# **CPLD Libraries Guide**

ISE 10.1

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# **About this Guide**

This guide is part of the ISE documentation collection and covers the use of Xilinx design elements in schematics. A separate version of this guide is also available if you prefer to work with Verilog or VHDL in your circuit design activities.

This guide contains the following:

- A general introduction to the design elements, including descriptions of the types of elements available in this architecture.
- A list of pre-existing design elements are automatically changed by the ISE software tools when they are used in this architecture, thus ensuring that you are always able to take full advantage of the latest circuit design advances.
- A list of the design elements that are supported in this architecture, organized by functional categories. Click on the element of your choice to immediately access its profile.
- Individual profiles describing each of the primitives and macros, and including, as appropriate, for each element:
- Its formal name
- A brief introduction to each element, including the names of all architectures in which it is supported
- Its schematic symbol
- Logic tables (if any)
- Port descriptions (if any)
- A list of available attributes
- VHDL and Verilog instantiation code
- References to any additional sources of information

#### **About this Architecture**

This version of the Libraries Guide describes the categories of design elements that comprise the Xilinx Unified Libraries for this architecture. These categories are:

- **Primitives** The simplest design elements in the Xilinx libraries. Primitives are the design element "atoms." Primitives can be created from primitives or macros. Examples of Xilinx primitives are the simple buffer, BUF, and the D flip-flop with clock enable and clear, FDCE.
- Macros The design element "molecules" of the Xilinx libraries. Macros can be created from the design element primitives or macros. For example, the FD4CE flip-flop macro is a composite of 4 FDCE primitives.

Xilinx maintains software libraries with hundreds of functional design elements (unimacros and primitives) for different device architectures. New functional elements are assembled with each release of development system software. In addition to a comprehensive Unified Library containing all design elements, this guide is one in a series of architecture-specific libraries.

# **Functional Categories**

This section categorizes, by function, the circuit design elements described in detail later in this guide. The elements (*primitives* and *macros*) are listed in alphanumeric order under each functional category.

Arithmetic Decoder Logic
Buffer Flip Flop Mux

Clock Divider General Shift Register

Comparator IO Shifter

Counter Latch

#### **Arithmetic**

Design Element	Description
ACC1	Macro: 1-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset
ACC16	Macro: 16-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset
ACC4	Macro: 4-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset
ACC8	Macro: 8-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset
ADD1	Macro: 1-Bit Full Adder with Carry-In and Carry-Out
ADD16	Macro: 16-Bit Cascadable Full Adder with Carry-In, Carry-Out, and Overflow
ADD4	Macro: 4-Bit Cascadable Full Adder with Carry-In, Carry-Out, and Overflow
ADD8	Macro: 8-Bit Cascadable Full Adder with Carry-In, Carry-Out, and Overflow
ADSU1	Macro: 1-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out
ADSU16	Macro: 16-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out, and Overflow
ADSU4	Macro: 4-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out, and Overflow
ADSU8	Macro: 8-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out, and Overflow

#### **Buffer**

Design Element	Description
BUF	Primitive: General Purpose Buffer
BUF16	Macro: 16-Bit General Purpose Buffer
BUF4	Macro: 4-Bit General Purpose Buffer
BUF8	Macro: 8-Bit General Purpose Buffer
BUFE	Primitive: Internal 3-State Buffer with Active High Enable
BUFE16	Macro: 16-Bit Internal 3-State Buffer with Active High Enable



Design Element	Description
BUFE4	Macro: 4-BitInternal 3-State Buffer with Active High Enable
BUFE8	Macro: 8-Bit Internal 3-State Buffer with Active High Enable
BUFG	Primitive: Global Clock Buffer
BUFGSR	Primitive: Global Set/Reset Input Buffer
BUFGTS	Primitive: Global 3-State Input Buffer
BUFT	Primitive: Internal 3-State Buffer with Active Low Enable
BUFT16	Macro: 16-Bit Internal 3-State Buffers with Active Low Enable
BUFT4	Macro: 4-Bit Internal 3-State Buffers with Active Low Enable
BUFT8	Macro: 8-Bit Internal 3-State Buffers with Active Low Enable

## **Clock Divider**

Design Element	Description
CLK_DIV10	Primitive: Simple Global Clock Divide by 10
CLK_DIV10R	Primitive: Global Clock Divide by 10 with Synchronous Reset
CLK_DIV10RSD	Primitive: Global Clock Divide by 10 with Synchronous Reset and Start Delay
CLK_DIV10SD	Primitive: Global Clock Divide by 10 with Start Delay
CLK_DIV12	Primitive: Simple Global Clock Divide by 12
CLK_DIV12R	Primitive: Global Clock Divide by 12 with Synchronous Reset
CLK_DIV12RSD	Primitive: Global Clock Divide by 12 with Synchronous Reset and Start Delay
CLK_DIV12SD	Primitive: Global Clock Divide by 12 with Start Delay
CLK_DIV14	Primitive: Simple Global Clock Divide by 14
CLK_DIV14R	Primitive: Global Clock Divide by 14 with Synchronous Reset
CLK_DIV14RSD	Primitive: Global Clock Divide by 14 with Synchronous Reset and Start Delay
CLK_DIV14SD	Primitive: Global Clock Divide by 14 with Start Delay
CLK_DIV16	Primitive: Simple Global Clock Divide by 16
CLK_DIV16R	Primitive: Global Clock Divide by 16 with Synchronous Reset
CLK_DIV16RSD	Primitive: Global Clock Divide by 16 with Synchronous Reset and Start Delay
CLK_DIV16SD	Primitive: Global Clock Divide by 16 with Start Delay
CLK_DIV2	Primitive: Simple Global Clock Divide by 2
CLK_DIV2R	Primitive: Global Clock Divide by 2 with Synchronous Reset
CLK_DIV2RSD	Primitive: Global Clock Divide by 2 with Synchronous Reset and Start Delay
CLK_DIV2SD	Primitive: Global Clock Divide by 2 with Start Delay
CLK_DIV4	Primitive: Simple Global Clock Divide by 4
CLK_DIV4R	Primitive: Global Clock Divide by 4 with Synchronous Reset
CLK_DIV4RSD	Primitive: Global Clock Divide by 4 with Synchronous Reset and Start Delay
CLK_DIV4SD	Primitive: Global Clock Divide by 4 with Start Delay
CLK_DIV6	Primitive: Simple Global Clock Divide by 6



Design Element	Description
CLK_DIV6R	Primitive: Global Clock Divide by 6 with Synchronous Reset
CLK_DIV6RSD	Primitive: Global Clock Divide by 6 with Synchronous Reset and Start Delay
CLK_DIV6SD	Primitive: Global Clock Divide by 6 with Start Delay
CLK_DIV8	Primitive: Simple Global Clock Divide by 8
CLK_DIV8R	Primitive: Global Clock Divide by 8 with Synchronous Reset
CLK_DIV8RSD	Primitive: Global Clock Divide by 8 with Synchronous Reset and Start Delay
CLK_DIV8SD	Primitive: Global Clock Divide by 8 with Start Delay

## Comparator

Design Element	Description
COMP16	Macro: 16-Bit Identity Comparator
COMP2	Macro: 2-Bit Identity Comparator
COMP4	Macro: 4-Bit Identity Comparator
COMP8	Macro: 8-Bit Identity Comparator
COMPM16	Macro: 16-Bit Magnitude Comparator
COMPM2	Macro: 2-Bit Magnitude Comparator
COMPM4	Macro: 4-Bit Magnitude Comparator
COMPM8	Macro: 8-Bit Magnitude Comparator

### Counter

Design Element	Description
CB16CE	Macro: 16-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear
CB16CLE	Macro: 16-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear
CB16CLED	Macro: 16-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear
CB16RE	Macro: 16-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB16RLE	Macro: 16-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB16X1	Macro: 16-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear
CB16X2	Macro: 16-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchro-nous Reset
CB2CE	Macro: 2-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear
CB2CLE	Macro: 2-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear
CB2CLED	Macro: 2-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear
CB2RE	Macro: 2-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset



Design Element	Description
CB2RLE	Macro: 2-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB2X1	Macro: 2-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear
CB2X2	Macro: 2-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchronous Reset
CB4CE	Macro: 4-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear
CB4CLE	Macro: 4-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear
CB4CLED	Macro: 4-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear
CB4RE	Macro: 4-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB4RLE	Macro: 4-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB4X1	Macro: 4-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear
CB4X2	Macro: 4-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchronous Reset
CB8CE	Macro: 8-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear
CB8CLE	Macro: 8-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear
CB8CLED	Macro: 8-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear
CB8RE	Macro: 8-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB8RLE	Macro: 8-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset
CB8X1	Macro: 8-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear
CB8X2	Macro: 8-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchronous Reset
CBD16CE	Macro: 16-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD16CLE	Macro: 16-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD16CLED	Macro: 16-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD16RE	Macro: 16-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD16RLE	Macro: 16-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD16X1	Macro: 16-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD16X2	Macro: 16-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD2CE	Macro: 2-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
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Design Element	Description
CBD2CLE	Macro: 2-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD2CLED	Macro: 2-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD2RE	Macro: 2-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD2RLE	Macro: 2-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD2X1	Macro: 2-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD2X2	Macro: 2-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD4CE	Macro: 4-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD4CLE	Macro: 4-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD4CLED	Macro: 4-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD4RE	Macro: 4-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD4RLE	Macro: 4-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD4X1	Macro: 4-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD4X2	Macro: 4-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD8CE	Macro: 8-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD8CLE	Macro: 8-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD8CLED	Macro: 8-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD8RE	Macro: 8-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD8RLE	Macro: 8-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CBD8X1	Macro: 8-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear
CBD8X2	Macro: 8-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset
CD4CE	Macro: 4-Bit Cascadable BCD Counter with Clock Enable and Asynchronous Clear
CD4CLE	Macro: 4-Bit Loadable Cascadable BCD Counter with Clock Enable and Asynchronous Clear
CD4RE	Macro: 4-Bit Cascadable BCD Counter with Clock Enable and Synchronous Reset
CD4RLE	Macro: 4-Bit Loadable Cascadable BCD Counter with Clock Enable and Synchronous Reset



Design Element	Description
CDD4CE	Macro: 4-Bit Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Asynchronous Clear
CDD4CLE	Macro: 4-Bit Loadable Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Asynchronous Clear
CDD4RE	Macro: 4-Bit Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Synchronous Reset
CDD4RLE	Macro: 4-Bit Loadable Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Synchronous Reset
CJ4CE	Macro: 4-Bit Johnson Counter with Clock Enable and Asynchronous Clear
CJ4RE	Macro: 4-Bit Johnson Counter with Clock Enable and Synchronous Reset
CJ5CE	Macro: 5-Bit Johnson Counter with Clock Enable and Asynchronous Clear
CJ5RE	Macro: 5-Bit Johnson Counter with Clock Enable and Synchronous Reset
CJ8CE	Macro: 8-Bit Johnson Counter with Clock Enable and Asynchronous Clear
CJ8RE	Macro: 8-Bit Johnson Counter with Clock Enable and Synchronous Reset
CJD4CE	Macro: 4-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Asynchronous Clear
CJD4RE	Macro: 4-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Synchronous Reset
CJD5CE	Macro: 5-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Asynchronous Clear
CJD5RE	Macro: 5-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Synchronous Reset
CJD8CE	Macro: 8-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Asynchronous Clear
CJD8RE	Macro: 8-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Synchronous Reset
CR16CE	Macro: 16-Bit Negative-Edge Binary Ripple Counter with Clock Enable and Asynchronous Clear
CR8CE	Macro: 8-Bit Negative-Edge Binary Ripple Counter with Clock Enable and Asynchronous Clear
CRD16CE	Macro: 16-Bit Dual-Edge Triggered Binary Ripple Counter with Clock Enable and Asynchronous Clear
CRD8CE	Macro: 8-Bit Dual-Edge Triggered Binary Ripple Counter with Clock Enable and Asynchronous Clear

### **Decoder**

Design Element	Description
D2_4E	Macro: 2- to 4-Line Decoder/Demultiplexer with Enable
D3_8E	Macro: 3- to 8-Line Decoder/Demultiplexer with Enable
D4_16E	Macro: 4- to 16-Line Decoder/Demultiplexer with Enable



## Flip Flop

Design Element	Description
FD	Macro: D Flip-Flop
FD16	Macro: Multiple D Flip-Flop
FD16CE	Macro: 16-Bit Data Register with Clock Enable and Asynchronous Clear
FD16RE	Macro: 16-Bit Data Register with Clock Enable and Synchronous Reset
FD4	Macro: Multiple D Flip-Flop
FD4CE	Macro: 4-Bit Data Register with Clock Enable and Asynchronous Clear
FD4RE	Macro: 4-Bit Data Register with Clock Enable and Synchronous Reset
FD8	Macro: Multiple D Flip-Flop
FD8CE	Macro: 8-Bit Data Register with Clock Enable and Asynchronous Clear
FD8RE	Macro: 8-Bit Data Register with Clock Enable and Synchronous Reset
FDC	Macro: D Flip-Flop with Asynchronous Clear
FDCE	Primitive: D Flip-Flop with Clock Enable and Asynchronous Clear
FDCP	Primitive: D Flip-Flop with Asynchronous Preset and Clear
FDCPE	Primitive: D Flip-Flop with Clock Enable and Asynchronous Preset and Clear
FDD	Macro: Dual Edge Triggered D Flip-Flop
FDD16	Macro: Multiple Dual Edge Triggered D Flip-Flop
FDD16CE	Macro: 16-Bit Dual Edge Triggered Data Register with Clock Enable and Asynchronous Clear
FDD16RE	Macro: 16-Bit Dual Edge Triggered Data Register with Clock Enable and Synchronous Reset
FDD4	Macro: Multiple Dual Edge Triggered D Flip-Flop
FDD4CE	Macro: 4-Bit Dual Edge Triggered Data Register with Clock Enable and Asynchronous Clear
FDD4RE	Macro: 4-Bit Dual Edge Triggered Data Register with Clock Enable and Synchronous Reset
FDD8	Macro: Multiple Dual Edge Triggered D Flip-Flop
FDD8CE	Macro: 8-Bit Dual Edge Triggered Data Register with Clock Enable and Asynchronous Clear
FDD8RE	Macro: 8-Bit Dual Edge Triggered Data Register with Clock Enable and Synchronous Reset
FDDC	Macro: D Dual Edge Triggered Flip-Flop with Asynchronous Clear
FDDCE	Primitive: Dual Edge Triggered D Flip-Flop with Clock Enable and Asynchronous Clear
FDDCP	Primitive: Dual Edge Triggered D Flip-Flop Asynchronous Preset and Clear
FDDCPE	Macro: Dual Edge Triggered D Flip-Flop with Clock Enable and Asynchronous Preset and Clear
FDDP	Macro: Dual Edge Triggered D Flip-Flop with Asynchronous Preset
FDDPE	Primitive: Dual Edge Triggered D Flip-Flop with Clock Enable and Asynchronous Preset
FDDR	Macro: Dual Edge Triggered D Flip-Flop with Synchronous Reset



Design Element	Description
FDDRE	Macro: Dual Edge Triggered D Flip-Flop with Clock Enable and Synchronous Reset
FDDRS	Macro: Dual Edge Triggered D Flip-Flop with Synchronous Reset and Set
FDDRSE	Macro: Dual Edge Triggered D Flip-Flop with Synchronous Reset and Set and Clock Enable
FDDS	Macro: Dual Edge Triggered D Flip-Flop with Synchronous Set
FDDSE	Macro: D Flip-Flop with Clock Enable and Synchronous Set
FDDSR	Macro: Dual Edge Triggered D Flip-Flop with Synchronous Set and Reset
FDDSRE	Macro: Dual Edge Triggered D Flip-Flop with Synchronous Set and Reset and Clock Enable
FDP	Macro: D Flip-Flop with Asynchronous Preset
FDPE	Primitive: D Flip-Flop with Clock Enable and Asynchronous Preset
FDR	Macro: D Flip-Flop with Synchronous Reset
FDRE	Macro: D Flip-Flop with Clock Enable and Synchronous Reset
FDRS	Macro: D Flip-Flop with Synchronous Reset and Set
FDRSE	Macro: D Flip-Flop with Synchronous Reset and Set and Clock Enable
FDS	Macro: D Flip-Flop with Synchronous Set
FDSE	Macro: D Flip-Flop with Clock Enable and Synchronous Set
FDSR	Macro: D Flip-Flop with Synchronous Set and Reset
FDSRE	Macro: D Flip-Flop with Synchronous Set and Reset and Clock Enable
FJKC	Macro: J-K Flip-Flop with Asynchronous Clear
FJKCE	Macro: J-K Flip-Flop with Clock Enable and Asynchronous Clear
FJKCP	Macro: J-K Flip-Flop with Asynchronous Clear and Preset
FJKCPE	Macro: J-K Flip-Flop with Asynchronous Clear and Preset and Clock Enable
FJKP	Macro: J-K Flip-Flop with Asynchronous Preset
FJKPE	Macro: J-K Flip-Flop with Clock Enable and Asynchronous Preset
FJKRSE	Macro: J-K Flip-Flop with Clock Enable and Synchronous Reset and Set
FJKSRE	Macro: J-K Flip-Flop with Clock Enable and Synchronous Set and Reset
FTC	Macro: Toggle Flip-Flop with Asynchronous Clear
FTCE	Macro: Toggle Flip-Flop with Clock Enable and Asynchronous Clear
FTCLE	Macro: Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear
FTCLEX	Macro: Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear
FTCP	Primitive: Toggle Flip-Flop with Asynchronous Clear and Preset
FTCPE	Macro: Toggle Flip-Flop with Clock Enable and Asynchronous Clear and Preset
FTCPLE	Macro: Loadable Toggle Flip-Flop with Clock Enable and Asynchronous Clear and Preset
FTDCE	Macro: Dual-Edge Triggered Toggle Flip-Flop with Clock Enable and Asynchronous Clear
FTDCLE	Macro: Dual-Edge Triggered Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear



Design Element	Description
FTDCLEX	Macro: Dual-Edge Triggered Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear
FTDCP	Primitive: Dual-Edge Triggered Toggle Flip-Flop with Asynchronous Clear and Preset
FTDRSE	Macro: Dual-Edge Triggered Toggle Flip-Flop with Synchronous Reset, Set, and Clock Enable
FTDRSLE	Macro: Dual-Edge Triggered Toggle Flip-Flop with Clock Enable and Synchronous Reset and Set
FTP	Macro: Toggle Flip-Flop with Asynchronous Preset
FTPE	Macro: Toggle Flip-Flop with Clock Enable and Asynchronous Preset
FTPLE	Macro: Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Preset
FTRSE	Macro: Toggle Flip-Flop with Clock Enable and Synchronous Reset and Set
FTRSLE	Macro: Toggle/Loadable Flip-Flop with Clock Enable and Synchronous Reset and Set
FTSRE	Macro: Toggle Flip-Flop with Clock Enable and Synchronous Set and Reset
FTSRLE	Macro: Toggle/Loadable Flip-Flop with Clock Enable and Synchronous Set and Reset

### General

Design Element	Description
GND	Primitive: Ground-Connection Signal Tag
KEEPER	Primitive: KEEPER Symbol
PULLDOWN	Primitive: Resistor to GND for Input Pads, Open-Drain, and 3-State Outputs
PULLUP	Primitive: Resistor to VCC for Input PADs, Open-Drain, and 3-State Outputs
VCC	Primitive: VCC-Connection Signal Tag

### 10

Design Element	Description
IBUF	Primitive: Input Buffer
IBUF16	Macro: 16-Bit Input Buffer
IBUF4	Macro: 4-Bit Input Buffer
IBUF8	Macro: 8-Bit Input Buffer
IOBUFE	Primitive: Bi-Directional Buffer
OBUF	Primitive: Output Buffer
OBUF16	Macro: 16-Bit Output Buffer
OBUF4	Macro: 4-Bit Output Buffer
OBUF8	Macro: 8-Bit Output Buffer
OBUFE	Macro: 3-State Output Buffer with Active-High Output Enable
OBUFE16	Macro: 16-Bit 3-State Output Buffer with Active-High Output Enable
OBUFE4	Macro: 4-Bit 3-State Output Buffer with Active-High Output Enable
OBUFE8	Macro: 8-Bit 3-State Output Buffer with Active-High Output Enable



Design Element	Description
OBUFT	Primitive: 3-State Output Buffer with Active Low Output Enable
OBUFT16	Macro: 16-Bit 3-State Output Buffer with Active Low Output Enable
OBUFT4	Macro: 4-Bit 3-State Output Buffers with Active-Low Output Enable
OBUFT8	Macro: 8-Bit 3-State Output Buffers with Active-Low Output Enable

#### Latch

Design Element	Description
LD	Macro: Transparent Data Latch
LD16	Macro: Multiple Transparent Data Latch
LD4	Macro: Multiple Transparent Data Latch
LD8	Macro: Multiple Transparent Data Latch
LDC	Macro: Transparent Data Latch with Asynchronous Clear
LDCP	Macro: Transparent Data Latch with Asynchronous Clear and Preset
LDG	Primitive: Transparent Datagate Latch
LDG16	Macro: 16-bit Transparent Datagate Latch
LDG4	Macro: 4-Bit Transparent Datagate Latch
LDG8	Macro: 8-Bit Transparent Datagate Latch
LDP	Macro: Transparent Data Latch with Asynchronous Preset

# Logic

Design Element	Description
AND2	Primitive: 2-Input AND Gate with Non-Inverted Inputs
AND2B1	Primitive: 2-Input AND Gate with 1 Inverted and 1 Non-Inverted Inputs
AND2B2	Primitive: 2-Input AND Gate with Inverted Inputs
AND3	Primitive: 3-Input AND Gate with Non-Inverted Inputs
AND3B1	Primitive: 3-Input AND Gate with 1 Inverted and 2 Non-Inverted Inputs
AND3B2	Primitive: 3-Input AND Gate with 2 Inverted and 1 Non-Inverted Inputs
AND3B3	Primitive: 3-Input AND Gate with Inverted Inputs
AND4	Primitive: 4-Input AND Gate with Non-Inverted Inputs
AND4B1	Primitive: 4-Input AND Gate with 1 Inverted and 3 Non-Inverted Inputs
AND4B2	Primitive: 4-Input AND Gate with 2 Inverted and 2 Non-Inverted Inputs
AND4B3	Primitive: 4-Input AND Gate with 3 Inverted and 1 Non-Inverted Inputs
AND4B4	Primitive: 4-Input AND Gate with Inverted Inputs
AND5	Primitive: 5-Input AND Gate with Non-Inverted Inputs
AND5B1	Primitive: 5-Input AND Gate with 1 Inverted and 4 Non-Inverted Inputs
AND5B2	Primitive: 5-Input AND Gate with 2 Inverted and 3 Non-Inverted Inputs



AND583 Primitive: 5-Input AND Gate with 3 Inverted and 2 Non-Inverted Inputs AND585 Primitive: 5-Input AND Gate with Inverted Inputs AND585 Primitive: 5-Input AND Gate with Inverted Inputs AND6 Macro: 6-Input AND Gate with Non-Inverted Inputs AND7 Macro: 7-Input AND Gate with Non-Inverted Inputs AND8 Macro: 8-Input AND Gate with Non-Inverted Inputs AND8 Macro: 9-Input AND Gate with Non-Inverted Inputs AND9 Macro: 9-Input AND Gate with Non-Inverted Inputs INV Primitive: Inverter INV16 Macro: 16 Inverters INV4 Macro: Four Inverters INV8 Macro: Four Inverters INV8 Macro: Four Inverters INV8 Macro: Fight Inverter INV8 Macro: Fight Inverters INV8 Macro: Fight Inverters INV8 Macro: Fight Inverters INAND2 Primitive: 2-Input NAND Gate with Non-Inverted Inputs INAND2B1 Primitive: 2-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND3B1 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B1 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B1 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B2 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with Non-Inverted Inputs INAND4 Primitive: 4-Input NAND Gate with Inverted Inputs INAND4B1 Primitive: 4-Input NAND Gate with Inverted Inputs INAND4B2 Primitive: 4-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND4B9 Primitive: 4-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND5B9 Primitive: 5-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND5B9 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B9 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B9 Primitive: 5-Input NAND Gate w	Design Element	Description
AND585 Primitive: 5-Input AND Gate with Inverted Inputs AND6 Macro: 6-Input AND Gate with Non-Inverted Inputs AND7 Macro: 8-Input AND Gate with Non-Inverted Inputs AND8 Macro: 8-Input AND Gate with Non-Inverted Inputs AND9 Macro: 9-Input AND Gate with Non-Inverted Inputs INV Primitive: Inverter INV16 Macro: 16 Inverters INV4 Macro: 16 Inverters INV8 Macro: Eight Inverters INV8 Macro: Eight Inverters INV8 Macro: Eight Inverters INV8 Macro: Four Inverters INV8 Macro: Eight Inverters INV8 Macro: Four Inverters INV8 Macro: Fight Inverters INV8 Macro: Four Inverters INV8 Macro: Four Inverters INV8 Macro: Four Inverters INV8 Macro: Four Inverters INV8 Macro: Fight Inverters INV8 Macro: Four Inverters INV8 Macro: Fight Inverters INV8 Macro: Fight Inverters INV8 Macro: Fight Inverted Inputs INAND2B1 Primitive: 2-Input NAND Gate with Inverted Inputs Invert	AND5B3	Primitive: 5-Input AND Gate with 3 Inverted and 2 Non-Inverted Inputs
AND6 Macro: 6-Input AND Gate with Non-Inverted Inputs AND7 Macro: 7-Input AND Gate with Non-Inverted Inputs AND8 Macro: 8-Input AND Gate with Non-Inverted Inputs AND9 Macro: 9-Input AND Gate with Non-Inverted Inputs INV Primitive: Inverter INV16 Macro: 16 Inverters INV4 Macro: 16 Inverters INV4 Macro: Four Inverters INV8 Macro: Fight Inverters INV8 Macro: Four Inverted Inputs INAND2 Primitive: 2-Input NAND Gate with Inverted Inputs INAND3 Primitive: 2-Input NAND Gate with Inverted Inputs INAND3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3 Primitive: 3-Input NAND Gate with Inverted and 2 Non-Inverted Inputs INAND3B1 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs INAND4 Primitive: 4-Input NAND Gate with Inverted and 3 Non-Inverted Inputs INAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND4B4 Primitive: 4-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND5B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B3 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with No	AND5B4	Primitive: 5-Input AND Gate with 4 Inverted and 1 Non-Inverted Inputs
AND7 Macro: 7-Input AND Gate with Non-Inverted Inputs AND8 Macro: 8-Input AND Gate with Non-Inverted Inputs AND9 Macro: 9-Input AND Gate with Non-Inverted Inputs INV Primitive: Inverter INV16 Macro: 16 Inverters INV4 Macro: Four Inverters INV8 Macro: Eight Inverted Inputs INAND28 Primitive: 2-Input NAND Gate with Non-Inverted Inputs INAND28 Primitive: 2-Input NAND Gate with Inverted and 1 Non-Inverted Inputs INAND3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B2 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND4B1 Primitive: 4-Input NAND Gate with Inverted Inputs INAND4B1 Primitive: 4-Input NAND Gate with Inverted and 3 Non-Inverted Inputs INAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 3 Non-Inverted Inputs INAND4B4 Primitive: 4-Input NAND Gate with 3 Inverted and 4 Non-Inverted Inputs INAND4B4 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs INAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs INAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs INAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs INAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6B Macro: 5-Input NAND Gate with Non-Inverted Inputs INAND6B Macro:	AND5B5	Primitive: 5-Input AND Gate with Inverted Inputs
AND8 Macro: 8-Input AND Gate with Non-Inverted Inputs  NAD9 Macro: 9-Input AND Gate with Non-Inverted Inputs  NAV Primitive: Inverter  INV16 Macro: 16 Inverters  INV4 Macro: 16 Inverters  INV8 Macro: Fight Inverters  INV8 Macro: Eight Inverters  INV8 Macro: Eight Inverters  INV8 Macro: Fight Inverters  NAND2 Primitive: 2-Input NAND Gate with Non-Inverted Inputs  NAND2 Primitive: 2-Input NAND Gate with Inverted and 1 Non-Inverted Inputs  NAND2B1 Primitive: 2-Input NAND Gate with Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Inverted and 3 Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND6B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6B4 Primitive: 5-Input NAND Gat	AND6	Macro: 6-Input AND Gate with Non-Inverted Inputs
AND9 Macro: 9-Input AND Gate with Non-Inverted Inputs  INV Primitive: Inverter  INV16 Macro: 16 Inverters  INV4 Macro: Four Inverters  INV8 Macro: Eight Inverters  INV8 Macro: Eight Inverters  INV8 Macro: Eight Inverters  INV8 Macro: 2-Input NAND Gate with Non-Inverted Inputs  NAND2 Primitive: 2-Input NAND Gate with Inverted and 1 Non-Inverted Inputs  NAND2B1 Primitive: 2-Input NAND Gate with Inverted Inputs  NAND2B2 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with 1 Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with 1 Inverted Inputs  NAND4B4 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 9-Input NAND Gate with Non	AND7	Macro: 7-Input AND Gate with Non-Inverted Inputs
INV Primitive: Inverter INV16 Macro: 16 Inverters INV4 Macro: Four Inverters INV8 Macro: Eight Inverters Inverted Inputs INAND3B1 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B2 Primitive: 4-Input NAND Gate with Inverted Inputs INAND4B1 Primitive: 4-Input NAND Gate with Inverted and 3 Non-Inverted Inputs INAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs INAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs INAND4B4 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND5 Primitive: 5-Input NAND Gate with 3 Inverted and 4 Non-Inverted Inputs INAND5B1 Primitive: 5-Input NAND Gate with 3 Inverted and 4 Non-Inverted Inputs INAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs INAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND6B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6B5 Primitive: 3-Input NAND Gate with Non-Inverted Inputs INAND6B6 Macro: 8-Input NAND Gate with Non-Inverted Inputs INAND8 Macro: 9-In	AND8	Macro: 8-Input AND Gate with Non-Inverted Inputs
INV16 Macro: 16 Inverters INV4 Macro: Four Inverters INV8 Macro: Eight Inverted Inputs INAND2B1 Primitive: 2-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs INAND3B2 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs INAND3B2 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs INAND4B4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs INAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs INAND4B2 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs INAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted Inputs INAND4B4 Primitive: 4-Input NAND Gate with 1 Inverted Inputs INAND4B4 Primitive: 5-Input NAND Gate with Inverted Inputs INAND5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND5 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs INAND5B2 Primitive: 5-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs INAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted Inputs INAND5B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs INAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs INAND6 Macro: 8-Input NAND Gate with Non-Inverted Inputs INAND6 Primitive: 2-Input NAND Gate with Non-Inverted Inputs INAND6 Primitive: 2-In	AND9	Macro: 9-Input AND Gate with Non-Inverted Inputs
INV4 Macro: Four Inverters  INV8 Macro: Eight Inverters  NAND2 Primitive: 2-Input NAND Gate with Non-Inverted Inputs  NAND2B1 Primitive: 2-Input NAND Gate with I Inverted and 1 Non-Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Inverted and 2 Non-Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 3 Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND4B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B6 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B6 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 5-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 5-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 5-Input NAND Gate with Non-Inverted Inputs  NAND9 Primitive: 2-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted and 1 Non-I	INV	Primitive: Inverter
NAND2	INV16	Macro: 16 Inverters
NAND2 Primitive: 2-Input NAND Gate with Non-Inverted Inputs  NAND2B1 Primitive: 2-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND2B2 Primitive: 2-Input NAND Gate with Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with 1 Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with 3 Inverted Inputs  NAND4B4 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 4 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6B Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND6 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NAND6 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted Inputs	INV4	Macro: Four Inverters
NAND2B1 Primitive: 2-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs  NAND3 Primitive: 2-Input NAND Gate with Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 3 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 3 Inverted and 4 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6B Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted and 1 Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs	INV8	Macro: Eight Inverters
NAND2B2 Primitive: 2-Input NAND Gate with Inverted Inputs  NAND3 Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 3 Inverted and 4 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted and 1 Non-Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs	NAND2	Primitive: 2-Input NAND Gate with Non-Inverted Inputs
Primitive: 3-Input NAND Gate with Non-Inverted Inputs  NAND3B1 Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs  NAND3B2 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs  NAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted and 1 Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 2-Input NOR Gate with Inverted Inputs	NAND2B1	Primitive: 2-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs
NAND3B1         Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs           NAND3B2         Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs           NAND3B3         Primitive: 3-Input NAND Gate with Inverted Inputs           NAND4         Primitive: 4-Input NAND Gate with Non-Inverted Inputs           NAND4B1         Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs           NAND4B2         Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs           NAND4B3         Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs           NAND4B4         Primitive: 4-Input NAND Gate with 1 Inverted Inputs           NAND5         Primitive: 5-Input NAND Gate with Non-Inverted Inputs           NAND5B1         Primitive: 5-Input NAND Gate with 2 Inverted and 4 Non-Inverted Inputs           NAND5B2         Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs           NAND5B3         Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs           NAND5B4         Primitive: 5-Input NAND Gate with Non-Inverted Inputs           NAND6         Macro: 6-Input NAND Gate with Non-Inverted Inputs           NAND6         Macro: 7-Input NAND Gate with Non-Inverted Inputs           NAND8         Macro: 8-Input NAND Gate with Non-Inverted Inputs           NAND9         Macro: 9-Input NAND Gate with Non-Inverted Inputs <td>NAND2B2</td> <td>Primitive: 2-Input NAND Gate with Inverted Inputs</td>	NAND2B2	Primitive: 2-Input NAND Gate with Inverted Inputs
NAND3B2 Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs  NAND4 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND4B4 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 2 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B6 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B6 Macro: 6-Input NAND Gate with Inverted Inputs  NAND6 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Inverted and 1 Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs	NAND3	Primitive: 3-Input NAND Gate with Non-Inverted Inputs
NAND3B3 Primitive: 3-Input NAND Gate with Inverted Inputs  NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 1 Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs	NAND3B1	Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs
NAND4 Primitive: 4-Input NAND Gate with Non-Inverted Inputs  NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND3B2	Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs
NAND4B1 Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs  NAND4B2 Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 4 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs	NAND3B3	Primitive: 3-Input NAND Gate with Inverted Inputs
Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs  NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 2-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Inverted Inputs	NAND4	Primitive: 4-Input NAND Gate with Non-Inverted Inputs
NAND4B3 Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs  NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND4B1	Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs
NAND4B4 Primitive: 4-Input NAND Gate with Inverted Inputs  NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND4B2	Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs
NAND5 Primitive: 5-Input NAND Gate with Non-Inverted Inputs  NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND4B3	Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs
NAND5B1 Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs  NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs  NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND4B4	Primitive: 4-Input NAND Gate with Inverted Inputs
NAND5B2 Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs NAND9 NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs NOR2 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs NOR2B1 Primitive: 2-Input NOR Gate with Inverted Inputs NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND5	Primitive: 5-Input NAND Gate with Non-Inverted Inputs
NAND5B3 Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs  NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND5B1	Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs
NAND5B4 Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs  Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND5B2	Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs
NAND5B5 Primitive: 5-Input NAND Gate with Inverted Inputs  NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with I Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND5B3	Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs
NAND6 Macro: 6-Input NAND Gate with Non-Inverted Inputs  NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND5B4	Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs
NAND7 Macro: 7-Input NAND Gate with Non-Inverted Inputs  NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND5B5	Primitive: 5-Input NAND Gate with Inverted Inputs
NAND8 Macro: 8-Input NAND Gate with Non-Inverted Inputs  NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND6	Macro: 6-Input NAND Gate with Non-Inverted Inputs
NAND9 Macro: 9-Input NAND Gate with Non-Inverted Inputs  NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND7	Macro: 7-Input NAND Gate with Non-Inverted Inputs
NOR2 Primitive: 2-Input NOR Gate with Non-Inverted Inputs  NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND8	Macro: 8-Input NAND Gate with Non-Inverted Inputs
NOR2B1 Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs  NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NAND9	Macro: 9-Input NAND Gate with Non-Inverted Inputs
NOR2B2 Primitive: 2-Input NOR Gate with Inverted Inputs  NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NOR2	Primitive: 2-Input NOR Gate with Non-Inverted Inputs
NOR3 Primitive: 3-Input NOR Gate with Non-Inverted Inputs	NOR2B1	Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs
	NOR2B2	Primitive: 2-Input NOR Gate with Inverted Inputs
NOR3B1 Primitive: 3-Input NOR Gate with 1 Inverted and 2 Non-Inverted Inputs	NOR3	Primitive: 3-Input NOR Gate with Non-Inverted Inputs
	NOR3B1	Primitive: 3-Input NOR Gate with 1 Inverted and 2 Non-Inverted Inputs



Design Element	Description
NOR3B2	Primitive: 3-Input NOR Gate with 2 Inverted and 1 Non-Inverted Inputs
NOR3B3	Primitive: 3-Input NOR Gate with Inverted Inputs
NOR4	Primitive: 4-Input NOR Gate with Non-Inverted Inputs
NOR4B1	Primitive: 4-Input NOR Gate with 1 Inverted and 3 Non-Inverted Inputs
NOR4B2	Primitive: 4-Input NOR Gate with 2 Inverted and 2 Non-Inverted Inputs
NOR4B3	Primitive: 4-Input NOR Gate with 3 Inverted and 1 Non-Inverted Inputs
NOR4B4	Primitive: 4-Input NOR Gate with Inverted Inputs
NOR5	Primitive: 5-Input NOR Gate with Non-Inverted Inputs
NOR5B1	Primitive: 5-Input NOR Gate with 1 Inverted and 4 Non-Inverted Inputs
NOR5B2	Primitive: 5-Input NOR Gate with 2 Inverted and 3 Non-Inverted Inputs
NOR5B3	Primitive: 5-Input NOR Gate with 3 Inverted and 2 Non-Inverted Inputs
NOR5B4	Primitive: 5-Input NOR Gate with 4 Inverted and 1 Non-Inverted Inputs
NOR5B5	Primitive: 5-Input NOR Gate with Inverted Inputs
NOR6	Macro: 6-Input NOR Gate with Non-Inverted Inputs
NOR7	Macro: 7-Input NOR Gate with Non-Inverted Inputs
NOR8	Macro: 8-Input NOR Gate with Non-Inverted Inputs
NOR9	Macro: 9-Input NOR Gate with Non-Inverted Inputs
OR2	Primitive: 2-Input OR Gate with Non-Inverted Inputs
OR2B1	Primitive: 2-Input OR Gate with 1 Inverted and 1 Non-Inverted Inputs
OR2B2	Primitive: 2-Input OR Gate with Inverted Inputs
OR3	Primitive: 3-Input OR Gate with Non-Inverted Inputs
OR3B1	Primitive: 3-Input OR Gate with 1 Inverted and 2 Non-Inverted Inputs
OR3B2	Primitive: 3-Input OR Gate with 2 Inverted and 1 Non-Inverted Inputs
OR3B3	Primitive: 3-Input OR Gate with Inverted Inputs
OR4	Primitive: 4-Input OR Gate with Non-Inverted Inputs
OR4B1	Primitive: 4-Input OR Gate with 1 Inverted and 3 Non-Inverted Inputs
OR4B2	Primitive: 4-Input OR Gate with 2 Inverted and 2 Non-Inverted Inputs
OR4B3	Primitive: 4-Input OR Gate with 3 Inverted and 1 Non-Inverted Inputs
OR4B4	Primitive: 4-Input OR Gate with Inverted Inputs
OR5	Primitive: 5-Input OR Gate with Non-Inverted Inputs
OR5B1	Primitive: 5-Input OR Gate with 1 Inverted and 4 Non-Inverted Inputs
OR5B2	Primitive: 5-Input OR Gate with 2 Inverted and 3 Non-Inverted Inputs
OR5B3	Primitive: 5-Input OR Gate with 3 Inverted and 2 Non-Inverted Inputs
OR5B4	Primitive: 5-Input OR Gate with 4 Inverted and 1 Non-Inverted Inputs
OR5B5	Primitive: 5-Input OR Gate with Inverted Inputs
OR6	Macro: 6-Input OR Gate with Non-Inverted Inputs
OR7	Macro: 7-Input OR Gate with Non-Inverted Inputs
OR8	Macro: 8-Input OR Gate with Non-Inverted Inputs



Design Element	Description
OR9	Macro: 9-Input OR Gate with Non-Inverted Inputs
XNOR2	Primitive: 2-Input XNOR Gate with Non-Inverted Inputs
XNOR3	Primitive: 3-Input XNOR Gate with Non-Inverted Inputs
XNOR4	Primitive: 4-Input XNOR Gate with Non-Inverted Inputs
XNOR5	Primitive: 5-Input XNOR Gate with Non-Inverted Inputs
XNOR6	Macro: 6-Input XNOR Gate with Non-Inverted Inputs
XNOR7	Macro: 7-Input XNOR Gate with Non-Inverted Inputs
XNOR8	Macro: 8-Input XNOR Gate with Non-Inverted Inputs
XNOR9	Macro: 9-Input XNOR Gate with Non-Inverted Inputs
XOR2	Primitive: 2-Input XOR Gate with Non-Inverted Inputs
XOR3	Primitive: 3-Input XOR Gate with Non-Inverted Inputs
XOR4	Primitive: 4-Input XOR Gate with Non-Inverted Inputs
XOR5	Primitive: 5-Input XOR Gate with Non-Inverted Inputs
XOR6	Macro: 6-Input XOR Gate with Non-Inverted Inputs
XOR7	Macro: 7-Input XOR Gate with Non-Inverted Inputs
XOR8	Macro: 8-Input XOR Gate with Non-Inverted Inputs
XOR9	Macro: 9-Input XOR Gate with Non-Inverted Inputs

### Mux

Design Element	Description
M16_1E	Macro: 16-to-1 Multiplexer with Enable
M2_1	Macro: 2-to-1 Multiplexer
M2_1B1	Macro: 2-to-1 Multiplexer with D0 Inverted
M2_1B2	Macro: 2-to-1 Multiplexer with D0 and D1 Inverted
M2_1E	Macro: 2-to-1 Multiplexer with Enable
M4_1E	Macro: 4-to-1 Multiplexer with Enable
M8_1E	Macro: 8-to-1 Multiplexer with Enable

# **Shift Register**

Design Element	Description
SR16CE	Macro: 16-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear
SR16CLE	Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear
SR16CLED	Macro: 16-Bit Shift Register with Clock Enable and Asynchronous Clear
SR16RE	Macro: 16-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



Design Element	Description
SR16RLE	Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset
SR16RLED	Macro: 16-Bit Shift Register with Clock Enable and Synchronous Reset
SR4CE	Macro: 4-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear
SR4CLE	Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear
SR4CLED	Macro: 4-Bit Shift Register with Clock Enable and Asynchronous Clear
SR4RE	Macro: 4-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset
SR4RLE	Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset
SR4RLED	Macro: 4-Bit Shift Register with Clock Enable and Synchronous Reset
SR8CE	Macro: 8-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear
SR8CLE	Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear
SR8CLED	Macro: 8-Bit Shift Register with Clock Enable and Asynchronous Clear
SR8RE	Macro: 8-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset
SR8RLE	Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset
SR8RLED	Macro: 8-Bit Shift Register with Clock Enable and Synchronous Reset
SRD16CE	Macro: 16-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD16CLE	Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD16CLED	Macro: 16-Bit Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD16RE	Macro: 16-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD16RLE	Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD16RLED	Macro: 16-Bit Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD4CE	Macro: 4-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD4CLE	Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD4CLED	Macro: 4-Bit Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD4RE	Macro: 4-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD4RLE	Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



Design Element	Description
SRD4RLED	Macro: 4-Bit Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD8CE	Macro: 8-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD8CLE	Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD8CLED	Macro: 8-Bit Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear
SRD8RE	Macro: 8-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD8RLE	Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset
SRD8RLED	Macro: 8-Bit Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset

## Shifter

Design Element	Description
BRLSHFT4	Macro: 4-Bit Barrel Shifter
BRLSHFT8	Macro: 8-Bit Barrel Shifter

# **About Design Elements**

This section describes the design elements that can be used with this architecture. The design elements are organized alphabetically.

The following information is provided for each design element, where applicable:

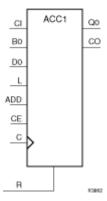
- Name of element
- Brief description
- Schematic symbol (if any)
- Logic Table (if any)
- Port Descriptions (if any)
- Usage
- Available Attributes (if any)
- For more information

You can find examples of VHDL and Verilog instantiation code in the ISE software (in the main menu, select **Edit** > **Language Templates** or in the *Libraries Guide for HDL Designs* for this architecture.



#### ACC1

Macro: 1-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element can add or subtract a 1-bit unsigned-binary word to or from the contents of a 1-bit data register and store the results in the register. The register can be loaded with a 1-bit word. The synchronous reset (R) has priority over all other inputs and, when High, causes the output to go to logic level zero during the Low-to-High clock (C) transition. Clock (C) transitions are ignored when clock enable (CE) is Low.

Load

When the load input (L) is High, CE is ignored and the data on the input D0 is loaded into the 1-bit register during the Low-to-High clock (C) transition.

Add

When control inputs ADD and CE are both High, the accumulator adds a 1-bit word (B0) and carry-in (CI) to the contents of the 1-bit register. The result is stored in the register and appears on output Q0 during the Low-to-High clock transition. The carry-out (CO) is not registered synchronously with the data output. CO always reflects the accumulation of input B0 and the contents of the register, which allows cascading of ACC1s by connecting CO of one stage to CI of the next stage. In add mode, CO acts as a carry-out, and CO and CI are active-High.

Subtract

When ADD is Low and CE is High, the 1-bit word B0 and CI are subtracted from the contents of the register. The result is stored in the register and appears on output Q0 during the Low-to-High clock transition. The carry-out (CO) is not registered synchronously with the data output. CO always reflects the accumulation of input B0 and the contents of the register, which allows cascading of ACC1s by connecting CO of one stage to CI of the next stage. In subtract mode, CO acts as a borrow, and CO and CI are active-Low.

This design element is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Design Entry Method**

This design element is only for use in schematics.



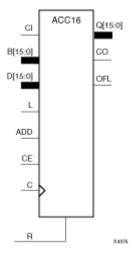
#### **For More Information**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### ACC<sub>16</sub>

Macro: 16-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element can add or subtract a 16-bit unsigned-binary, respectively or twos-complement word to or from the contents of a 16-bit data register and store the results in the register. The register can be loaded with the 16-bit word.

When the load input (L) is High, CE is ignored and the data on the D inputs is loaded into the register during the Low-to-High clock (C) transition. ACC16 loads the data on inputs D15 – D0 into the 16-bit register.

This design element operates on either 16-bit unsigned binary numbers or 16-bit twos-complement numbers. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is how they determine when "overflow" occurs. Unsigned binary uses carry-out (CO), while twos complement uses OFL to determine when "overflow" occurs.

• For unsigned binary operation, ACC16 can represent numbers between 0 and 15, inclusive. In add mode, CO is active (High) when the sum exceeds the bounds of the adder/subtracter. In subtract mode, CO is an active-Low borrow-out and goes Low when the difference exceeds the bounds. The carry-out (CO) is not registered synchronously with the data outputs. CO always reflects the accumulation of the B inputs (B15 – B0 for ACC16). This allows the cascading of ACC16s by connecting CO of one stage to CI of the next stage. An unsigned binary "overflow" that is always active-High can be generated by gating the ADD signal and CO as follows:

unsigned overflow = CO XOR ADD

Ignore OFL in unsigned binary operation.

For twos-complement operation, ACC16 represents numbers between -8 and +7, inclusive. If an addition
or subtraction operation result exceeds this range, the OFL output goes High. The overflow (OFL) is not
registered synchronously with the data outputs. OFL always reflects the accumulation of the B inputs (B15 –



B0 for ACC16) and the contents of the register, which allows cascading of ACC4s by connecting OFL of one stage to CI of the next stage.

Ignore CO in twos-complement operation.

The synchronous reset (R) has priority over all other inputs, and when set to High, causes all outputs to go to logic level zero during the Low-to-High clock (C) transition. Clock (C) transitions are ignored when clock enable (CE) is Low.

This design element is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### **Logic Table**

Input						Output
R	L	CE	ADD	D	С	Q
1	х	х	х	х	Rising	0
0	1	х	х	Dn	Rising	Dn
0	0	1	1	х	Rising	Q0+Bn+CI
0	0	1	0	х	Rising	Q0-Bn-CI
0	0	0	х	х	Rising	No Change

Q0: Previous value of Q

Bn: Value of Data input B

CI: Value of input CI

#### **Design Entry Method**

This design element is only for use in schematics.

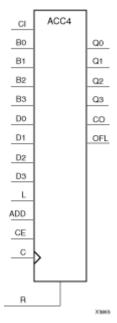
#### For More Information

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### ACC4

Macro: 4-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element can add or subtract a 4-bit unsigned-binary, respectively or twos-complement word to or from the contents of a 4-bit data register and store the results in the register. The register can be loaded with the 4-bit word.

When the load input (L) is High, CE is ignored and the data on the D inputs is loaded into the register during the Low-to-High clock (C) transition. ACC4 loads the data on inputs D3 – D0 into the 4-bit register.

This design element operates on either 4-bit unsigned binary numbers or 4-bit twos-complement numbers. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is how they determine when "overflow" occurs. Unsigned binary uses carry-out (CO), while twos complement uses OFL to determine when "overflow" occurs.

• For unsigned binary operation, ACC4 can represent numbers between 0 and 15, inclusive. In add mode, CO is active (High) when the sum exceeds the bounds of the adder/subtracter. In subtract mode, CO is an active-Low borrow-out and goes Low when the difference exceeds the bounds. The carry-out (CO) is not registered synchronously with the data outputs. CO always reflects the accumulation of the B inputs (B3 – B0 for ACC4). This allows the cascading of ACC4s by connecting CO of one stage to CI of the next stage. An unsigned binary "overflow" that is always active-High can be generated by gating the ADD signal and CO as follows:

unsigned overflow = CO XOR ADD

Ignore OFL in unsigned binary operation.



• For twos-complement operation, ACC4 represents numbers between -8 and +7, inclusive. If an addition or subtraction operation result exceeds this range, the OFL output goes High. The overflow (OFL) is not registered synchronously with the data outputs. OFL always reflects the accumulation of the B inputs (B3 – B0 for ACC4) and the contents of the register, which allows cascading of ACC4s by connecting OFL of one stage to CI of the next stage.

Ignore CO in twos-complement operation.

The synchronous reset (R) has priority over all other inputs, and when set to High, causes all outputs to go to logic level zero during the Low-to-High clock (C) transition. Clock (C) transitions are ignored when clock enable (CE) is Low.

This design element is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Input						Output
R	L	CE	ADD	D	С	Q
1	х	х	х	х	Rising	0
0	1	х	х	Dn	Rising	Dn
0	0	1	1	х	Rising	Q0+Bn+CI
0	0	1	0	х	Rising	Q0-Bn-CI
0	0	0	х	х	Rising	No Change

Q0: Previous value of Q

Bn: Value of Data input B

CI: Value of input CI

#### **Design Entry Method**

This design element is only for use in schematics.

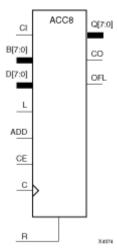
#### For More Information

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### ACC8

Macro: 8-Bit Loadable Cascadable Accumulator with Carry-In, Carry-Out, and Synchronous Reset



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element can add or subtract a 8-bit unsigned-binary, respectively or twos-complement word to or from the contents of a 8-bit data register and store the results in the register. The register can be loaded with the 8-bit word.

When the load input (L) is High, CE is ignored and the data on the D inputs is loaded into the register during the Low-to-High clock (C) transition. ACC8 loads the data on inputs D7 – D0 into the 8-bit register.

This design element operates on either 8-bit unsigned binary numbers or 8-bit twos-complement numbers. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is how they determine when "overflow" occurs. Unsigned binary uses carry-out (CO), while twos complement uses OFL to determine when "overflow" occurs.

• For unsigned binary operation, ACC8 can represent numbers between 0 and 255, inclusive. In add mode, CO is active (High) when the sum exceeds the bounds of the adder/subtracter. In subtract mode, CO is an active-Low borrow-out and goes Low when the difference exceeds the bounds. The carry-out (CO) is not registered synchronously with the data outputs. CO always reflects the accumulation of the B inputs (B3 – B0 for ACC4). This allows the cascading of ACC8s by connecting CO of one stage to CI of the next stage. An unsigned binary "overflow" that is always active-High can be generated by gating the ADD signal and CO as follows:

unsigned overflow = CO XOR ADD

Ignore OFL in unsigned binary operation.

• For twos-complement operation, ACC8 represents numbers between -128 and +127, inclusive. If an addition or subtraction operation result exceeds this range, the OFL output goes High. The overflow (OFL) is not registered synchronously with the data outputs. OFL always reflects the accumulation of the B inputs (B3 –



B0 for ACC8) and the contents of the register, which allows cascading of ACC8s by connecting OFL of one stage to CI of the next stage.

Ignore CO in twos-complement operation.

The synchronous reset (R) has priority over all other inputs, and when set to High, causes all outputs to go to logic level zero during the Low-to-High clock (C) transition. Clock (C) transitions are ignored when clock enable (CE) is Low.

This design element is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Input					Output	
R	L	CE	ADD	D	С	Q
1	х	х	х	х	Rising	0
0	1	х	х	Dn	Rising	Dn
0	0	1	1	х	Rising	Q0+Bn+CI
0	0	1	0	х	Rising	Q0-Bn-CI
0	0	0	х	х	Rising	No Change

Q0: Previous value of Q

Bn: Value of Data input B

CI: Value of input CI

## **Design Entry Method**

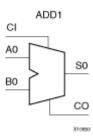
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## ADD1

### Macro: 1-Bit Full Adder with Carry-In and Carry-Out



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a cascadable 1-bit full adder with carry-in and carry-out. It adds two 1-bit words (A and B) and a carry-in (CI), producing a binary sum (S0) output and a carry-out (CO).

## **Logic Table**

Inputs			Outputs	Outputs	
A0	ВО	CI	S0	СО	
0	0	0	0	0	
1	0	0	1	0	
0	1	0	1	0	
1	1	0	0	1	
0	0	1	1	0	
1	0	1	0	1	
0	1	1	0	1	
1	1	1	1	1	

# **Design Entry Method**

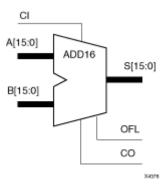
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### ADD16

#### Macro: 16-Bit Cascadable Full Adder with Carry-In, Carry-Out, and Overflow



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element adds two words and a carry-in (CI), producing a sum output and carry-out (CO) or overflow (OFL). The factors added are A15 – A0, B15 – B0 and CI, producing the sum output S15 – S0 and CO (or OFL).

#### Logic Table

Input	Output		
A B		S	
An Bn		An+Bn+CI	
CI: Value of input CI.			

### **Unsigned Binary Versus Twos Complement**

This design element can operate on either 16-bit unsigned binary numbers or 16-bit twos-complement numbers, respectively. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is the way they determine when "overflow" occurs. Unsigned binary uses CO, while twos-complement uses OFL to determine when "overflow" occurs. To interpret the inputs as unsigned binary, follow the CO output. To interpret the inputs as twos complement, follow the OFL output.

### **Unsigned Binary Operation**

For unsigned binary operation, this element represents numbers between 0 and 65535, inclusive. OFL is ignored in unsigned binary operation.

#### **Twos-Complement Operation**

For twos-complement operation, this element can represent numbers between -32768 and +32767, inclusive. OFL is active (High) when the sum exceeds the bounds of the adder. CO is ignored in twos-complement operation.



# **Design Entry Method**

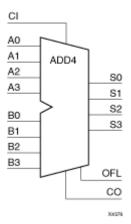
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### ADD4

#### Macro: 4-Bit Cascadable Full Adder with Carry-In, Carry-Out, and Overflow



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element adds two words and a carry-in (CI), producing a sum output and carry-out (CO) or overflow (OFL). The factors added are A3 – A0, B3 – B0, and CI producing the sum output S3 – S0 and CO (or OFL).

#### Logic Table

Input	Output		
A B		S	
An Bn		An+Bn+CI	
CI: Value of input CI.			

### **Unsigned Binary Versus Twos Complement**

This design element can operate on either 4-bit unsigned binary numbers or 4-bit twos-complement numbers, respectively. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is the way they determine when "overflow" occurs. Unsigned binary uses CO, while twos-complement uses OFL to determine when "overflow" occurs. To interpret the inputs as unsigned binary, follow the CO output. To interpret the inputs as twos complement, follow the OFL output.

### **Unsigned Binary Operation**

For unsigned binary operation, this element represents numbers from 0 to 15, inclusive. OFL is ignored in unsigned binary operation.



# **Twos-Complement Operation**

For twos-complement operation, this element can represent numbers between -8 and +7, inclusive. OFL is active (High) when the sum exceeds the bounds of the adder. CO is ignored in twos-complement operation.

### **Design Entry Method**

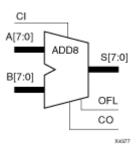
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### ADD8

#### Macro: 8-Bit Cascadable Full Adder with Carry-In, Carry-Out, and Overflow



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element adds two words and a carry-in (CI), producing a sum output and carry-out (CO) or overflow (OFL). The factors added are A7 – A0, B7 – B0, and CI, producing the sum output S7 – S0 and CO (or OFL).

### **Logic Table**

Input	Output			
A B		S		
An Bn		An+Bn+CI		
CI: Value of input CI.	CI: Value of input CI.			

### **Unsigned Binary Versus Twos Complement**

This design element can operate on either 8-bit unsigned binary numbers or 8-bit twos-complement numbers, respectively. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is the way they determine when "overflow" occurs. Unsigned binary uses CO, while twos-complement uses OFL to determine when "overflow" occurs. To interpret the inputs as unsigned binary, follow the CO output. To interpret the inputs as twos complement, follow the OFL output.

## **Unsigned Binary Operation**

For unsigned binary operation, this element represents numbers between 0 and 255, inclusive. OFL is ignored in unsigned binary operation.

### **Twos-Complement Operation**

For twos-complement operation, this element can represent numbers between -128 and +127, inclusive. OFL is active (High) when the sum exceeds the bounds of the adder. CO is ignored in twos-complement operation.



# **Design Entry Method**

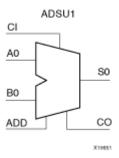
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# ADSU1

#### Macro: 1-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

When the ADD input is High, two 1-bit words (A0 and B0) are added with a carry-in (CI), producing a 1-bit output (S0) and a carry-out (CO). When the ADD input is Low, B0 is subtracted from A0, producing a result (S0) and borrow (CO). In add mode, CO represents a carry-out, and CO and CI are active-High. In subtract mode, CO represents a borrow, and CO and CI are active-Low.

Add Function, ADD=1

Inputs			Outputs	
A0	В0	CI	S0	СО
0	0	0	0	0
0	1	0	1	0
1	0	0	1	0
1	1	0	0	1
0	0	1	1	0
0	1	1	0	1
1	0	1	0	1
1	1	1	1	1

Subtract Function, ADD=0

Inputs			Outputs	
A0	B0	CI	S0	СО
0	0	0	1	0
0	1	0	0	0
1	0	0	0	1
1	1	0	1	0
0	0	1	0	1



Inputs			Outputs	
A0	B0	CI	S0	СО
0	1	1	1	0
1	0	1	1	1
1	1	1	0	1
1	0	1	1	1
1	1	1	0	1

# **Design Entry Method**

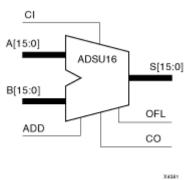
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## ADSU16

#### Macro: 16-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out, and Overflow



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

When the ADD input is High, ADSU16 adds two 16-bit words (A15 – A0 and B15 – B0) and a CI are added, producing a 16-bit sum output (S15 – S0) and CO or OFL. ADSU4 adds two 4-bit words (A3 – A0 and B3 – B0) and a CI, producing a 4-bit sum output (S3 – S0) and CO or OFL. ADSU8 adds two 8-bit words (A7 – A0 and B7 – B0) and a CI producing, an 8-bit sum output (S7 – S0) and CO or OFL. ADSU16 adds two 16-bit words (A15 – A0 and B15 – B0) and a CI, producing a 16-bit sum output (S15 – S0) and CO or OFL.

When the ADD input is Low, this element subtracts Bz – B0 from Az– A0, producing a difference output and CO or OFL. It subtracts B3 – B0 from A3 – A0, producing a 4-bit difference (S3 – S0) and CO or OFL.

In add mode, CO and CI are active-High. In subtract mode, CO and CI are active-Low. OFL is active-High in add and subtract modes.

## Logic Table

Input	Output				
ADD	A	В	S		
1	An	Bn	An+Bn+CI*		
0	An	Bn	An-Bn-CI*		
CI*: ADD = 0, CI, CO active LOW					
CI*: ADD = 1, CI, CO active I	HIGH				

Unsigned Binary Versus Twos Complement

This design element can operate on either 16-bit unsigned binary numbers or 16-bit twos-complement numbers. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is the way they determine when "overflow" occurs. Unsigned binary uses CO, while twos complement uses OFL to determine when "overflow" occurs.



With adder/subtracters, either unsigned binary or twos-complement operations cause an overflow. If the result crosses the overflow boundary, an overflow is generated. Similarly, when the result crosses the carry-out boundary, a carry-out is generated.

Unsigned Binary Operation

For unsigned binary operation, this element can represent numbers between 0 and 65535, inclusive. In add mode, CO is active (High) when the sum exceeds the bounds of the adder/subtracter. In subtract mode, CO is an active-Low borrow-out and goes Low when the difference exceeds the bounds.

An unsigned binary "overflow" that is always active-High can be generated by gating the ADD signal and CO as follows:

unsigned overflow = CO XOR ADD

OFL is ignored in unsigned binary operation.

Twos-Complement Operation

For twos-complement operation, this element can represent numbers between -32768 and +32767, inclusive.

If an addition or subtraction operation result exceeds this range, the OFL output goes High. CO is ignored in twos-complement operation.

### **Design Entry Method**

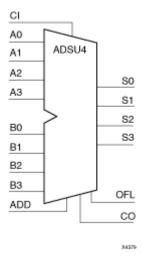
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# ADSU4

Macro: 4-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out, and Overflow



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

When the ADD input is High, ADSU4 adds two 4-bit words (A3 - A0 and B3 - B0) and a CI are added, producing a 4-bit sum output (S3 - S0) and CO or OFL. For this element, two 4-bit words (A3 - A0 and B3 - B0) and a CI are added, producing a 4-bit sum output (S3 - S0) and CO or OFL

When the ADD input is Low, this element subtracts Bz - B0 from Az- A0, producing a 4-bit difference (S3 – S0) and CO or OFL.

In add mode, CO and CI are active-High. In subtract mode, CO and CI are active-Low. OFL is active-High in add and subtract modes.

### **Logic Table**

Input	Output				
ADD	A	В	S		
1	An	Bn	An+Bn+CI*		
0	An	Bn	An-Bn-CI*		
CI*: ADD = 0, CI, CO active LOW					
CI*: ADD = 1, CI, CO active	CI*: ADD = 1, CI, CO active HIGH				

Unsigned Binary Versus Twos Complement



This design element can operate on either 4-bit unsigned binary numbers or 4-bit twos-complement numbers. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is the way they determine when "overflow" occurs. Unsigned binary uses CO, while twos complement uses OFL to determine when "overflow" occurs.

With adder/subtracters, either unsigned binary or twos-complement operations cause an overflow. If the result crosses the overflow boundary, an overflow is generated. Similarly, when the result crosses the carry-out boundary, a carry-out is generated.

Unsigned Binary Operation

For unsigned binary operation, ADSU4 can represent numbers between 0 and 15, inclusive. In add mode, CO is active (High) when the sum exceeds the bounds of the adder/subtracter. In subtract mode, CO is an active-Low borrow-out and goes Low when the difference exceeds the bounds.

An unsigned binary "overflow" that is always active-High can be generated by gating the ADD signal and CO as follows:

unsigned overflow = CO XOR ADD

OFL is ignored in unsigned binary operation.

Twos-Complement Operation

For twos-complement operation, this element can represent numbers between -8 and +7, inclusive.

If an addition or subtraction operation result exceeds this range, the OFL output goes High. CO is ignored in twos-complement operation.

### **Design Entry Method**

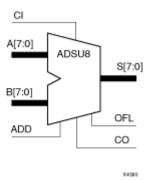
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## ADSU8

#### Macro: 8-Bit Cascadable Adder/Subtracter with Carry-In, Carry-Out, and Overflow



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

When the ADD input is High, ADSU8 adds two 8-bit words (A7 - A0 and B7 - B0) and a CI are added, producing, an 8-bit sum output (S7 - S0) and CO or OFL. For this element, two 8-bit words (A7 - A0 and B7 - B0) and a CI producing, an 8-bit sum output (S7 - S0) and CO or OFL.

When the ADD input is Low, this element subtracts B7 - B0 from A7 - A0, producing an 8-bit difference (S7 - S0) and CO or OFL.

In add mode, CO and CI are active-High. In subtract mode, CO and CI are active-Low. OFL is active-High in add and subtract modes.

### **Logic Table**

Input	Output			
ADD	A	В	S	
1	An	Bn	An+Bn+CI*	
0	An	Bn	An-Bn-CI*	
CI*: ADD = 0, CI, CO active LOW				
CI*: ADD = 1, CI, CO active HIGH				

Unsigned Binary Versus Twos Complement

This design element can operate on either 8-bit unsigned binary numbers or 8-bit twos-complement numbers. If the inputs are interpreted as unsigned binary, the result can be interpreted as unsigned binary. If the inputs are interpreted as twos complement, the output can be interpreted as twos complement. The only functional difference between an unsigned binary operation and a twos-complement operation is the way they determine when "overflow" occurs. Unsigned binary uses CO, while twos complement uses OFL to determine when "overflow" occurs.

With adder/subtracters, either unsigned binary or twos-complement operations cause an overflow. If the result crosses the overflow boundary, an overflow is generated. Similarly, when the result crosses the carry-out boundary, a carry-out is generated.



Unsigned Binary Operation

For unsigned binary operation, this element can represent numbers between 0 and 255, inclusive. In add mode, CO is active (High) when the sum exceeds the bounds of the adder/subtracter. In subtract mode, CO is an active-Low borrow-out and goes Low when the difference exceeds the bounds.

An unsigned binary "overflow" that is always active-High can be generated by gating the ADD signal and CO as follows:

unsigned overflow = CO XOR ADD

OFL is ignored in unsigned binary operation.

Twos-Complement Operation

For twos-complement operation, this element can represent numbers between -128 and +127, inclusive.

If an addition or subtraction operation result exceeds this range, the OFL output goes High. CO is ignored in twos-complement operation.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND<sub>2</sub>

#### Primitive: 2-Input AND Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND2B1

#### Primitive: 2-Input AND Gate with 1 Inverted and 1 Non-Inverted Inputs

AND2B1



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND2B2

#### Primitive: 2-Input AND Gate with Inverted Inputs

AND2B2



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND3

#### Primitive: 3-Input AND Gate with Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## AND3B1

#### Primitive: 3-Input AND Gate with 1 Inverted and 2 Non-Inverted Inputs

AND3B1



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND3B2

#### Primitive: 3-Input AND Gate with 2 Inverted and 1 Non-Inverted Inputs

AND3B2



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## AND3B3

#### Primitive: 3-Input AND Gate with Inverted Inputs

AND3B3



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND4

#### Primitive: 4-Input AND Gate with Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 4-Input AND Gate with 1 Inverted and 3 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 4-Input AND Gate with 2 Inverted and 2 Non-Inverted Inputs

AND4B2



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 4-Input AND Gate with 3 Inverted and 1 Non-Inverted Inputs

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 4-Input AND Gate with Inverted Inputs

AND4B4

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

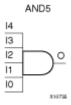
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND5

#### Primitive: 5-Input AND Gate with Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input AND Gate with 1 Inverted and 4 Non-Inverted Inputs

### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

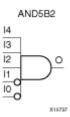
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input AND Gate with 2 Inverted and 3 Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

### **Design Entry Method**

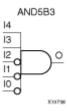
This design element is only for use in schematics.

#### **Available Attributes**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input AND Gate with 3 Inverted and 2 Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

# **Design Entry Method**

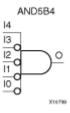
This design element is only for use in schematics.

#### **Available Attributes**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input AND Gate with 4 Inverted and 1 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input AND Gate with Inverted Inputs

AND5B5



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

### **Design Entry Method**

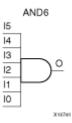
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND<sub>6</sub>

#### Macro: 6-Input AND Gate with Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

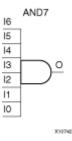
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### AND7

#### Macro: 7-Input AND Gate with Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

# **Design Entry Method**

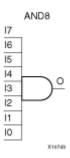
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## AND8

#### Macro: 8-Input AND Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

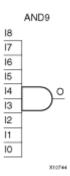
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## AND9

#### Macro: 9-Input AND Gate with Non-Inverted Inputs



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

AND functions of up to five inputs are available in any combination of inverting and non-inverting inputs. AND functions of six to nine inputs, 12 inputs, and 16 inputs are available with noninverting inputs. To make some or all inputs inverting, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the appropriate number of inputs.

## **Design Entry Method**

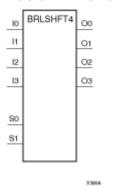
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **BRLSHFT4**

#### Macro: 4-Bit Barrel Shifter



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 4-bit barrel shifter that can rotate four inputs (I3 - I0) up to four places. The control inputs (S1 and S0) determine the number of positions, from one to four, that the data is rotated. The four outputs (O3 - O0) reflect the shifted data inputs.

#### **Logic Table**

Inputs				Outputs					
S1	S0	10	I1	I2	I3	O0	O1	O2	O3
0	0	a	b	С	d	a	b	С	d
0	1	a	b	С	d	b	С	d	a
1	0	a	b	С	d	С	d	a	b
1	1	a	b	С	d	d	a	b	С

## **Design Entry Method**

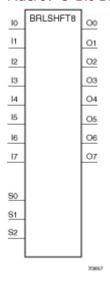
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **BRLSHFT8**

#### Macro: 8-Bit Barrel Shifter



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an 8-bit barrel shifter, can rotate the eight inputs (I7 - I0) up to eight places. The control inputs (S2 - S0) determine the number of positions, from one to eight, that the data is rotated. The eight outputs (O7 - O0) reflect the shifted data inputs.

## **Logic Table**

Inpu	Inputs				Outputs													
S2	S1	S0	10	I1	I2	<b>I</b> 3	<b>I4</b>	<b>I</b> 5	<b>I</b> 6	I7	O0	01	O2	О3	O4	O5	O6	<b>O</b> 7
0	0	0	a	b	С	d	e	f	g	h	a	b	С	d	e	f	g	h
0	0	1	a	b	С	d	e	f	g	h	b	С	d	e	f	g	h	a
0	1	0	a	b	С	d	e	f	g	h	С	d	e	f	g	h	a	b
0	1	1	a	b	С	d	e	f	g	h	d	e	f	g	h	a	b	С
1	0	0	a	b	С	d	e	f	g	h	e	f	g	h	a	b	С	d
1	0	1	a	b	С	d	e	f	g	h	f	g	h	a	b	С	d	e
1	1	0	a	b	С	d	e	f	g	h	g	h	a	b	С	d	e	f
1	1	1	a	b	С	d	e	f	g	h	h	a	b	С	d	e	f	g

## **Design Entry Method**

This design element is only for use in schematics.



- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: General Purpose Buffer

BUF

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This is a general-purpose, non-inverting buffer.

This element is not necessary and is removed by the partitioning software (MAP).

## **Design Entry Method**

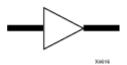
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



Macro: 16-Bit General Purpose Buffer

BUF16



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This is a 16-bit, general purpose, non-inverting buffer. In working with CPLDs, this element is usually removed, unless you inhibit optimization by applying the OPT=OFF attribute to the symbol, or by using the LOGIC\_OPT=OFF global attribute.

## **Design Entry Method**

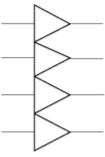
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 4-Bit General Purpose Buffer





## **Supported Architectures**

X4614

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This is a 4-bit, general purpose, non-inverting buffer. In working with CPLDs, this element is usually removed, unless you inhibit optimization by applying the OPT=OFF attribute to the symbol, or by using the LOGIC\_OPT=OFF global attribute.

# **Design Entry Method**

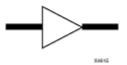
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 8-Bit General Purpose Buffer

BUF8



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This is a 8-bit, general purpose, non-inverting buffer. In working with CPLDs, this element is usually removed, unless you inhibit optimization by applying the OPT=OFF attribute to the symbol, or by using the LOGIC\_OPT=OFF global attribute.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: Internal 3-State Buffer with Active High Enable

E BUFE

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a single, 3-state buffer with inputs I and output O, and an active-High output enable (E). When E is High, data on the input of the buffer is transferred to the corresponding output. When E is Low, the output is high impedance (Z state or Off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures.

The outputs of separate symbols for this entity can be tied together to form a bus or a multiplexer. Make sure that only one E is High at any one time. If none of the E inputs is active-High, a "weak-keeper" circuit keeps the output bus from floating but does not guarantee that the bus remains at the last value driven onto it. For certain CPLD devices, output from nets assume the High logic level when all connected BUFE/BUFT buffers are disabled. For FPGA devices, elements need a PULLUP element connected to their output. NGDBuild inserts a PULLUP element if one is not connected.

## **Logic Table**

Inputs	Outputs	
E	I	0
0	X	Z
1	1	1
1	0	0

## **Design Entry Method**

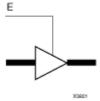
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 16-Bit Internal 3-State Buffer with Active High Enable

BUFE16



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a multiple 3-state buffer with inputs of I15 – I0 and outputs of O15 – O0 and an active-High output enable (E). When E is High, data on the inputs of the buffers is transferred to the corresponding outputs.

When E is Low, the output is high impedance (Z state or Off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures. The outputs of separate BUFE elements can be tied together to form a bus or a multiplexer. Make sure that only one E is High at any one time. If none of the E inputs is active-High, a "weak-keeper" circuit keeps the output bus from floating but does not guarantee that the bus remains at the last value driven onto it.

#### **Logic Table**

Inputs	Outputs	
Е	I	0
0	Х	Z
1	1	1
1	0	0

## **Design Entry Method**

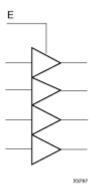
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 4-BitInternal 3-State Buffer with Active High Enable

BUFE4



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a multiple 3-state buffer with inputs of I3 - I0 and outputs of O3 - O0 and an active-High output enable (E). When E is High, data on the inputs of the buffers is transferred to the corresponding outputs.

When E is Low, the output is high impedance (Z state or Off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures. The outputs of separate BUFE elements can be tied together to form a bus or a multiplexer. Make sure that only one E is High at any one time. If none of the E inputs is active-High, a "weak-keeper" circuit keeps the output bus from floating but does not guarantee that the bus remains at the last value driven onto it.

## **Logic Table**

Inputs	Outputs	
E	I	0
0	Х	Z
1	1	1
1	0	0

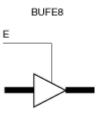
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 8-Bit Internal 3-State Buffer with Active High Enable



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a multiple 3-state buffer with inputs of I7 - I0 and outputs of O7 - O0 and an active-High output enable (E). When E is High, data on the inputs of the buffers is transferred to the corresponding outputs.

When E is Low, the output is high impedance (Z state or Off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures. The outputs of separate BUFE elements can be tied together to form a bus or a multiplexer. Make sure that only one E is High at any one time. If none of the E inputs is active-High, a "weak-keeper" circuit keeps the output bus from floating but does not guarantee that the bus remains at the last value driven onto it.

#### **Logic Table**

Inputs	Outputs	
E	I	0
0	Х	Z
1	1	1
1	0	0

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **BUFG**

#### Primitive: Global Clock Buffer

BUFG

1 O

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a high-fanout buffer that connects signals to the global routing resources for low skew distribution of the signal. BUFGs are typically used on clock nets as well other high fanout nets like sets/resets and clock enables.

## **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

## **Verilog Instantiation Template**

```
// BUFG: Global Clock Buffer (source by an internal signal)
// All FPGAs
// Xilinx HDL Libraries Guide, version 10.1.1
```



- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **BUFGSR**

Primitive: Global Set/Reset Input Buffer

BUFGSR

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element distributes global set/reset signals throughout selected flip-flops of an XC9500/XV/XL, CoolRunner XPLA3, or CoolRunner-II device. Global Set/Reset (GSR) control pins are available on these CPLD devices. Consult device data sheets for availability.

This design element always acts as an input buffer. To use it in a schematic, connect the input of the design element symbol to an IPAD or an IOPAD representing the GSR signal source. GSR signals generated on-chip must be passed through an OBUF-type buffer before they are connected to the design element.

For global set/reset control, the output of the design element normally connects to the CLR or PRE input of a flip-flop symbol, like FDCP, or any registered symbol with asynchronous clear or preset. The global set/reset control signal may pass through an inverter to perform an active-low set/reset. The output of the design element may also be used as an ordinary input signal to other logic elsewhere in the design. This design element can control any number of flip-flops in a design.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **BUFGTS**

Primitive: Global 3-State Input Buffer

BUFGTS

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element distributes global output-enable signals throughout the output pad drivers of CPLD devices. Global Three-State (GTS) control pins are available on these CPLD devices. Consult device data sheets for availability.

This element always acts as an input buffer. To use it in a schematic, connect the input of the BUFGTS symbol to an IPAD or an IOPAD representing the GTS signal source. GTS signals generated on-chip must be passed through an OBUF-type buffer before they are connected to this element.

For global 3-state control, the output of this element normally connects to the E input of a 3-state output buffer symbol, OBUFE. The global 3-state control signal may pass through an inverter or control an OBUFT symbol to perform an active-low output-enable. The same 3-state control signal may even be used both inverted and non-inverted to enable alternate groups of device outputs. The output of BUFGTS may also be used as an ordinary input signal to other logic elsewhere in the design. Each BUFGTS can control any number of output buffers in a design.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: Internal 3-State Buffer with Active Low Enable

BUFT



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a single 3-state buffer with input I and an output of O and active-Low output enable (T). When T is Low, data on the input of the buffer is transferred to the corresponding output. When T is High, the output is high impedance (Z state or off). The output of the buffer is connected to a horizontal longline in FPGA architectures.

The output of separate BUFT symbols can be tied together to form a bus or a multiplexer. Make sure that only one T is Low at one time. For CPLD devices, BUFT output nets assume the High logic level when all connected BUFE/BUFT buffers are disabled. For FPGAs, when all BUFTs on a net are disabled, the net is High. For correct simulation of this effect, a PULLUP element must be connected to the net. NGDBuild inserts a PULLUP element if one is not connected so that back-annotation simulation reflects the true state of the device.

#### **Logic Table**

Inputs	Outputs	
Т	I	0
1	X	Z
0	1	1
0	0	0

#### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 16-Bit Internal 3-State Buffers with Active Low Enable

BUFT1€



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a multiple 3-state buffer with inputs I15 - 10 and outputs O15 - O0 and active-Low output enable (T). When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (Z state or off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures.

The output of separate BUFT symbols can be tied together to form a bus or a multiplexer. Make sure that only one T is Low at one time. For CPLD devices, BUFT output nets assume the High logic level when all connected BUFE/BUFT buffers are disabled. For FPGAs, when all BUFTs on a net are disabled, the net is High. For correct simulation of this effect, a PULLUP element must be connected to the net. NGDBuild inserts a PULLUP element if one is not connected so that back-annotation simulation reflects the true state of the device.

# Logic Table

Inputs	Outputs	
Т	I	0
1	X	Z
0	1	1
0	0	0

## **Design Entry Method**

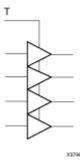
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 4-Bit Internal 3-State Buffers with Active Low Enable

BUFT4



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a multiple 3-state buffer with inputs I3 – I0 and outputs O3 – O0 and active-Low output enable (T). When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (Z state or off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures.

The output of separate BUFT symbols can be tied together to form a bus or a multiplexer. Make sure that only one T is Low at one time. For CPLD devices, BUFT output nets assume the High logic level when all connected BUFE/BUFT buffers are disabled. For FPGAs, when all BUFTs on a net are disabled, the net is High. For correct simulation of this effect, a PULLUP element must be connected to the net. NGDBuild inserts a PULLUP element if one is not connected so that back-annotation simulation reflects the true state of the device.

## **Logic Table**

Inputs	Outputs	
Т	I	0
1	Х	Z
0	1	1
0	0	0

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 8-Bit Internal 3-State Buffers with Active Low Enable





## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3

#### Introduction

This design element is a multiple 3-state buffer with inputs I7 – I0 and outputs O7 – O0 and active-Low output enable (T). When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (Z state or off). The outputs of the buffers are connected to horizontal longlines in FPGA architectures.

The output of separate BUFT symbols can be tied together to form a bus or a multiplexer. Make sure that only one T is Low at one time. For CPLD devices, BUFT output nets assume the High logic level when all connected BUFE/BUFT buffers are disabled. For FPGAs, when all BUFTs on a net are disabled, the net is High. For correct simulation of this effect, a PULLUP element must be connected to the net. NGDBuild inserts a PULLUP element if one is not connected so that back-annotation simulation reflects the true state of the device.

## **Logic Table**

Inputs	Outputs	
Т	I	0
1	X	Z
0	1	1
0	0	0

#### **Design Entry Method**

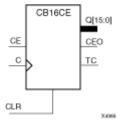
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB16CE

Macro: 16-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs			
CLR	CE	С	Qz-Q0	TC	CEO	
1	X	X	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

$$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$$

$$CEO = TC \cdot CE$$

#### **Design Entry Method**

This design element is only for use in schematics.

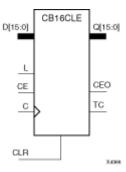


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB16CLE

Macro: 16-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This element is a synchronously loadable, asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs				
CLR	L	CE	С	Dz – D0	Qz – Q0	TC	CEO
1	Χ	Χ	X	Χ	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	X	X	No change	No change	0
0	0	1	-	X	Inc	TC	CEO

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

CEO = TC • CE



# **Design Entry Method**

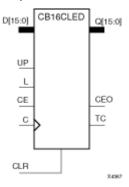
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB16CLED

Macro: 16-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

For CPLD parts, see "CB2X1", "CB4X1", "CB8X1", "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs			Outputs					
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO
1	Χ	Χ	X	Χ	Χ	0	0	0
0	1	X	-	X	Dn	Dn	TC	CEO



Inputs			Outputs					
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO
0	0	0	X	X	Χ	No change	No change	0
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO

z = bit width - 1

$$\mathsf{TC} = (\mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0 \bullet \mathsf{UP}) + (\mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0 \bullet \mathsf{UP})$$

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

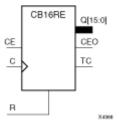
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB16RE

Macro: 16-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs			Outputs			
R	CE	С	Qz-Q0	TC	CEO	
1	X	-	0	0	0	
0	0	Х	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0)$ 

 $CEO = TC \cdot CE$ 

## **Design Entry Method**

This design element is only for use in schematics.

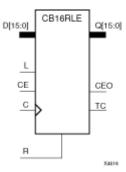


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CB16RLE

Macro: 16-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs				
R	L	CE	С	Dz – D0	Qz - Q0	TC	CEO
1	X	X	-	X	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	Χ	Χ	No change	No change	0



Inputs						Outputs			
R	L	CE	С	Dz – D0	Qz – Q0	TC	CEO		
0	0	1	-	X	Inc	TC	CEO		
z = bit wic	dth - 1	•					•		
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC	CEO = TC∙CE								

# **Design Entry Method**

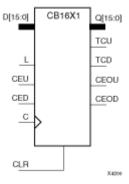
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **CB16X1**

Macro: 16-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, connect the CEOU and CEOD outputs of each counter directly to the CEU and CED inputs, respectively, of the next stage. Connect the clock, L, and CLR inputs in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs	Inputs							Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD		
1	Χ	Χ	Χ	Χ	Χ	0	0	1	0	CEOD		
0	1	X	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD		
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0		
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0		
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD		
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid		

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $TCD = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD = TCD $\bullet$ CED

# **Design Entry Method**

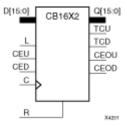
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **CB16X2**

Macro: 16-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchro-nous Reset



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading in CPLD architectures.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs	Inputs						Outputs				
R	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD	
1	X	Χ	Χ	-	X	0	0	1	0	CEOD	
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD	
0	0	0	0	Χ	Χ	No Chg	No Chg	No Chg	0	0	
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0	
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD	
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid	

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD =  $TCD \bullet CED$ 

# **Design Entry Method**

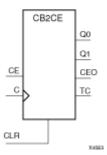
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CB2CE

Macro: 2-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear



#### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### **Logic Table**

Inputs			Outputs				
CLR	CE	С	Qz-Q0	TC	CEO		
1	X	Χ	0	0	0		
0	0	X	No change	No change	0		
0	1	-	Inc	TC	CEO		

z = bit width - 1

$$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0$$

CEO = TC • CE

#### **Design Entry Method**

This design element is only for use in schematics.

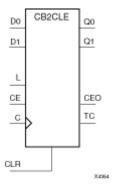


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **CB2CLE**

Macro: 2-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This element is a synchronously loadable, asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs					Outputs			
CLR	L	CE	С	Dz – D0	Qz - Q0	TC	CEO	
1	Х	Χ	Χ	X	0	0	0	
0	1	X	-	Dn	Dn	TC	CEO	
0	0	0	Х	Х	No change	No change	0	



Inputs						Outputs			
CLR	C				Dz - D0 Qz - Q0 TC CEO				
0	0	1	X	Inc	TC	CEO			
z = bit width	- 1								
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC•CE									

# **Design Entry Method**

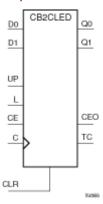
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB2CLED

Macro: 2-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

For CPLD parts, see "CB2X1", "CB4X1", "CB8X1", "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs					Outputs			
CLR L CE C UP Dz - D0					Qz - Q0	TC	CEO	
1	Χ	Χ	Χ	X	X	0	0	0



Inputs						Outputs			
CLR	L	CE	С	UP	Dz – D0	Qz - Q0	TC	CEO	
0	1	X	-	X	Dn	Dn	TC	CEO	
0	0	0	X	X	Χ	No change	No change	0	
0	0	1	-	1	X	Inc	TC	CEO	
0	0	1	-	0	X	Dec	TC	CEO	

z = bit width - 1

 $\mathsf{TC} = (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet UP) + (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet UP)$ 

CEO = TC•CE

# **Design Entry Method**

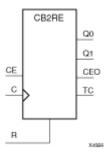
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB2RE

Macro: 2-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs			Outputs			
R	CE	С	Qz-Q0	TC	CEO	
1	X	-	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

$$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0)$$

CEO = TC • CE

## **Design Entry Method**

This design element is only for use in schematics.

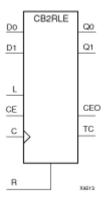


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CB2RLE**

Macro: 2-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Inputs					Outputs			
R	L	CE	С	Dz – D0	Qz – Q0	TC	CEO		
1	X	X	-	X	0	0	0		
0	1	X	-	Dn	Dn	TC	CEO		
0	0	0	Х	Х	No change	No change	0		



Inputs					Outputs			
R	L	CE	Qz - Q0	TC	CEO			
0	0	1	X	Inc	TC	CEO		
z = bit width -	z = bit width - 1							
$TC = Qz \bullet Q(z-$	1)•Q(z-2)••Q	0						

# **Design Entry Method**

CEO =  $TC \bullet CE$ 

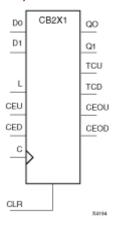
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB2X1

Macro: 2-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, connect the CEOU and CEOD outputs of each counter directly to the CEU and CED inputs, respectively, of the next stage. Connect the clock, L, and CLR inputs in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.



The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	Χ	Χ	X	Χ	Χ	0	0	1	0	CEOD
0	1	X	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $TCD = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD = TCD • CED

# **Design Entry Method**

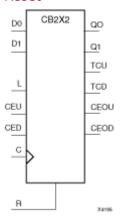
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB2X2

Macro: 2-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchronous Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading in CPLD architectures.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.



The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs						Outputs					
R	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD	
1	X	Χ	Х	-	X	0	0	1	0	CEOD	
0	1	X	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD	
0	0	0	0	Х	Χ	No Chg	No Chg	No Chg	0	0	
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0	
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD	
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid	

z = bit width - 1

 $TCU = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD = TCD • CED

# **Design Entry Method**

This design element is only for use in schematics.

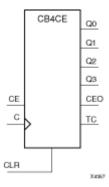
## **Available Attributes**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB4CE

Macro: 4-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs			
CLR	CE	С	Qz-Q0	TC	CEO	
1	X	X	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

 $TC = Qz \cdot Q(z-1) \cdot Q(z-2) \cdot ... \cdot Q0$ 

CEO = TC • CE



# **Design Entry Method**

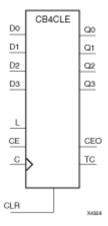
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CB4CLE**

Macro: 4-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This element is a synchronously loadable, asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs					Outputs			
CLR	L	CE	С	Dz – D0	Qz - Q0	TC	CEO	
1	Х	X	X	X	0	0	0	
0	1	X	-	Dn	Dn	TC	CEO	
0	0	0	Х	Х	No change	No change	0	



Inputs						Outputs			
CLR	R L CE C Dz – D0				Qz - Q0	TC	CEO		
0	0	1	X	Inc	TC	CEO			
z = bit width -	z = bit width - 1								
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC•CE									

# **Design Entry Method**

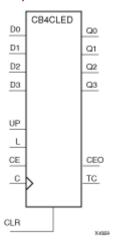
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB4CLED

Macro: 4-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

For CPLD parts, see "CB2X1", "CB4X1", "CB8X1", "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs	Inputs						Outputs		
CLR	L	CE	С	UP	Dz – D0	Qz - Q0	TC	CEO	
1	Х	Χ	Х	Х	Х	0	0	0	
0	1	X	-	X	Dn	Dn	TC	CEO	
0	0	0	Х	Х	Х	No change	No change	0	
0	0	1	-	1	X	Inc	TC	CEO	
0	0	1	-	0	X	Dec	TC	CEO	

z = bit width - 1

 $\mathsf{TC} = (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet \mathsf{UP}) + (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet \mathsf{UP})$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

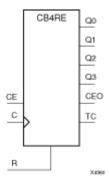
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB4RE

Macro: 4-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs			
R	CE	С	Qz-Q0	TC	CEO	
1	X	-	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0)$ 

CEO = TC • CE



# **Design Entry Method**

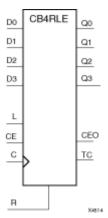
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CB4RLE**

Macro: 4-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs			Outputs				
R	L	CE	С	Dz – D0	Qz - Q0	TC	CEO
1	X	X	-	X	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	X	X	No change	No change	0



Inputs			Outputs					
R	L	CE	С	Dz – D0	Qz - Q0	TC	CEO	
0	0	1	-	X	Inc	TC	CEO	
z = bit width -	· 1							
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$								
CEO = TC•CE								

# **Design Entry Method**

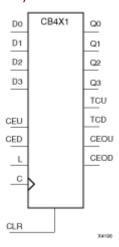
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB4X1

Macro: 4-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, connect the CEOU and CEOD outputs of each counter directly to the CEU and CED inputs, respectively, of the next stage. Connect the clock, L, and CLR inputs in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.



When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD	
1	Χ	Χ	Χ	Χ	Χ	0	0	1	0	CEOD	
0	1	Χ	Χ	-	Dn	Dn	TCU	TCD	CEOU	CEOD	
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0	
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0	
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD	
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid	

z = bit width - 1

 $TCU = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU = TCU • CEU

CEOD = TCD • CED

# **Design Entry Method**

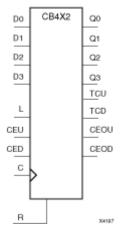
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB4X2

Macro: 4-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading in CPLD architectures.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.



The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Inputs						Outputs				
R	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD	
1	X	Χ	X	-	Χ	0	0	1	0	CEOD	
0	1	Χ	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD	
0	0	0	0	X	Х	No Chg	No Chg	No Chg	0	0	
0	0	1	0	-	Χ	Inc	TCU	TCD	CEOU	0	
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD	
0	0	1	1	-	Х	Inc	TCU	TCD	Invalid	Invalid	

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD = TCD • CED

# **Design Entry Method**

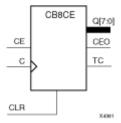
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB8CE

Macro: 8-Bit Cascadable Binary Counter with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs			Outputs			
CLR	CE	С	Qz-Q0	TC	CEO	
1	Χ	Χ	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

$$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0$$

 $CEO = TC \cdot CE$ 

# **Design Entry Method**

This design element is only for use in schematics.

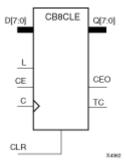


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **CB8CLE**

Macro: 8-Bit Loadable Cascadable Binary Counters with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This element is a synchronously loadable, asynchronously clearable, cascadable binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs			Outputs				
CLR	L	CE	С	Dz – D0	Qz – Q0	TC	CEO
1	Χ	X	X	X	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	X	X	No change	No change	0
0	0	1	-	X	Inc	TC	CEO

z = bit width - 1

 $TC = Qz \cdot Q(z-1) \cdot Q(z-2) \cdot ... \cdot Q0$ 

CEO = TC • CE



# **Design Entry Method**

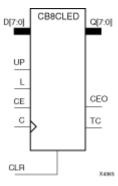
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB8CLED

Macro: 8-Bit Loadable Cascadable Bidirectional Binary Counters with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

For CPLD parts, see "CB2X1", "CB4X1", "CB8X1", "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs						Outputs		
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO
1	Χ	Χ	X	X	Χ	0	0	0
0	1	X	-	X	Dn	Dn	TC	CEO



Inputs						Outputs		
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO
0	0	0	X	X	Χ	No change	No change	0
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO

z = bit width - 1

$$\mathsf{TC} = (\mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0 \bullet \mathsf{UP}) + (\mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0 \bullet \mathsf{UP})$$

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

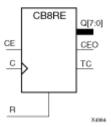
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CB8RE

Macro: 8-Bit Cascadable Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs			Outputs			
R	CE	С	Qz-Q0	TC	CEO	
1	Х	-	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0)$ 

CEO = TC • CE

# **Design Entry Method**

This design element is only for use in schematics.

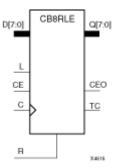


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CB8RLE**

Macro: 8-Bit Loadable Cascadable Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to zero on the Low-to-High clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs			Outputs				
R	L	CE	С	Dz – D0	Qz - Q0	TC	CEO
1	X	X	-	X	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	Х	Х	No change	No change	0



Inputs			Outputs						
R	L	CE	Qz - Q0	TC	CEO				
0	0	1	-	X	Inc	TC	CEO		
z = bit width -	1								
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC•CE	CEO = TC•CE								

# **Design Entry Method**

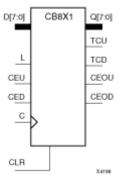
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB8X1

Macro: 8-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, connect the CEOU and CEOD outputs of each counter directly to the CEU and CED inputs, respectively, of the next stage. Connect the clock, L, and CLR inputs in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	X	Х	X	X	Χ	0	0	1	0	CEOD
0	1	Χ	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	X	No Change	No Change	No Change	0	0
0	0	1	0	-	Χ	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	Χ	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD =  $TCD \cdot CED$ 

# **Design Entry Method**

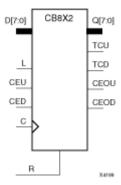
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CB8X2

Macro: 8-Bit Loadable Cascadable Bidirectional Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading in CPLD architectures.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

The counter is initialized to zero (TCU Low and TCD High) when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
R	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	X	Χ	Χ	-	X	0	0	1	0	CEOD
0	1	Χ	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Χ	Χ	No Chg	No Chg	No Chg	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD = TCD $\bullet$ CED

# **Design Entry Method**

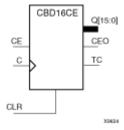
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD16CE

Macro: 16-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is an asynchronously clearable, cascadable dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs			Outputs				
CLR	CE	С	Qz – Q0	TC	CEO		
1	Χ	Χ	0	0	0		
0	0	Χ	No change	No change	0		
0	1	-	Inc	TC	CEO		
0	1	-	Inc	TC	CEO		

z = bit width - 1

$$\mathsf{TC} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$$

CEO = TC•CE

## **Design Entry Method**

This design element is only for use in schematics.

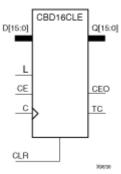


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD16CLE

Macro: 16-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This is a synchronously loadable, asynchronously clearable, cascadable dual edge triggered binary counters. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### **Logic Table**

Inputs			Outputs	Outputs			
CLR	L	CE	С	Dz – D0	Qz – Q0	TC	CEO
1	X	Х	Х	X	0	0	0
0	1	Х	-	Dn	Dn	TC	CEO
0	1	Х	-	Dn	Dn	TC	CEO
0	0	0	Х	X	No change	No change	0
0	0	1	-	X	Inc	TC	CEO



Inputs			Outputs					
CLR	R L CE C Dz – D0					TC	CEO	
0	0	1	-	X	Inc	TC	CEO	
z = bit width	- 1							
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$								
CEO = TC•CE								

# **Design Entry Method**

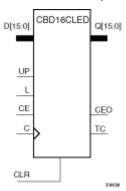
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CBD16CLED

Macro: 16-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



#### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High and High-to-Low clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

See "CB2X1," "CB4X1," "CB8X1," "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs			Outputs					
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO
1	Χ	Χ	Χ	Χ	Χ	0	0	0
0	1	X	-	X	Dn	Dn	TC	CEO
0	1	X	=	X	Dn	Dn	TC	CEO
0	0	0	Х	Х	Х	No change	No change	0



Inputs			Outputs					
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO

z = bit width - 1

 $\mathsf{TC} = (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet UP) + (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet UP)$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

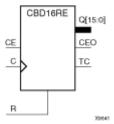
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD16RE

Macro: 16-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero during the Low-to-High or High-to-Low clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Inputs			Outputs				
R	CE	С	Qz-Q0	TC	CEO		
1	X	-	0	0	0		
1	Х	-	0	0	0		
0	0	X	No change	No change	0		
0	1	-	Inc	TC	CEO		
0	1	-	Inc	TC	CEO		

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0)$ 

CEO = TC•CE

## **Design Entry Method**

This design element is only for use in schematics.

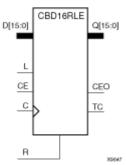


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD16RLE

Macro: 16-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) outputs to Low on the Low-to-High or High-to-Low clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High and High-to-Low clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs					Outputs			
R	L	CE	С	Dz - D0	Qz – Q0	TC	CEO	
1	X	X	-	X	0	0	0	
1	X	X	-	X	0	0	0	
0	1	X	-	Dn	Dn	TC	CEO	
0	1	X	-	Dn	Dn	TC	CEO	
0	0	0	X	Χ	No change	No change	0	
0	0	1	-	X	Inc	TC	CEO	



Inputs			Outputs					
R L CE C Dz – D0					Qz – Q0	TC	CEO	
0	0	1	-	X	Inc	TC	CEO	
z = bit width - 1								
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$								

# **Design Entry Method**

CEO = TC∙CE

This design element is only for use in schematics.

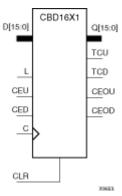
# **Available Attributes**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD16X1

Macro: 16-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional dual edge triggered binary counters. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High and High-to-Low clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are connected directly to the CEU and CED inputs, respectively, of the next stage. The clock, L, and CLR inputs are connected in parallel.

The maximum clocking frequency of these counters is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	X	Χ	Χ	Χ	Χ	0	0	1	0	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	-	Х	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	-	Х	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $\mathsf{TCU} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD =  $TCD \bullet CED$ 

# **Design Entry Method**

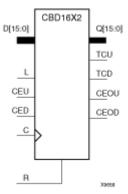
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD16X2

Macro: 16-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional dual edge triggered binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High and High-to-Low clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High and High-to-Low clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
R	L	CEU	CED	С	Dz – D0	Qz – Q0	TCU	TCD	CEOU	CEOD
1	X	X	X	-	X	0	0	1	0	CEOD
1	X	X	X	-	X	0	0	1	0	CEOD
0	1	Х	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	_	X	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	_	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU = TCU•CEU

CEOD = TCD • CED

# **Design Entry Method**

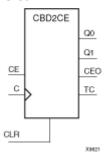
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD2CE

Macro: 2-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is an asynchronously clearable, cascadable dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs			
CLR	CE	С	Qz – Q0	TC	CEO	
1	X	X	0	0	0	
0	0	Х	No change	No change	0	
0	1	-	Inc	TC	CEO	
0	1	-	Inc	TC	CEO	

z = bit width - 1

$$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0$$

CEO = TC•CE

#### **Design Entry Method**

This design element is only for use in schematics.

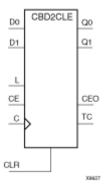


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD2CLE

Macro: 2-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This is a synchronously loadable, asynchronously clearable, cascadable dual edge triggered binary counters. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### **Logic Table**

Inputs			Outputs				
CLR	L	CE	С	Dz - D0	Qz - Q0	TC	CEO
1	Χ	Χ	Χ	Χ	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	X	Χ	No change	No change	0
0	0	1	-	Χ	Inc	TC	CEO



Inputs			Outputs					
CLR	L	CE	С	Dz – D0	Qz – Q0	TC	CEO	
0	0	1	-	Χ	Inc	TC	CEO	
z = bit widt	h - 1							
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$								
CEO = TC•	CE							

# **Design Entry Method**

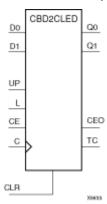
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD2CLED

Macro: 2-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High and High-to-Low clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

See "CB2X1," "CB4X1," "CB8X1," "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs					
CLR	L	CE	С	UP	Dz – D0	Qz - Q0	TC	CEO
1	Χ	Χ	X	Χ	Χ	0	0	0
0	1	X	-	X	Dn	Dn	TC	CEO
0	1	X	-	X	Dn	Dn	TC	CEO



Inputs			Outputs					
CLR	L	CE	С	UP	Dz – D0	Qz - Q0	TC	CEO
0	0	0	Χ	X	X	No change	No change	0
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO

z = bit width - 1

 $\mathsf{TC} = (Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0 \bullet UP) + (Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0 \bullet UP)$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

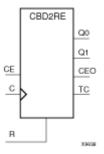
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD2RE

# Macro: 2-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero during the Low-to-High or High-to-Low clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Inputs			Outputs				
R	CE	С	Qz-Q0	TC	CEO		
1	X	-	0	0	0		
1	Х	-	0	0	0		
0	0	X	No change	No change	0		
0	1	-	Inc	TC	CEO		
0	1	-	Inc	TC	CEO		

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0)$ 

CEO = TC•CE



# **Design Entry Method**

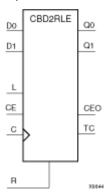
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD2RLE

Macro: 2-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



#### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) outputs to Low on the Low-to-High or High-to-Low clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High and High-to-Low clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs				
R	L	CE	С	Dz - D0	Qz - Q0	TC	CEO
1	X	X	-	X	0	0	0
1	Х	X	-	Х	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	X	Х	No change	No change	0



Inputs			Outputs				
R	L	CE	С	Dz – D0	Qz - Q0	TC	CEO
0	0	1	-	X	Inc	TC	CEO
0	0	1	-	X	Inc	TC	CEO

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

CEO =  $TC \bullet CE$ 

# **Design Entry Method**

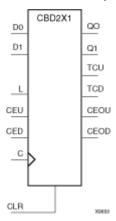
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD2X1

Macro: 2-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional dual edge triggered binary counters. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High and High-to-Low clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are connected directly to the CEU and CED inputs, respectively, of the next stage. The clock, L, and CLR inputs are connected in parallel.

The maximum clocking frequency of these counters is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	X	Χ	Χ	Χ	Χ	0	0	1	0	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	-	Х	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	-	Х	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $\mathsf{TCU} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD =  $TCD \bullet CED$ 

# **Design Entry Method**

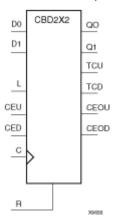
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD2X2

Macro: 2-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional dual edge triggered binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High and High-to-Low clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High and High-to-Low clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
R	L	CEU	CED	С	Dz – D0	Qz – Q0	TCU	TCD	CEOU	CEOD
1	X	X	X	-	X	0	0	1	0	CEOD
1	X	X	X	-	X	0	0	1	0	CEOD
0	1	Х	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	_	X	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	_	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU = TCU•CEU

CEOD = TCD • CED

# **Design Entry Method**

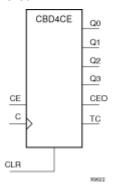
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD4CE

Macro: 4-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is an asynchronously clearable, cascadable dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs			
CLR	CE	С	Qz – Q0	TC	CEO	
1	Х	Χ	0	0	0	
0	0	X	No change	No change	0	
0	1	-	Inc	TC	CEO	
0	1	-	Inc	TC	CEO	

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0$ 

CEO = TC • CE



# **Design Entry Method**

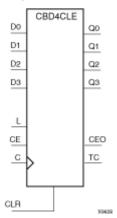
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### CBD4CLE

Macro: 4-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only: CoolRunner-II

#### Introduction

This is a synchronously loadable, asynchronously clearable, cascadable dual edge triggered binary counters. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Inputs					Outputs	Outputs		
CLR	L	CE	С	Dz – D0	Qz – Q0	TC	CEO	
1	Х	X	X	X	0	0	0	
0	1	X	-	Dn	Dn	TC	CEO	
0	1	Х	-	Dn	Dn	TC	CEO	
0	0	0	Х	X	No change	No change	0	
0	0	1	-	X	Inc	TC	CEO	



Inputs						Outputs			
CLR	LR L CE C Dz – D0					TC	CEO		
0	0	Inc	TC	CEO					
z = bit width -	- 1								
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC∙CE									

# **Design Entry Method**

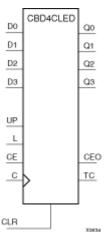
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD4CLED

Macro: 4-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High and High-to-Low clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

See "CB2X1," "CB4X1," "CB8X1," "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Inputs						Outputs			
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO		
1	Χ	Χ	Χ	Χ	X	0	0	0		
0	1	X	-	X	Dn	Dn	TC	CEO		



Inputs						Outputs			
CLR	L	CE	С	UP	Dz – D0	Qz - Q0	TC	CEO	
0	1	X	-	X	Dn	Dn	TC	CEO	
0	0	0	X	X	Χ	No change	No change	0	
0	0	1	-	1	X	Inc	TC	CEO	
0	0	1	-	1	X	Inc	TC	CEO	
0	0	1	-	0	X	Dec	TC	CEO	
0	0	1	-	0	Х	Dec	TC	CEO	

z = bit width - 1

 $\mathsf{TC} = (Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0 \bullet UP) + (Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0 \bullet UP)$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

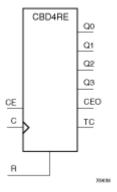
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD4RE

Macro: 4-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero during the Low-to-High or High-to-Low clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs				
R	CE	С	Qz-Q0	TC	CEO		
1	X	-	0	0	0		
1	Х	-	0	0	0		
0	0	Х	No change	No change	0		
0	1	-	Inc	TC	CEO		
0	1	-	Inc	TC	CEO		

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0)$ 

CEO = TC•CE



# **Design Entry Method**

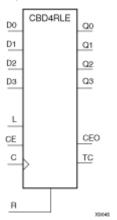
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **CBD4RLE**

Macro: 4-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) outputs to Low on the Low-to-High or High-to-Low clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High and High-to-Low clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Inputs					Outputs			
R	L	CE	С	Dz - D0	Qz - Q0	TC	CEO		
1	X	X	-	X	0	0	0		
1	X	X	-	X	0	0	0		
0	1	X	-	Dn	Dn	TC	CEO		
0	1	X	-	Dn	Dn	TC	CEO		



Inputs			Outputs				
R	L	CE	С	Dz - D0	Qz - Q0	TC	CEO
0	0	0	Χ	Χ	No change	No change	0
0	0	1	-	X	Inc	TC	CEO
0	0	1	-	Х	Inc	TC	CEO

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

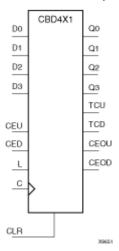
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD4X1

Macro: 4-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional dual edge triggered binary counters. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High and High-to-Low clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are connected directly to the CEU and CED inputs, respectively, of the next stage. The clock, L, and CLR inputs are connected in parallel.

The maximum clocking frequency of these counters is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.



This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	Χ	X	Χ	X	Χ	0	0	1	0	CEOD
0	1	Х	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $\mathsf{TCU} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

 $TCD = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$ 

CEOU = TCU•CEU

CEOD = TCD • CED

# **Design Entry Method**

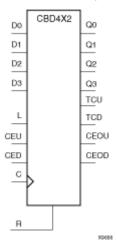
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD4X2

Macro: 4-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional dual edge triggered binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High and High-to-Low clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High and High-to-Low clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.



This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs						Outputs				
R	L	CEU	CED	С	Dz – D0	Qz – Q0	TCU	TCD	CEOU	CEOD
1	Χ	X	X	-	X	0	0	1	0	CEOD
1	Х	Х	Х	-	Х	0	0	1	0	CEOD
0	1	Χ	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	-	Х	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	X	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	-	Х	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD = TCD • CED

# **Design Entry Method**

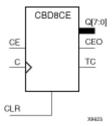
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CBD8CE

Macro: 8-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is an asynchronously clearable, cascadable dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs			Outputs				
CLR	CE	С	Qz – Q0	TC	CEO		
1	Χ	Χ	0	0	0		
0	0	Χ	No change	No change	0		
0	1	-	Inc	TC	CEO		
0	1	-	Inc	TC	CEO		

z = bit width - 1

$$TC = Qz \cdot Q(z-1) \cdot Q(z-2) \cdot ... \cdot Q0$$

CEO = TC•CE

### **Design Entry Method**

This design element is only for use in schematics.

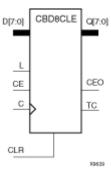


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD8CLE

Macro: 8-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This is a synchronously loadable, asynchronously clearable, cascadable dual edge triggered binary counters. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock transition, independent of the state of clock enable (CE). The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Inputs					Outputs			
CLR L CE C			С	Dz - D0	Qz - Q0	TC	CEO		
1	Χ	Χ	Χ	Χ	0	0	0		
0	1	X	-	Dn	Dn	TC	CEO		
0	1	X	-	Dn	Dn	TC	CEO		
0	0	0	Х	Χ	No change	No change	0		
0	0	1	-	X	Inc	TC	CEO		



Inputs						Outputs			
CLR	.R L CE C Dz – D0					TC	CEO		
0	0	1	Χ	Inc	TC	CEO			
z = bit width	ı - 1								
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC•CE									

# **Design Entry Method**

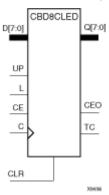
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CBD8CLED

Macro: 8-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, cascadable, bidirectional dual edge triggered binary counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition, independent of the state of clock enable (CE). The Q outputs decrement when CE is High and UP is Low during the Low-to-High and High-to-Low clock transition. The Q outputs increment when CE and UP are High. The counter ignores clock transitions when CE is Low.

For counting up, the TC output is High when all Q outputs and UP are High. For counting down, the TC output is High when all Q outputs and UP are Low.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, UP, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

See "CB2X1," "CB4X1," "CB8X1," "CB16X1" for high-performance cascadable, bidirectional counters.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs	Inputs						Outputs			
CLR	L	CE	С	UP	Dz - D0	Qz - Q0	TC	CEO		
1	X	Χ	Χ	X	Χ	0	0	0		
0	1	Χ	-	X	Dn	Dn	TC	CEO		
0	1	X	-	X	Dn	Dn	TC	CEO		
0	0	0	Χ	Х	Х	No change	No change	0		



Inputs			Outputs					
CLR	L	CE	С	UP	Dz – D0	Qz - Q0	TC	CEO
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	1	X	Inc	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO
0	0	1	-	0	X	Dec	TC	CEO

z = bit width - 1

 $\mathsf{TC} = (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet UP) + (Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0 \bullet UP)$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

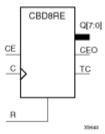
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD8RE

Macro: 8-Bit Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) to logic level zero during the Low-to-High or High-to-Low clock transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when both Q outputs are High.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs			Outputs				
R	CE	С	Qz-Q0	TC	CEO		
1	X	-	0	0	0		
1	Х	-	0	0	0		
0	0	Χ	No change	No change	0		
0	1	-	Inc	TC	CEO		
0	1	-	Inc	TC	CEO		

z = bit width - 1

 $TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet ... \bullet Q0)$ 

CEO = TC•CE

## **Design Entry Method**

This design element is only for use in schematics.

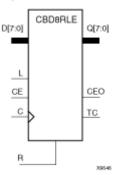


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD8RLE

Macro: 8-Bit Loadable Cascadable Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, cascadable dual edge triggered binary counter. The synchronous reset (R), when High, overrides all other inputs and forces the Q outputs, terminal count (TC), and clock enable out (CEO) outputs to Low on the Low-to-High or High-to-Low clock (C) transition.

The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High and High-to-Low clock (C) transition, independent of the state of CE. The Q outputs increment when CE is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when all Q outputs are High. The CEO output is High when all Q outputs and CE are High to allow direct cascading of counters.

Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs					Outputs		
R	L	CE	С	Dz – D0	Qz – Q0	TC	CEO
1	X	X	-	X	0	0	0
1	X	X	-	X	0	0	0
0	1	X	-	Dn	Dn	TC	CEO
0	1	X	-	Dn	Dn	TC	CEO
0	0	0	X	X	No change	No change	0
0	0	1	-	X	Inc	TC	CEO



Inputs					Outputs	Outputs			
R	L CE C Dz – D0					TC	CEO		
0	0	1	Inc	TC	CEO				
z = bit wid	th - 1		•	•			•		
$TC = Qz \bullet Q(z-1) \bullet Q(z-2) \bullet \dots \bullet Q0$									
CEO = TC∙CE									

# **Design Entry Method**

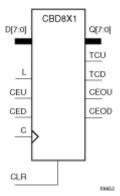
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD8X1

Macro: 8-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronously loadable, asynchronously clearable, bidirectional dual edge triggered binary counters. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high speed cascading.

The asynchronous clear (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, clock enable outputs CEOU and CEOD go to Low and High, respectively, independent of clock transitions. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

The Q outputs increment when CEU is High, provided CLR and L are Low, during the Low-to-High and High-to-Low clock transition. The Q outputs decrement when CED is High, provided CLR and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are connected directly to the CEU and CED inputs, respectively, of the next stage. The clock, L, and CLR inputs are connected in parallel.

The maximum clocking frequency of these counters is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
CLR	L	CEU	CED	С	Dz-D0	Qz-Q0	TCU	TCD	CEOU	CEOD
1	X	Χ	Χ	Χ	Χ	0	0	1	0	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	X	X	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	-	Х	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $\mathsf{TCU} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU =  $TCU \cdot CEU$ 

CEOD =  $TCD \bullet CED$ 

# **Design Entry Method**

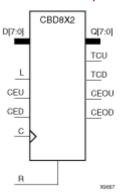
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CBD8X2

Macro: 8-Bit Loadable Cascadable Bidirectional Dual Edge Triggered Binary Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a synchronous, loadable, resettable, bidirectional dual edge triggered binary counter. It has separate count-enable inputs and synchronous terminal-count outputs for up and down directions to support high-speed cascading.

The synchronous reset (R) is the highest priority input. When R is High, all other inputs are ignored; the data outputs (Q) go to logic level zero, terminal count outputs TCU and TCD go to zero and one, respectively, and clock enable outputs CEOU and CEOD go to Low and High, respectively, on the Low-to-High and High-to-Low clock (C) transition. The data on the D inputs loads into the counter on the Low-to-High and High-to-Low clock (C) transition when the load enable input (L) is High, independent of the CE inputs.

All Q outputs increment when CEU is High, provided R and L are Low during the Low-to-High and High-to-Low clock transition. All Q outputs decrement when CED is High, provided R and L are Low. The counter ignores clock transitions when CEU and CED are Low. Both CEU and CED should not be High during the same clock transition; the CEOU and CEOD outputs might not function properly for cascading when CEU and CED are both High.

For counting up, the CEOU output is High when all Q outputs and CEU are High. For counting down, the CEOD output is High when all Q outputs are Low and CED is High. To cascade counters, the CEOU and CEOD outputs of each counter are, respectively, connected directly to the CEU and CED inputs of the next stage. The C, L, and R inputs are connected in parallel.

The maximum clocking frequency of these counter components is unaffected by the number of cascaded stages for all counting and loading functions. The TCU terminal count output is High when all Q outputs are High, regardless of CEU. The TCD output is High when all Q outputs are Low, regardless of CED.

When cascading counters, the final terminal count signals can be produced by AND wiring all the TCU outputs (for the up direction) and all the TCD outputs (for the down direction). The TCU, CEOU, and CEOD outputs are produced by optimizable AND gates within the component. This results in zero propagation from the CEU and CED inputs and from the Q outputs, provided all connections from each such output remain on-chip. Otherwise, a macrocell buffer delay is introduced.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs						Outputs				
R	L	CEU	CED	С	Dz – D0	Qz – Q0	TCU	TCD	CEOU	CEOD
1	Χ	X	X	-	X	0	0	1	0	CEOD
1	Χ	X	Χ	-	X	0	0	1	0	CEOD
0	1	Х	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	1	Х	Х	-	Dn	Dn	TCU	TCD	CEOU	CEOD
0	0	0	0	Х	Х	No Change	No Change	No Change	0	0
0	0	1	0	-	X	Inc	TCU	TCD	CEOU	0
0	0	1	0	-	Х	Inc	TCU	TCD	CEOU	0
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	0	1	-	Х	Dec	TCU	TCD	0	CEOD
0	0	1	1	-	Х	Inc	TCU	TCD	Invalid	Invalid
0	0	1	1	-	X	Inc	TCU	TCD	Invalid	Invalid

z = bit width - 1

 $TCU = Qz \bullet Q(z\text{-}1) \bullet Q(z\text{-}2) \bullet \dots \bullet Q0$ 

 $\mathsf{TCD} = \mathsf{Qz} \bullet \mathsf{Q}(\mathsf{z}\text{-}1) \bullet \mathsf{Q}(\mathsf{z}\text{-}2) \bullet \dots \bullet \mathsf{Q}0$ 

CEOU = TCU•CEU

CEOD = TCD • CED

# **Design Entry Method**

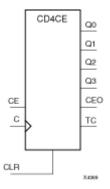
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CD4CE

Macro: 4-Bit Cascadable BCD Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

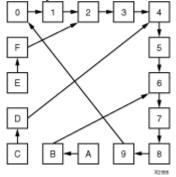
This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

CD4CE is a 4-bit (stage), asynchronous clearable, cascadable binary-coded-decimal (BCD) counter. The asynchronous clear input (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; the Q outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero, independent of clock transitions. The Q outputs increment when clock enable (CE) is High during the Low-to-High clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when Q3 and Q0 are High and Q2 and Q1 are Low.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs			Outputs							
CLR	CE	С	Q3	Q2	Q1	Q0	TC	CEO		
1	X	X	0	0	0	0	0	0		
0	1	-	Inc	Inc	Inc	Inc	TC	CEO		
0	0	X	No Change	No Change	No Change	No Change	TC	0		
0	1	Χ	1	0	0	1	1	1		

 $\mathsf{TC} = \mathsf{Q3} \bullet ! \mathsf{Q2} \bullet ! \mathsf{Q1} \bullet \mathsf{Q0}$ 

CEO = TC•CE

# **Design Entry Method**

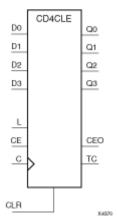
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CD4CLE

Macro: 4-Bit Loadable Cascadable BCD Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

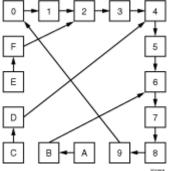
This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

CD4CLE is a 4-bit (stage), synchronously loadable, asynchronously clearable, binarycoded- decimal (BCD) counter. The asynchronous clear input (CLR) is the highest priority input. When (CLR) is High, all other inputs are ignored; the (Q) outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero, independent of clock transitions. The data on the (D) inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition. The (Q) outputs increment when clock enable input (CE) is High during the Low- to-High clock transition. The counter ignores clock transitions when (CE) is Low. The (TC) output is High when Q3 and Q0 are High and Q2 and Q1 are Low.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.



This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs					Outputs	Outputs						
CLR	L	CE	D3 – D0	С	Q3	Q2	Q1	Q0	TC	CEO		
1	Х	Х	Х	Х	0	0	0	0	0	0		
0	1	Х	D3 – D0	-	D3	D2	D1	D0	TC	CEO		
0	0	1	Х	-	Inc	Inc	Inc	Inc	TC	CEO		
0	0	0	X	Х	No Change	No Change	No Change	No Change	TC	0		
0	0	1	Х	Х	1	0	0	1	1	1		

 $TC = Q3 \bullet !Q2 \bullet !Q1 \bullet Q0$ 

CEO = TC•CE

# **Design Entry Method**

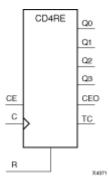
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CD4RE

Macro: 4-Bit Cascadable BCD Counter with Clock Enable and Synchronous Reset



# **Supported Architectures**

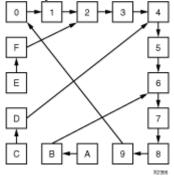
This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

CD4RE is a 4-bit (stage), synchronous resettable, cascadable binary-coded-decimal (BCD) counter. The synchronous reset input (R) is the highest priority input. When (R) is High, all other inputs are ignored; the (Q) outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero on the Low-to-High clock (C) transition. The (Q) outputs increment when the clock enable input (CE) is High during the Low-to-High clock transition. The counter ignores clock transitions when (CE) is Low. The (TC) output is High when Q3 and Q0 are High and Q2 and Q1 are Low.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs			Outputs							
R	CE	С	Q3	Q2	Q1	Q0	TC	CEO		
1	X	-	0	0	0	0	0	0		
0	1	-	Inc	Inc	Inc	Inc	TC	CEO		
0	0	X	No Change	No Change	No Change	No Change	TC	0		
0	1	Χ	1	0	0	1	1	1		

 $TC = Q3 \bullet ! Q2 \bullet ! Q1 \bullet Q0$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

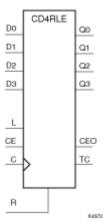
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CD4RLE

Macro: 4-Bit Loadable Cascadable BCD Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

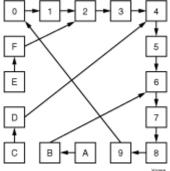
This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

CD4RLE is a 4-bit (stage), synchronous loadable, resettable, binary-coded-decimal (BCD) counter. The synchronous reset input (R) is the highest priority input. When R is High, all other inputs are ignored; the Q outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero on the Low-to-High clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High clock (C) transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when Q3 and Q0 are High and Q2 and Q1 are Low.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.



This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Input	s				Outputs					
R	L	CE	D3 – D0	С	Q3	Q2	Q1	Q0	TC	CEO
1	Х	X	X	-	0	0	0	0	0	0
0	1	Χ	D3 – D0	-	D3	D	D	D0	TC	CEO
0	0	1	X	-	Inc	Inc	Inc	Inc	TC	CEO
0	0	0	Х	Х	No Change	No Change	No Change	No Change	TC	0
0	0	1	X	X	1	0	0	1	1	1

 $\mathsf{TC} = \mathsf{Q3} \bullet ! \mathsf{Q2} \bullet ! \mathsf{Q1} \bullet \mathsf{Q0}$ 

CEO = TC•CE

# **Design Entry Method**

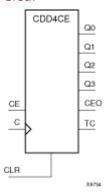
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CDD4CE

Macro: 4-Bit Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

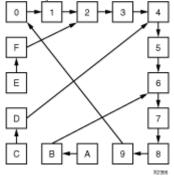
This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

CDD4CE is a 4-bit (stage), asynchronous clearable, cascadable dual edge triggered Binary-coded-decimal (BCD) counter. The asynchronous clear input (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; the Q outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero, independent of clock transitions. The Q outputs increment when clock enable (CE) is High during the Low-to-High and High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low. The TC output is High when Q3 and Q0 are High and Q2 and Q1 are Low. The counter recovers to zero from any illegal state within the first clock cycle.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs			Outputs					
CLR	CE	С	Q3	Q2	Q1	Q0	TC	CEO
1	Х	X	0	0	0	0	0	0
0	1	-	Inc	Inc	Inc	Inc	TC	CEO
0	1	-	Inc	Inc	Inc	Inc	TC	CEO
0	0	Х	No Change	No Change	No Change	No Change	TC	0
0	1	Х	1	0	0	1	1	1
TC = Q3•	!Q2•!Q1•Q0	)						

# **Design Entry Method**

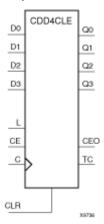
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CDD4CLE

Macro: 4-Bit Loadable Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

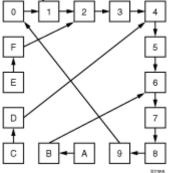
This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

CDD4CLE is a 4-bit (stage), synchronously loadable, asynchronously clearable, dual edge triggered Binary-coded-decimal (BCD) counter. The asynchronous clear input (CLR) is the highest priority input. When CLR is High, all other inputs are ignored; the Q outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero, independent of clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High and High-to-Low clock (C) transitions. The Q outputs increment when clock enable input (CE) is High during the Low- to-High clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when Q3 and Q0 are High and Q2 and Q1 are Low. The counter recovers to zero from any illegal state within the first clock cycle.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C, L, and CLR inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.



This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs							
CLR	L	CE	D3 – D0	С	Q3	Q2	Q1	Q0	TC	CEO
1	X	Х	Х	Х	0	0	0	0	0	0
0	1	X	D3 – D0	-	D3	D2	D1	D0	TC	CEO
0	1	Х	D3 – D0	-	D3	D2	D1	D0	TC	CEO
0	0	1	Х	-	Inc	Inc	Inc	Inc	TC	CEO
0	0	1	Х	-	Inc	Inc	Inc	Inc	TC	CEO
0	0	0	Х	Х	No Change	No Change	No Change	No Change	TC	0
0	0	1	Х	Х	1	0	0	1	1	1

 $\mathsf{TC} = \mathsf{Q3} \bullet ! \mathsf{Q2} \bullet ! \mathsf{Q1} \bullet \mathsf{Q0}$ 

CEO = TC•CE

## **Design Entry Method**

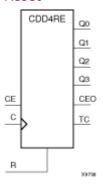
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CDD4RE

# Macro: 4-Bit Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

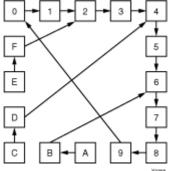
This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

CDD4RE is a 4-bit (stage), synchronous resettable, cascadable dual edge triggered binary-coded-decimal (BCD) counter. The synchronous reset input (R) is the highest priority input. When R is High, all other inputs are ignored; the Q outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero on the Low-to-High or High-to-Low clock (C) transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when Q3 and Q0 are High and Q2 and Q1 are Low. The counter recovers to zero from any illegal state within the first clock cycle.

The counter recovers from any of six possible illegal states and returns to a normal count sequence within two clock cycles for Xilinx® devices, as shown in the following state diagram:



Create larger counters by connecting the CEO output of each stage to the CE input of the next stage and connecting the C and R inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n ( $t_{\text{CE-TC}}$ ), where n is the number of stages and the time  $t_{\text{CE-TC}}$  is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input or use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.



# **Logic Table**

Inputs			Outputs	Outputs					
R	CE	С	Q3	Q2	Q1	Q0	TC	CEO	
1	X	-	0	0	0	0	0	0	
1	Х	-	0	0	0	0	0	0	
0	1	-	Inc	Inc	Inc	Inc	TC	CEO	
0	1	-	Inc	Inc	Inc	Inc	TC	CEO	
0	0	Х	No Change	No Change	No Change	No Change	TC	0	
0	1	Χ	1	0	0	1	1	1	

 $TC = Q3 \bullet !Q2 \bullet !Q1 \bullet Q0$ 

 $CEO = TC \bullet CE$ 

# **Design Entry Method**

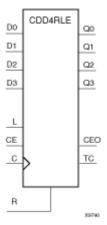
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CDD4RLE

Macro: 4-Bit Loadable Cascadable Dual Edge Triggered BCD Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This is a 4-bit (stage), synchronous loadable, resettable, dual edge triggered binary-coded-decimal (BCD) counter. The synchronous reset input (R) is the highest priority input. When R is High, all other inputs are ignored; the Q outputs, terminal count (TC), and clock enable out (CEO) go to logic level zero on the Low-to-High or High-to-Low clock transitions. The data on the D inputs is loaded into the counter when the load enable input (L) is High during the Low-to-High and High-to-Low clock (C) transition. The Q outputs increment when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. The counter ignores clock transitions when CE is Low. The TC output is High when Q3 and Q0 are High and Q2 and Q1 are Low. The counter recovers to zero from any illegal state within the first clock cycle.

Larger counters are created by connecting the count enable out (CEO) output of the first stage to the CE input of the next stage and connecting the R, L, and C inputs in parallel. CEO is active (High) when TC and CE are High. The maximum length of the counter is determined by the accumulated CE-to-TC propagation delays versus the clock period. The clock period must be greater than n(tCE-TC), where n is the number of stages and the time tCE-TC is the CE-to-TC propagation delay of each stage. When cascading counters, use the CEO output if the counter uses the CE input; use the TC output if it does not.

This counter is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Design Entry Method**

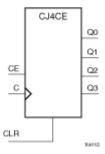
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CJ4CE

Macro: 4-Bit Johnson Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a clearable Johnson/shift counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the data (Q) outputs to logic level zero, independent of clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2, and so forth) when the clock enable input (CE) is High during the Low-to-High clock transition. Clock transitions are ignored when (CE) is Low.

The Q3 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .

### **Logic Table**

Inputs			Outputs		
CLR	CE	С	Q0	Q1 through Q3	
1	Χ	X	0	0	
0	0	X	No change	No change	
0	1	-	!q3	q0 through q2	

q = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

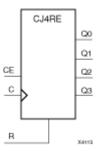
#### **Available Attributes**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CJ4RE

#### Macro: 4-Bit Johnson Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a resettable Johnson/shift counter. The synchronous reset (R) input, when High, overrides all other inputs and forces the data (Q) outputs to logic level zero during the Low-to-High clock (C) transition. The counter increments (shifts Q0 to Q1, Q1 to Q2, and so forth) when the clock enable input (CE) is High during the Low-to-High clock transition. Clock transitions are ignored when CE is Low.

The Q3 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .

#### Logic Table

Inputs			Outputs		
R	CE	С	Q0	Q1 through Q3	
1	X	-	0	0	
0	0	Х	No change	No change	
0	1	-	!q3	q0 through q2	

q = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

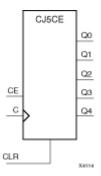
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CJ5CE

#### Macro: 5-Bit Johnson Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a clearable Johnson/shift counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the data (Q) outputs to logic level zero, independent of clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2, and so forth) when the clock enable input (CE) is High during the Low-to-High clock transition. Clock transitions are ignored when (CE) is Low.

The Q4 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .

## **Logic Table**

Inputs			Outputs	
CLR	CE	С	Q0	Q1 through Q4
1	Χ	X	0	0
0	0	Х	No change	No change
0	1	-	!q4	q0 through q3

 $\mathbf{q}$  = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

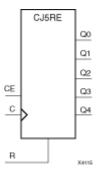
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **CJ5RE**

Macro: 5-Bit Johnson Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a resettable Johnson/shift counter. The synchronous reset (R) input, when High, overrides all other inputs and forces the data (Q) outputs to logic level zero during the Low-to-High clock (C) transition. The counter increments (shifts Q0 to Q1, Q1 to Q2, and so forth) when the clock enable input (CE) is High during the Low-to-High clock transition. Clock transitions are ignored when CE is Low.

The Q4 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .

## **Logic Table**

Inputs			Outputs	
R	CE	С	Q0	Q1 through Q4
1	X	-	0	0
0	0	X	No change	No change
0	1	-	!q4	q0 through q3

q = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

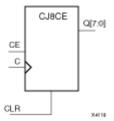
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CJ8CE

#### Macro: 8-Bit Johnson Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a clearable Johnson/shift counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and forces the data (Q) outputs to logic level zero, independent of clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2, and so forth) when the clock enable input (CE) is High during the Low-to-High clock transition. Clock transitions are ignored when (CE) is Low.

The Q7 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .

## **Logic Table**

Inputs			Outputs		
CLR	CE	С	Q0	Q1 through Q8	
1	Χ	X	0	0	
0	0	X	No change	No change	
0	1	-	!q7	q0 through q7	

q = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

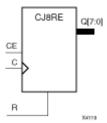
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CJ8RE

Macro: 8-Bit Johnson Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a resettable Johnson/shift counter. The synchronous reset (R) input, when High, overrides all other inputs and forces the data (Q) outputs to logic level zero during the Low-to-High clock (C) transition. The counter increments (shifts Q0 to Q1, Q1 to Q2, and so forth) when the clock enable input (CE) is High during the Low-to-High clock transition. Clock transitions are ignored when CE is Low.

The Q7 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .

## Logic Table

Inputs			Outputs		
R	CE	С	Q0	Q1 through Q7	
1	X	-	0	0	
0	0	X	No change	No change	
0	1	-	!q7	q0 through q6	

 ${\bf q}$  = state of referenced output one setup time prior to active clock transition

### **Design Entry Method**

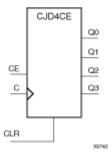
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CJD4CE

Macro: 4-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is a dual edge triggered clearable Johnson/shift counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and causes the data (Q) outputs to go to logic level zero independent of clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2,etc.) when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. Clock transitions are ignored when CE is Low.

The Q3 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs		Outputs	Outputs			
CLR	CE	С	Q0	Q1 through Q3		
1	Χ	X	0	0		
0	0	X	No Change	No Change		
0	1	-	!q3	q0 through q2		
0	1	-	!q3	q0 through q2		
q = state of referenced output one setup time prior to active clock transition						

## **Design Entry Method**

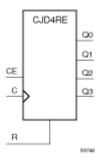
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CJD4RE

Macro: 4-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a resettable dual edge triggered Johnson/shift counter. The synchronous reset (R) input, when High, overrides all other inputs and causes the data (Q) outputs to go to logic level zero during the Low-to-High and High-to-Low clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2, etc.) when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. Clock transitions are ignored when CE is Low.

The Q3 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs			Outputs	Outputs		
R	CE	С	Q0	Q1 – Q3		
1	X	-	0	0		
1	X	-	0	0		
0	0	X	No Change	No Change		
0	1	-	!q3	q0 – q2		
0	1	-	!q3	q0 – q2		
q = state of referenced output one setup time prior to active clock transition						

## **Design Entry Method**

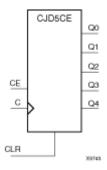
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CJD5CE

Macro: 5-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is a dual edge triggered clearable Johnson/shift counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and causes the data (Q) outputs to go to logic level zero independent of clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2,etc.) when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. Clock transitions are ignored when CE is Low.

The Q4 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs	Outputs		
CLR	CE	С	Q0	Q1 through Q4		
1	X	X	0	0		
0	0	X	No Change	No Change		
0	1	-	!q4	q0 through q3		
0	1	-	!q4	q0 through q3		
q = state of referenced output one setup time prior to active clock transition						

## **Design Entry Method**

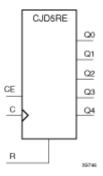
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CJD5RE**

Macro: 5-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a resettable dual edge triggered Johnson/shift counter. The synchronous reset (R) input, when High, overrides all other inputs and causes the data (Q) outputs to go to logic level zero during the Low-to-High and High-to-Low clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2, etc.) when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. Clock transitions are ignored when CE is Low.

The Q4 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs			Outputs	Outputs		
R	CE	С	Q0	Q1 – Q4		
1	X	-	0	0		
1	X	-	0	0		
0	0	X	No Change	No Change		
0	1	-	!q4	q0 - q3		
0	1	-	!q4	q0 - q3		
q = state of ref	ferenced output one setu	o time prior to active clo	ock transition	•		

## **Design Entry Method**

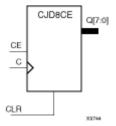
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CJD8CE

Macro: 8-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This element is a dual edge triggered clearable Johnson/shift counter. The asynchronous clear (CLR) input, when High, overrides all other inputs and causes the data (Q) outputs to go to logic level zero independent of clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2,etc.) when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. Clock transitions are ignored when CE is Low.

The Q7 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs			Outputs	
CLR	CE	С	Q0	Q1 through Q7
1	X	X	0	0
0	0	X	No Change	No Change
0	1	-	!q7	q0 through q6
0	1	-	!q7	q0 through q6
g = state of refer	renced output one setu	time prior to active clo	ck transition	•

## **Design Entry Method**

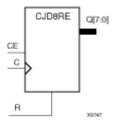
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CJD8RE

Macro: 8-Bit Dual Edge Triggered Johnson Counter with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a resettable dual edge triggered Johnson/shift counter. The synchronous reset (R) input, when High, overrides all other inputs and causes the data (Q) outputs to go to logic level zero during the Low-to-High and High-to-Low clock (C) transitions. The counter increments (shifts Q0 to Q1, Q1 to Q2, etc.) when the clock enable input (CE) is High during the Low-to-High and High-to-Low clock transition. Clock transitions are ignored when CE is Low.

The Q7 output is inverted and fed back to input Q0 to provide continuous counting operation.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs			Outputs	
R	CE	С	Q0	Q1 – Q7
1	X	-	0	0
1	X	-	0	0
0	0	X	No Change	No Change
0	1	-	!q7	q0 – q6
0	1	-	!q7	q0 – q6
q = state of ref	ferenced output one setu	p time prior to active clo	ck transition	•

## **Design Entry Method**

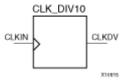
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK\_DIV10**

#### Primitive: Simple Global Clock Divide by 10



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 10.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

# **Verilog Instantiation Template**

```
// CLK_DIV10: Simple Clock Divide by 10
```



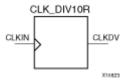
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV10 CLK_DIV10_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV10_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CLK\_DIV10R

Primitive: Global Clock Divide by 10 with Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 10.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV10R: Clock Divide by 10 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV10R_inst : CLK_DIV10R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV10R_inst instantiation
```



# **Verilog Instantiation Template**

```
// CLK_DIV10R: Clock Divide by 10 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV10R CLK_DIV10R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

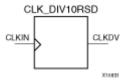
// End of CLK_DIV10R_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV10RSD**

Primitive: Global Clock Divide by 10 with Synchronous Reset and Start Delay



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 10.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV10RSD: Clock Divide by 10 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV10RSD_inst: CLK_DIV10RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV10RSD_inst instantiation
```

## **Verilog Instantiation Template**

```
// CLK_DIV12RSD: Clock Divide by 12 with Synchronous Reset and Start
// Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV12RSD CLK_DIV12RSD_inst (
    .CLKDV(CLKDV), // Divided clock output
    .CDRST(CDRST), // Synchronous reset input
    .CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

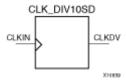
defparam CLK_DIV12RSD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV12RSD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV10SD**

Primitive: Global Clock Divide by 10 with Start Delay



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 10.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV10SD: Clock Divide by 10 with Start Delay
-- CoolRunner-II

-- Xilinx HDL Language Template, version 10.1

CLK_DIV10SD_inst : CLK_DIV10SD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV10SD_inst instantiation
```



# **Verilog Instantiation Template**

```
// CLK_DIV10SD: Clock Divide by 10 with Start Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV10SD CLK_DIV10SD_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

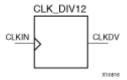
defparam CLK_DIV10SD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV10SD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CLK\_DIV12

#### Primitive: Simple Global Clock Divide by 12



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 12.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- CLK_DIV12: Simple Clock Divide by 12
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV12_inst : CLK_DIV12
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV12_inst instantiation
```

## **Verilog Instantiation Template**

```
// CLK_DIV12: Simple Clock Divide by 12
```



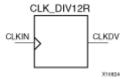
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV12 CLK_DIV12_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV12_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CLK\_DIV12R

Primitive: Global Clock Divide by 12 with Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 12.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- CLK_DIV12R: Clock Divide by 12 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV12R_inst : CLK_DIV12R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV12R_inst instantiation
```



# **Verilog Instantiation Template**

```
// CLK_DIV12R: Clock Divide by 12 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV12R CLK_DIV12R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

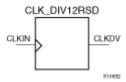
// End of CLK_DIV12R_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV12RSD**

Primitive: Global Clock Divide by 12 with Synchronous Reset and Start Delay



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 12.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV12RSD: Clock Divide by 12 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV12RSD_inst : CLK_DIV12RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV12RSD_inst instantiation
```

## **Verilog Instantiation Template**

```
// CLK_DIV12RSD: Clock Divide by 12 with Synchronous Reset and Start
// Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV12RSD CLK_DIV12RSD_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

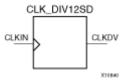
defparam CLK_DIV12RSD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV12RSD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV12SD**

#### Primitive: Global Clock Divide by 12 with Start Delay



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 12.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV12SD: Clock Divide by 12 with Start Delay
-- CoolRunner-II

-- Xilinx HDL Language Template, version 10.1

CLK_DIV12SD_inst: CLK_DIV12SD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV12SD_inst instantiation
```



# **Verilog Instantiation Template**

```
// CLK_DIV12SD: Clock Divide by 12 with Start Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV12SD CLK_DIV12SD_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

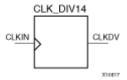
defparam CLK_DIV12SD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV12SD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV14**

#### Primitive: Simple Global Clock Divide by 14



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 14.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV14: Simple Clock Divide by 14

-- CoolRunner-II

-- Xilinx HDL Language Template, version 10.1

CLK_DIV14_inst : CLK_DIV14
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV14_inst instantiation
```

# **Verilog Instantiation Template**

```
// CLK_DIV14: Simple Clock Divide by 14
```



240

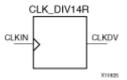
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV14 CLK_DIV14_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV14_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV14R**

Primitive: Global Clock Divide by 14 with Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element divides a user-provided external clock signal gclk<2> by 14.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

## **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV14R: Clock Divide by 14 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV14R_inst : CLK_DIV14R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV14R_inst instantiation
```



# **Verilog Instantiation Template**

```
// CLK_DIV14R: Clock Divide by 14 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV14R CLK_DIV14R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

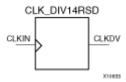
// End of CLK_DIV14R_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CLK\_DIV14RSD

Primitive: Global Clock Divide by 14 with Synchronous Reset and Start Delay



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 14.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV14RSD: Clock Divide by 14 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV14RSD_inst: CLK_DIV14RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV14RSD_inst instantiation
```

```
// CLK_DIV14RSD: Clock Divide by 14 with Synchronous Reset and Start
// Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV14RSD CLK_DIV14RSD_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

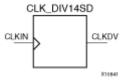
defparam CLK_DIV14RSD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV14RSD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV14SD**

#### Primitive: Global Clock Divide by 14 with Start Delay



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 14.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### VHDL Instantiation Template

```
use UNISIM.vcomponents.all;

-- CLK_DIV14SD: Clock Divide by 14 with Start Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV14SD_inst: CLK_DIV14SD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV14SD_inst instantiation
```



```
// CLK_DIV14SD: Clock Divide by 14 with Start Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV14SD CLK_DIV14SD_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

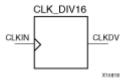
defparam CLK_DIV14SD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV14SD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CLK\_DIV16

#### Primitive: Simple Global Clock Divide by 16



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 16.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

When using this component, the dedicated clock divider reset pin on the device is reserved and may not be used by user logic.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV16: Simple Clock Divide by 16

-- CoolRunner-II

-- Xilinx HDL Language Template, version 10.1

CLK_DIV16_inst : CLK_DIV16
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV16_inst instantiation
```



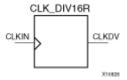
```
// CLK_DIV16: Simple Clock Divide by 16
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV16 CLK_DIV16_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV16_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV16R**

Primitive: Global Clock Divide by 16 with Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 16.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV16R: Clock Divide by 16 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV16R_inst: CLK_DIV16R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV16R_inst instantiation
```

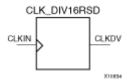


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV16RSD**

Primitive: Global Clock Divide by 16 with Synchronous Reset and Start Delay



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 16.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV16RSD: Clock Divide by 16 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV16RSD_inst: CLK_DIV16RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV16RSD_inst instantiation
```

```
// CLK_DIV16RSD: Clock Divide by 16 with Synchronous Reset and Start
// Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV16RSD CLK_DIV16RSD_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

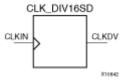
defparam CLK_DIV16RSD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV16RSD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV16SD**

#### Primitive: Global Clock Divide by 16 with Start Delay



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 16.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

-- End of CLK\_DIV16SD\_inst instantiation

```
use UNISIM.vcomponents.all;

-- CLK_DIV16SD: Clock Divide by 16 with Start Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV16SD_inst : CLK_DIV16SD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);
```

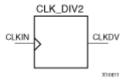


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV2**

#### Primitive: Simple Global Clock Divide by 2



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 2.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

# **Verilog Instantiation Template**

```
// CLK_DIV2: Simple Clock Divide by 2
```



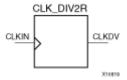
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV2 CLK_DIV2_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV2_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV2R**

Primitive: Global Clock Divide by 2 with Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 2.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;
-- CLK_DIV2R: Clock Divide by 2 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV2R_inst : CLK_DIV2R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV2R_inst instantiation
```



```
// CLK_DIV2R: Clock Divide by 2 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV2R CLK_DIV2R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// End of CLK_DIV2R_inst instantiation
```

### **For More Information**

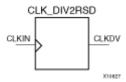
- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.

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# **CLK DIV2RSD**

Primitive: Global Clock Divide by 2 with Synchronous Reset and Start Delay



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 2.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV2RSD: Clock Divide by 2 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV2RSD_inst: CLK_DIV2RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV2RSD_inst instantiation
```

#### For More Information

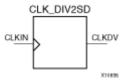
260

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# CLK\_DIV2SD

#### Primitive: Global Clock Divide by 2 with Start Delay



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 2.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# VHDL Instantiation Template

-- End of CLK\_DIV2SD\_inst instantiation

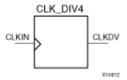


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV4**

#### Primitive: Simple Global Clock Divide by 4



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 4.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

# **Verilog Instantiation Template**

```
// CLK_DIV4: Simple Clock Divide by 4
```



264

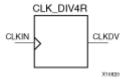
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV4 CLK_DIV4_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV4_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV4R**

Primitive: Global Clock Divide by 4 with Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element divides a user-provided external clock signal gclk<2> by 4.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV4R: Clock Divide by 4 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV4R_inst : CLK_DIV4R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV4R_inst instantiation
```



```
// CLK_DIV4R: Clock Divide by 4 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV4R CLK_DIV4R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

// End of CLK_DIV4R_inst instantiation
```

### **For More Information**

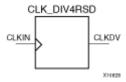
- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.

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# **CLK DIV4RSD**

Primitive: Global Clock Divide by 4 with Synchronous Reset and Start Delay



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 4.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV4RSD: Clock Divide by 4 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV4RSD_inst : CLK_DIV4RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV4RSD_inst instantiation
```

#### For More Information

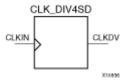
268

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV4SD**

#### Primitive: Global Clock Divide by 4 with Start Delay



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 4.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### VHDL Instantiation Template

```
use UNISIM.vcomponents.all;
-- CLK_DIV4SD: Clock Divide by 4 with Start Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV4SD_inst: CLK_DIV4SD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV4SD_inst instantiation
```

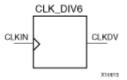


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV6**

#### Primitive: Simple Global Clock Divide by 6



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 6.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- CLK_DIV6: Simple Clock Divide by 6
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV6_inst : CLK_DIV6
port map (
CLKDV => CLKDV, -- Divided clock output
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV6_inst instantiation
```

# **Verilog Instantiation Template**

```
// CLK_DIV6: Simple Clock Divide by 6
```



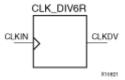
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV6 CLK_DIV6_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV6_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV6R**

Primitive: Global Clock Divide by 6 with Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 6.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV6R: Clock Divide by 6 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV6R_inst : CLK_DIV6R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);

-- End of CLK_DIV6R_inst instantiation
```



```
// CLK_DIV6R: Clock Divide by 6 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV6R CLK_DIV6R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

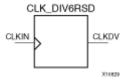
// End of CLK_DIV6R_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK\_DIV6RSD**

Primitive: Global Clock Divide by 6 with Synchronous Reset and Start Delay



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 6.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

```
use UNISIM.vcomponents.all;

-- CLK_DIV6RSD: Clock Divide by 6 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV6RSD_inst : CLK_DIV6RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV6RSD_inst instantiation
```

```
// CLK_DIV6RSD: Clock Divide by 6 with Synchronous Reset and Start
// Delay
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV6RSD CLK_DIV6RSD_inst (
    .CLKDV(CLKDV), // Divided clock output
    .CDRST(CDRST), // Synchronous reset input
    .CLKIN(CLKIN) // Clock input
);

// Edit the following defparam to specify the number of clock
// cycles to delay before starting. If the instance name to
// the clock divider is changed, that change needs to be
// reflected in the defparam statements.

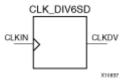
defparam CLK_DIV6RSD_inst.DIVIDER_DELAY = 1;
// End of CLK_DIV6RSD_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV6SD**

#### Primitive: Global Clock Divide by 6 with Start Delay



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 6.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

# **VHDL Instantiation Template**

-- End of CLK\_DIV4SD\_inst instantiation

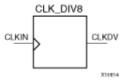


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV8**

#### Primitive: Simple Global Clock Divide by 8



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 8.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CLKDV output is reset low by power-on reset circuitry.

### **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

# **Verilog Instantiation Template**

```
// CLK_DIV8: Simple Clock Divide by 8
```



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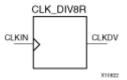
```
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1
CLK_DIV8 CLK_DIV8_inst (
.CLKDV(CLKDV), // Divided clock output
.CLKIN(CLKIN) // Clock input
);
// End of CLK_DIV8_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK\_DIV8R**

Primitive: Global Clock Divide by 8 with Synchronous Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 8.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- CLK_DIV8R: Clock Divide by 8 with Synchronous Reset
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV8R_inst : CLK_DIV8R
port map (
CLKDV => CLKDV, -- Divided clock output
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV8R_inst instantiation
```



# **Verilog Instantiation Template**

```
// CLK_DIV8R: Clock Divide by 8 with Synchronous Reset
// CoolRunner-II
// Xilinx HDL Language Template, version 10.1

CLK_DIV8R CLK_DIV8R_inst (
.CLKDV(CLKDV), // Divided clock output
.CDRST(CDRST), // Synchronous reset input
.CLKIN(CLKIN) // Clock input
);

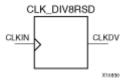
// End of CLK_DIV8R_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV8RSD**

Primitive: Global Clock Divide by 8 with Synchronous Reset and Start Delay



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 8.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner-II devices, but not the XC2C32 or XC2C64. The CLKIN and CDRST inputs can only be connected to the device gclk<2> and CDRST pins. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The CDRST input is an active High synchronous reset. If CDRST is input High when the CLKDV output is High, the CLKDV output remains High to complete the last clock pulse, and then goes Low.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;

-- CLK_DIV8RSD: Clock Divide by 8 with Synchronous Reset and Start
-- Delay
-- CoolRunner-II
-- Xilinx HDL Language Template, version 10.1

CLK_DIV8RSD_inst: CLK_DIV8RSD
-- Edit the following generic to specify the number of clock cycles
-- to delay before starting.
generic map (
DIVIDER_DELAY => 1)
port map (
CLKDV => CLKDV, -- Divided clock output
```



```
CDRST => CDRST, -- Synchronous reset input
CLKIN => CLKIN -- Clock input
);
-- End of CLK_DIV8RSD_inst instantiation
```

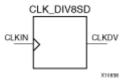
### **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **CLK DIV8SD**

#### Primitive: Global Clock Divide by 8 with Start Delay



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element divides a user-provided external clock signal gclk<2> by 8.

Only one clock divider may be used per design. The global clock divider is available on the XC2C128, XC2C256, XC2C384, and XC2C512 CoolRunner<sup>TM</sup>-IIdevices, but not the XC2C32 or XC2C64. The CLKIN input can only be connected to the device gclk<2> pin. The duty cycle of the CLKDV output is 50-50. The CLKDV output can only connect to clock inputs of synchronous elements. It cannot be used as combinatorial logic, and should not be routed directly to an output pin.

The start delay function delays the start of the CLKDV output by (n + 1) clocks, where n is the divisor for the clock divider.

The CLKDV output is reset low by power-on reset circuitry.

# **Design Entry Method**

Instantiation	Recommended
Inference	No
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### VHDL Instantiation Template

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

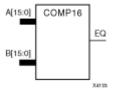


### **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 16-Bit Identity Comparator



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 16-bit identity comparator. The equal output (EQ) is high when A15 – A0 and B15 – B0 are equal.

Equality is determined by a bit comparison of the two words. When any two of the corresponding bits from each word are not the same, the EQ output is Low.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 2-Bit Identity Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 2-bit identity comparator. The equal output (EQ) is High when the two words A1 - A0 and B1 - B0 are equal.

Equality is determined by a bit comparison of the two words. When any two of the corresponding bits from each word are not the same, the EQ output is Low.

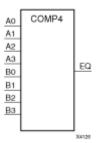
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 4-Bit Identity Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 4-bit identity comparator. The equal output (EQ) is high when A3 – A0 and B3 – B0 are equal.

Equality is determined by a bit comparison of the two words. When any two of the corresponding bits from each word are not the same, the EQ output is Low.

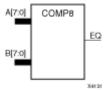
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 8-Bit Identity Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an 8-bit identity comparator. The equal output (EQ) is high when A7 – A0 and B7 – B0 are equal.

Equality is determined by a bit comparison of the two words. When any two of the corresponding bits from each word are not the same, the EQ output is Low.

# **Design Entry Method**

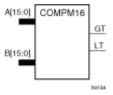
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# COMPM<sub>16</sub>

#### Macro: 16-Bit Magnitude Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 16-bit magnitude comparator that compare two positive Binary-weighted words. It compares A15 – A0 and B15 – B0, where A15 and B15 are the most significant bits.

The greater-than output (GT) is High when A > B, and the less-than output (LT) is High when A < B When the two words are equal, both GT and LT are Low. Equality can be measured with this macro by comparing both outputs with a NOR gate.

### Logic Table

See COMPM8 for a representative logic table.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# COMPM2

#### Macro: 2-Bit Magnitude Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 2-bit magnitude comparator that compare two positive Binary-weighted words. It compares A1 - A0 and B1 - B0, where A1 and B1 are the most significant bits.

The greater-than output (GT) is High when A > B, and the less-than output (LT) is High when A < B When the two words are equal, both GT and LT are Low. Equality can be measured with this macro by comparing both outputs with a NOR gate.

### **Logic Table**

Inputs		Outputs			
A1	B1	A0	В0	GT	LT
0	0	0	0	0	0
0	0	1	0	1	0
0	0	0	1	0	1
0	0	1	1	0	0
1	1	0	0	0	0
1	1	1	0	1	0
1	1	0	1	0	1
1	1	1	1	0	0
1	0	Χ	X	1	0
0	1	Χ	Χ	0	1

### **Design Entry Method**

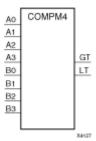
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### COMPM4

#### Macro: 4-Bit Magnitude Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 4-bit magnitude comparator that compare two positive Binary-weighted words. It compares A3 – A0 and B3 – B0, where A3 and B3 are the most significant bits.

The greater-than output (GT) is High when A > B, and the less-than output (LT) is High when A < B When the two words are equal, both GT and LT are Low. Equality can be measured with this macro by comparing both outputs with a NOR gate.

### **Logic Table**

Inputs				Outputs	
A3, B3	A2, B2	A1, B1	A0, B0	GT	LT
A3>B3	Χ	Х	X	1	0
A3 <b3< td=""><td>Χ</td><td>Х</td><td>X</td><td>0</td><td>1</td></b3<>	Χ	Х	X	0	1
A3=B3	A2>B2	Х	X	1	0
A3=B3	A2 <b2< td=""><td>Х</td><td>X</td><td>0</td><td>1</td></b2<>	Х	X	0	1
A3=B3	A2=B2	A1>B1	X	1	0
A3=B3	A2=B2	A1 <b1< td=""><td>X</td><td>0</td><td>1</td></b1<>	X	0	1
A3=B3	A2=A2	A1=B1	A0>B0	1	0
A3=B3	A2=B2	A1=B1	A0 <b0< td=""><td>0</td><td>1</td></b0<>	0	1
A3=B3	A2=B2	A1=B1	A0=B0	0	0

### **Design Entry Method**

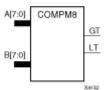
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# COMPM8

Macro: 8-Bit Magnitude Comparator



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an 8-bit magnitude comparator that compare two positive Binary-weighted words. It compares A7 - A0 and B7 - B0, where A7 and B7 are the most significant bits.

The greater-than output (GT) is High when A > B, and the less-than output (LT) is High when A < B When the two words are equal, both GT and LT are Low. Equality can be measured with this macro by comparing both outputs with a NOR gate.

# Logic Table

Inputs						Outputs			
A7, B7	A6, B6	A5, B5	A4, B4	A3, B3	A2, B2	A1, B1	A0, B0	GT	LT
A7>B7	Х	Х	Χ	Х	Х	Х	Х	1	0
A7 <b7< td=""><td>Х</td><td>Х</td><td>Χ</td><td>Х</td><td>Х</td><td>Х</td><td>Χ</td><td>0</td><td>1</td></b7<>	Х	Х	Χ	Х	Х	Х	Χ	0	1
A7=B7	A6>B6	Х	Х	Х	Х	Х	Х	1	0
A7=B7	A6 <b6< td=""><td>Х</td><td>Χ</td><td>Х</td><td>Х</td><td>Х</td><td>Х</td><td>0</td><td>1</td></b6<>	Х	Χ	Х	Х	Х	Х	0	1
A7=B7	A6=B6	A5>B5	Χ	Х	Х	Х	Х	1	0
A7=B7	A6=B6	A5 <b5< td=""><td>Х</td><td>Х</td><td>Х</td><td>Х</td><td>Х</td><td>0</td><td>1</td></b5<>	Х	Х	Х	Х	Х	0	1
A7=B7	A6=B6	A5=B5	A4>B4	Х	Х	Х	Х	1	0
A7=B7	A6=B6	A5=B5	A4 <b4< td=""><td>Х</td><td>Х</td><td>Х</td><td>Х</td><td>0</td><td>1</td></b4<>	Х	Х	Х	Х	0	1
A7=B7	A6=B6	A5=B5	A4=B4	A3>B3	Х	Х	Χ	1	0
A7=B7	A6=B6	A5=B5	A4=B4	A3 <b3< td=""><td>Х</td><td>Х</td><td>Х</td><td>0</td><td>1</td></b3<>	Х	Х	Х	0	1
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2>B2	Х	Х	1	0
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2 <b2< td=""><td>Х</td><td>Х</td><td>0</td><td>1</td></b2<>	Х	Х	0	1
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2=B2	A1>B1	Χ	1	0
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2=B2	A1 <b1< td=""><td>Χ</td><td>0</td><td>1</td></b1<>	Χ	0	1
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2=B2	A1=B1	A0>B0	1	0
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2=B2	A1=B1	A0 <b0< td=""><td>0</td><td>1</td></b0<>	0	1
A7=B7	A6=B6	A5=B5	A4=B4	A3=B3	A2=B2	A1=B1	A0=B0	0	0



# **Design Entry Method**

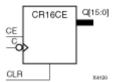
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CR16CE

Macro: 16-Bit Negative-Edge Binary Ripple Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 16-bit cascadable, clearable, binary ripple counter with clock enable and asynchronous clear.

Larger counters can be created by connecting the last Q output of the first stage to the clock input of the next stage. CLR and CE inputs are connected in parallel. The clock period is not affected by the overall length of a ripple counter. The overall clock-to-output propagation is  $n(t_{C-Q})$ , where n is the number of stages and the time  $t_{C-Q}$  is the C-to-Qz propagation delay of each stage.

This counter is asynchronously cleared, outputs Low, when power is applied. .

### **Logic Table**

Inputs			Outputs
CLR	CE	С	Qz - Q0
1	X	X	0
0	0	X	No Change
0	1	Ø	Inc

z = bit width - 1

# **Design Entry Method**

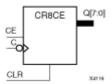
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## CR8CE

Macro: 8-Bit Negative-Edge Binary Ripple Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an 8-bit cascadable, clearable, binary, ripple counter with clock enable and asynchronous clear.

The asynchronous clear (CLR), when High, overrides all other inputs and causes the Q outputs to go to logic level zero. The counter increments when the clock enable input (CE) is High during the High-to-Low clock (C) transition. The counter ignores clock transitions when CE is Low.

Larger counters can be created by connecting the last Q output of the first stage to the clock input of the next stage. CLR and CE inputs are connected in parallel. The clock period is not affected by the overall length of a ripple counter. The overall clock-to-output propagation is  $n(t_{C-Q})$ , where n is the number of stages and the time  $t_{C-Q}$  is the C-to-Qz propagation delay of each stage.

This counter is asynchronously cleared, outputs Low, when power is applied. .

### **Logic Table**

Inputs			Outputs
CLR	CE	С	Qz - Q0
1	X	X	0
0	0	X	No Change
0	1	Ø	Inc

z = bit width - 1

### **Design Entry Method**

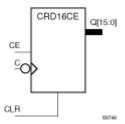
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CRD16CE

Macro: 16-Bit Dual-Edge Triggered Binary Ripple Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered 16-bit cascadable, clearable, binary ripple counter.

The asynchronous clear (CLR), when High, overrides all other inputs and causes the Q outputs to go to logic level zero. The counter increments when the clock enable input (CE) is High during the High-to-Low and Low-to-High clock (C) transitions. The counter ignores clock transitions when CE is Low.

Larger counters can be created by connecting the last Q output of the first stage to the clock input of the next stage. CLR and CE inputs are connected in parallel. The clock period is not affected by the overall length of a ripple counter. The overall clock-to-output propagation is  $n(t_{C-Q})$ , where n is the number of stages and the time  $t_{C-Q}$  is the C-to-Qz propagation delay of each stage.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs			Outputs
CLR	CE	С	Qz - Q0
1	X	X	0
0	0	X	No Change
0	1	-	Inc
0	1	-	Inc

z = bit width - 1

# **Design Entry Method**

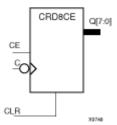
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### CRD8CE

Macro: 8-Bit Dual-Edge Triggered Binary Ripple Counter with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered 8-bit cascadable, clearable, binary ripple counter.

The asynchronous clear (CLR), when High, overrides all other inputs and causes the Q outputs to go to logic level zero. The counter increments when the clock enable input (CE) is High during the High-to-Low and Low-to-High clock (C) transitions. The counter ignores clock transitions when CE is Low.

Larger counters can be created by connecting the last Q output of the first stage to the clock input of the next stage. CLR and CE inputs are connected in parallel. The clock period is not affected by the overall length of a ripple counter. The overall clock-to-output propagation is  $n(t_{C-Q})$ , where n is the number of stages and the time  $t_{C-Q}$  is the C-to-Qz propagation delay of each stage.

This counter is asynchronously cleared, outputs Low, when power is applied. .For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Inputs			Outputs
CLR	CE	С	Qz - Q0
1	X	X	0
0	0	X	No Change
0	1	-	Inc
0	1	-	Inc

z = bit width - 1

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# D2 4E

#### Macro: 2- to 4-Line Decoder/Demultiplexer with Enable



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a decoder/demultiplexer. When the enable (E) input of this element is High, one of four active-High outputs (D3 - D0) is selected with a 2-bit binary address (A1 - A0) input. The non-selected outputs are Low. Also, when the E input is Low, all outputs are Low. In demultiplexer applications, the E input is the data input.

## **Logic Table**

Inputs			Outputs	Outputs			
A1	A0	E	D3	D2	D1	D0	
Х	Х	0	0	0	0	0	
0	0	1	0	0	0	1	
0	1	1	0	0	1	0	
1	0	1	0	1	0	0	
1	1	1	1	0	0	0	

# **Design Entry Method**

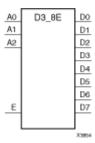
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# D3\_8E

#### Macro: 3- to 8-Line Decoder/Demultiplexer with Enable



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

When the enable (E) input of the D3\_8E decoder/demultiplexer is High, one of eight active-High outputs (D7 – D0) is selected with a 3-bit binary address (A2 – A0) input. The non-selected outputs are Low. Also, when the E input is Low, all outputs are Low. In demultiplexer applications, the E input is the data input.

# **Logic Table**

Inputs			Outputs								
A2	A1	A0	E	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Х	X	0	0	0	0	0	0	0	0	0
0	0	0	1	0	0	0	0	0	0	0	1
0	0	1	1	0	0	0	0	0	0	1	0
0	1	0	1	0	0	0	0	0	1	0	0
0	1	1	1	0	0	0	0	1	0	0	0
1	0	0	1	0	0	0	1	0	0	0	0
1	0	1	1	0	0	1	0	0	0	0	0
1	1	0	1	0	1	0	0	0	0	0	0
1	1	1	1	1	0	0	0	0	0	0	0

#### **Design Entry Method**

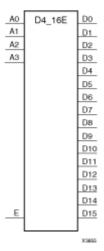
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **D4\_16E**

#### Macro: 4- to 16-Line Decoder/Demultiplexer with Enable



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a decoder/demultiplexer. When the enable (E) input of this design element is High, one of 16 active-High outputs (D15 – D0) is selected with a 4-bit binary address (A3 – A0) input. The non-selected outputs are Low. Also, when the E input is Low, all outputs are Low. In demultiplexer applications, the E input is the data input.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FD

#### Macro: D Flip-Flop



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a D-type flip-flop with data input (D) and data output (Q). The data on the D inputs is loaded into the flip-flop during the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Outputs	
D	С	Q
0	-	0
1	-	1

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

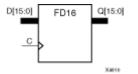
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	0, 1	0	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FD16**

#### Macro: Multiple D Flip-Flop



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a multiple D-type flip-flops with data inputs (D) and data outputs (Q), with a 16-bit register, each with a common clock (C). The data on the D inputs is loaded into the flip-flop during the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs		Outputs		
Dz – D0	С	Qz – Q0		
0	-	0		
1	-	1bit-width - 1		
z = bit-width - 1				

# **Design Entry Method**

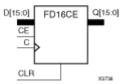
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## FD16CE

#### Macro: 16-Bit Data Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 16-bit data register with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q) during the Low-to-High clock (C) transition. When CLR is High, it overrides all other inputs and resets the data outputs (Q) Low. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
CLR	CE	Dz – D0	С	Qz - Q0	
1	Χ	Χ	Χ	0	
0	0	X	Χ	No Change	
0	1	Dn	-	Dn	
z = bit-width - 1					

## **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

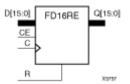
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	16-bit Binary	All zeros	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FD16RE

Macro: 16-Bit Data Register with Clock Enable and Synchronous Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 16-bit data registers. When the clock enable (CE) input is High, and the synchronous reset (R) input is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q0) during the Low-to-High clock (C) transition. When R is High, it overrides all other inputs and resets the data outputs (Q) Low on the Low-to-High clock transition. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
R	CE	Dz - D0	С	Qz - Q0	
1	X	X	-	0	
0	0	Χ	Χ	No Change	
0	1	Dn	-	Dn	
z = bit-width - 1					

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

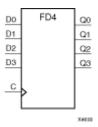
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	16-bit Binary	All zeros	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FD4

#### Macro: Multiple D Flip-Flop



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a multiple D-type flip-flops with data inputs (D) and data outputs (Q), with a 4-bit register, each with a common clock (C). The data on the D inputs is loaded into the flip-flop during the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Outputs			
Dz – D0	С	Qz – Q0		
0	-	0		
1	-	1bit-width - 1		
z = bit-width - 1				

### **Design Entry Method**

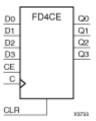
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FD4CE

Macro: 4-Bit Data Register with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 4-bit data register with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q) during the Low-to-High clock (C) transition. When CLR is High, it overrides all other inputs and resets the data outputs (Q) Low. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
CLR	CE	Dz - D0	С	Qz - Q0	
1	Χ	Χ	Χ	0	
0	0	Χ	X	No Change	
0	1	Dn	-	Dn	
z = bit-width - 1					

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	Binary	4-bit Binary	All zeros	Sets the initial value of Q output after configuration

#### For More Information

• See the appropriate CPLD User Guide.

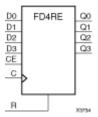


• See the appropriate CPLD Data Sheets.



### FD4RE

Macro: 4-Bit Data Register with Clock Enable and Synchronous Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 4-bit data registers. When the clock enable (CE) input is High, and the synchronous reset (R) input is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q0) during the Low-to-High clock (C) transition. When R is High, it overrides all other inputs and resets the data outputs (Q) Low on the Low-to-High clock transition. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
R	CE	Dz - D0	С	Qz - Q0	
1	X	X	-	0	
0	0	Χ	Χ	No Change	
0	1	Dn	-	Dn	
z = bit-width - 1					

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	Binary	4-bit Binary	All zeros	Sets the initial value of Q output after configuration

#### For More Information

• See the appropriate CPLD User Guide.

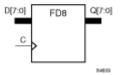


• See the appropriate CPLD Data Sheets.



### FD8

#### Macro: Multiple D Flip-Flop



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a multiple D-type flip-flops with data inputs (D) and data outputs (Q), with a 8-bit register, each with a common clock (C). The data on the D inputs is loaded into the flip-flop during the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs		Outputs	
Dz – D0	С	Qz – Q0	
0	-	0	
1	-	1bit-width - 1	
z = bit-width - 1			

# **Design Entry Method**

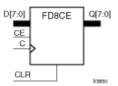
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FD8CE

#### Macro: 8-Bit Data Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 8-bit data register with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q) during the Low-to-High clock (C) transition. When CLR is High, it overrides all other inputs and resets the data outputs (Q) Low. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs			Outputs	
CLR	CE	Dz - D0	С	Qz - Q0
1	Χ	Χ	Χ	0
0	0	Χ	Χ	No Change
0	1	Dn	-	Dn
z = bit-width - 1				

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

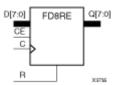
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	8-bit Binary	All zeros	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FD8RE

Macro: 8-Bit Data Register with Clock Enable and Synchronous Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is an 8-bit data register. When the clock enable (CE) input is High, and the synchronous reset (R) input is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q0) during the Low-to-High clock (C) transition. When R is High, it overrides all other inputs and resets the data outputs (Q) Low on the Low-to-High clock transition. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs			Outputs	
R	CE	Dz - D0	С	Qz - Q0
1	X	X	-	0
0	0	Χ	Χ	No Change
0	1	Dn	-	Dn
z = bit-width - 1				

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	Binary	Any 8-Bit Value	All zeros	Sets the initial value of Q output after configuration.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDC**

#### Macro: D Flip-Flop with Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with data (D) and asynchronous clear (CLR) inputs and data output (Q). The asynchronous CLR, when High, overrides all other inputs and sets the (Q) output Low. The data on the (D) input is loaded into the flip-flop when CLR is Low on the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs			Outputs
CLR	D	С	Q
1	X	X	0
0	D	-	D

# **Design Entry Method**

This design element is only for use in schematics.

### **Available Attributes**

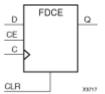
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	0, 1	0	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDCE**

Primitive: D Flip-Flop with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data input (D) of this design element is transferred to the corresponding data output (Q) during the Low-to-High clock (C) transition. When CLR is High, it overrides all other inputs and resets the data output (Q) Low. When CE is Low, clock transitions are ignored.

For XC9500XL and XC9500XV devices, logic connected to the clock enable (CE) input may be implemented using the clock enable product term (p-term) in the macrocell, provided the logic can be completely implemented using the single p-term available for clock enable without requiring feedback from another macrocell. Only FDCE and FDPE flip-flops may take advantage of the clock-enable p-term.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### Logic Table

Inputs	Outputs			
CLR	CE	D	С	Q
1	Χ	Χ	Χ	0
0	0	Χ	Χ	No Change
0	1	D	-	D

### **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **Available Attributes**



### **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- FDCE: Single Data Rate D Flip-Flop with Asynchronous Clear and
         Clock Enable (posedge clk). All families.
-- Xilinx HDL Libraries Guide, version 10.1.1
FDCE_inst : FDCE
generic map (
INIT => '0') -- Initial value of register ('0' or '1')
port map (
Q => Q,
             -- Data output
C => C,
            -- Clock input
CE => CE, -- Clock enable input
CLR => CLR, -- Asynchronous clear input
             -- Data input
D => D
);
-- End of FDCE_inst instantiation
```

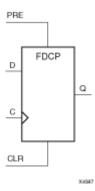
# **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDCP**

#### Primitive: D Flip-Flop with Asynchronous Preset and Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with data (D), asynchronous preset (PRE) and clear (CLR) inputs, and data output (Q). The asynchronous PRE, when High, sets the (Q) output High; CLR, when High, resets the output Low. Data on the (D) input is loaded into the flip-flop when PRE and CLR are Low on the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs				Outputs
CLR	PRE	D	С	Q
1	Χ	Χ	X	0
0	1	Χ	X	1
0	0	D	-	D

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration

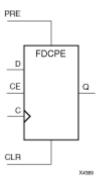


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDCPE**

Primitive: D Flip-Flop with Clock Enable and Asynchronous Preset and Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with data (D), clock enable (CE), asynchronous preset (PRE), and asynchronous clear (CLR) inputs. The asynchronous active high PRE sets the Q output High; that active high CLR resets the output Low and has precedence over the PRE input. Data on the D input is loaded into the flip-flop when PRE and CLR are Low and CE is High on the Low-to-High clock (C) transition. When CE is Low, the clock transitions are ignored and the previous value is retained. The FDCPE is generally implemented as a slice or IOB register within the device.

For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net. For FPGA devices, upon power-up, the initial value of this component is specified by the INIT attribute. If a subsequent GSR (Global Set/Reset) is asserted, the flop is asynchronously set to the INIT value.

**Note** While this device supports the use of asynchronous set and reset, it is not generally recommended to be used for in most cases. Use of asynchronous signals pose timing issues within the design that are difficult to detect and control and also have an adverse affect on logic optimization causing a larger design that can consume more power than if a synchronous set or reset is used.

### Logic Table

Inputs					Outputs
CLR	PRE	CE	D	С	Q
1	X	Χ	Χ	Χ	0
0	1	Χ	Χ	Χ	1
0	0	0	Χ	Χ	No Change
0	0	1	D	-	D



# **Port Descriptions**

Name	Direction	Width	Function
Q	Output	1-Bit	Data output
С	Input	1-Bit	Clock input
CE	Input	1-Bit	Clock enable input
CLR	Input	1-Bit	Asynchronous clear input
D	Input	1-Bit	Data input
PRE	Input	1-Bit	Asynchronous set input

### **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration and on GSR

### **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- FDCPE: Single Data Rate D Flip-Flop with Asynchronous Clear, Set and
          Clock Enable (posedge clk). All families.
-- Xilinx HDL Libraries Guide, version 10.1.1
FDCPE_inst : FDCPE
generic map (
INIT => '0') -- Initial value of register ('0' or '1')
port map (
Q => Q,
            -- Data output
C => C, -- Clock input
CE => CE, -- Clock enable input
CLR => CLR, -- Asynchronous clear input
D => D,
            -- Data input
PRE => PRE -- Asynchronous set input
-- End of FDCPE_inst instantiation
```

# **Verilog Instantiation Template**

// FDCPE: Single Data Rate D Flip-Flop with Asynchronous Clear, Set and

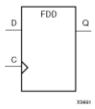


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDD**

### Macro: Dual Edge Triggered D Flip-Flop



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element is a single dual edge triggered D-type flip-flop with data input (D) and data output (Q). The data on the D input is loaded into the flip-flop during the Low-to-High and the High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

Inputs		Outputs
D	С	Q
0	-	0
1	-	1
0	-	0
1	-	1

### **Design Entry Method**

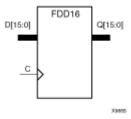
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD16

#### Macro: Multiple Dual Edge Triggered D Flip-Flop



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a multiple dual edge triggered D-type flip-flop with data inputs (D) and data outputs (Q). It is a 16-bit register with a common clock (C). The data on the D inputs is loaded into the flip-flop during the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs		Outputs	
Dz – D0	С	Qz – Q0	
0	-	0	
1	-	1	
0	-	0	
1	-	1	
z = bit-width - 1			

# **Design Entry Method**

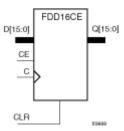
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD16CE

Macro: 16-Bit Dual Edge Triggered Data Register with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a 16-bit data registers with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q) during the Low-to-High and High-to-Low clock (C) transitions. When CLR is High, it overrides all other inputs and resets the data outputs (Q) Low. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs		Outputs		
CLR	CE	Dz - D0	С	Qz – Q0
1	X	X	X	0
0	0	X	X	No Change
0	1	Dn	-	Dn
0	1	Dn	-	Dn
z = bit-width - 1				

### **Design Entry Method**

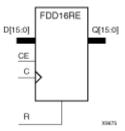
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD16RE

Macro: 16-Bit Dual Edge Triggered Data Register with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a 16-bit data register. When the clock enable (CE) input is High, and the synchronous reset (R) input is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q0) during the Low-to-High or High-to-Low clock (C) transition. When R is High, it overrides all other inputs and resets the data outputs (Q) Low on the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
R	CE	Dz – D0	С	Qz – Q0	
1	X	X	-	0	
1	X	X	-	0	
0	0	Х	X	No Change	
0	1	Dn	-	Dn	
0	1	Dn	-	Dn	
z = bit-width - 1					

### **Design Entry Method**

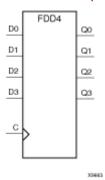
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD4

### Macro: Multiple Dual Edge Triggered D Flip-Flop



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a multiple dual edge triggered D-type flip-flop with data inputs (D) and data outputs (Q). It is a 4-bit register with a common clock (C). The data on the D inputs is loaded into the flip-flop during the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs			
Dz – D0	С	Qz – Q0		
0	-	0		
1	-	1		
0	-	0		
1	-	1		
z = bit-width - 1				

### **Design Entry Method**

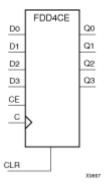
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FDD4CE

Macro: 4-Bit Dual Edge Triggered Data Register with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a 4-bit data registers with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q) during the Low-to-High and High-to-Low clock (C) transitions. When CLR is High, it overrides all other inputs and resets the data outputs (Q) Low. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### Logic Table

Inputs		Outputs		
CLR	CE	Dz - D0	С	Qz – Q0
1	X	X	X	0
0	0	X	X	No Change
0	1	Dn	-	Dn
0	1	Dn	-	Dn
z = bit-width - 1				

# **Design Entry Method**

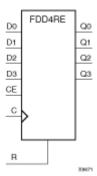
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD4RE

Macro: 4-Bit Dual Edge Triggered Data Register with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a 4-bit data register. When the clock enable (CE) input is High, and the synchronous reset (R) input is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q0) during the Low-to-High or High-to-Low clock (C) transition. When R is High, it overrides all other inputs and resets the data outputs (Q) Low on the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs			
R	CE	Dz – D0	С	Qz – Q0
1	X	X	-	0
1	X	X	-	0
0	0	Х	Х	No Change
0	1	Dn	-	Dn
0	1	Dn	-	Dn
z = bit-width	- 1	•	•	

# **Design Entry Method**

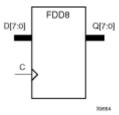
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### FDD8

#### Macro: Multiple Dual Edge Triggered D Flip-Flop



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element is a multiple dual edge triggered D-type flip-flop with data inputs (D) and data outputs (Q). It is an 8-bit register with a common clock (C). The data on the D inputs is loaded into the flip-flop during the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
Dz – D0	С	Qz – Q0		
0	-	0		
1	-	1		
0	-	0		
1	-	1		
z = bit-width - 1				

# **Design Entry Method**

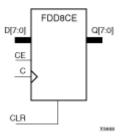
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD8CE

#### Macro: 8-Bit Dual Edge Triggered Data Register with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a 8-bit data registers with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q) during the Low-to-High and High-to-Low clock (C) transitions. When CLR is High, it overrides all other inputs and resets the data outputs (Q) Low. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Inputs		Outputs		
CLR	CE	Dz – D0	С	Qz - Q0
1	Χ	Χ	X	0
0	0	Χ	Χ	No Change
0	1	Dn	-	Dn
0	1	Dn	-	Dn
z = bit-width - 1				

#### **Design Entry Method**

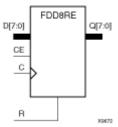
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# FDD8RE

Macro: 8-Bit Dual Edge Triggered Data Register with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a 8-bit data register. When the clock enable (CE) input is High, and the synchronous reset (R) input is Low, the data on the data inputs (D) is transferred to the corresponding data outputs (Q0) during the Low-to-High or High-to-Low clock (C) transition. When R is High, it overrides all other inputs and resets the data outputs (Q) Low on the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This register is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
R	CE	Dz – D0	С	Qz – Q0	
1	X	X	-	0	
1	X	X	-	0	
0	0	X	X	No Change	
0	1	Dn	-	Dn	
0	1	Dn	-	Dn	
z = bit-width - 1					

# **Design Entry Method**

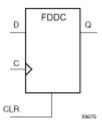
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDC**

#### Macro: D Dual Edge Triggered Flip-Flop with Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element is a single dual edge triggered D-type flip-flop with data (D) and asynchronous clear (CLR) inputs and data output (Q). The asynchronous CLR, when High, overrides all other inputs and sets the Q output Low. The data on the D input is loaded into the flip-flop when CLR is Low on the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs			Outputs
CLR	D	С	Q
1	X	X	0
0	1	-	1
0	1	-	1
0	0	-	0
0	0	-	0

### **Design Entry Method**

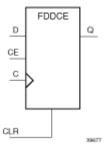
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FDDCE**

Primitive: Dual Edge Triggered D Flip-Flop with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a single dual edge triggered D-type flip-flop with clock enable and asynchronous clear. When clock enable (CE) is High and asynchronous clear (CLR) is Low, the data on the data input (D) of FDDCE is transferred to the corresponding data output (Q) during the Low-to-High and High-to-Low clock (C) transitions. When CLR is High, it overrides all other inputs and resets the data output (Q) Low. When CE is Low, clock transitions are ignored.

Logic connected to the clock enable (CE) input may be implemented using the clock enable product term (p-term) in the macrocell, provided the logic can be completely implemented using the single p-term available for clock enable without requiring feedback from another macrocell. Only FDDCE and FDDPE flip-flops can take advantage of the clock-enable p-term.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### **Port Descriptions**

Inputs				Outputs
CLR	CE	D	С	Q
1	X	X	X	0
0	0	X	Χ	No Change
0	1	1	-	1
0	1	0	-	0
0	1	1	-	1
0	1	0	-	0

### **Design Entry Method**

This design element is only for use in schematics.

### For More Information

• See the appropriate CPLD User Guide.

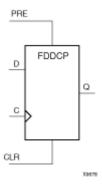


• See the appropriate CPLD Data Sheets.



# **FDDCP**

Primitive: Dual Edge Triggered D Flip-Flop Asynchronous Preset and Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a single dual edge triggered D-type flip-flop with data (D), asynchronous preset (PRE) and clear (CLR) inputs, and data output (Q). The asynchronous PRE, when High, sets the Q output High; CLR, when High, resets the output Low. Data on the D input is loaded into the flip-flop when PRE and CLR are Low on the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Port Descriptions**

Inputs		Outputs			
CLR	PRE	D	С	Q	
1	X	Χ	X	0	
0	1	Χ	X	1	
0	0	0	-	0	
0	0	1	-	1	
0	0	0	-	0	
0	0	1	-	1	

# **Design Entry Method**

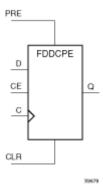
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDCPE**

Macro: Dual Edge Triggered D Flip-Flop with Clock Enable and Asynchronous Preset and Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a single dual edge triggered D-type flip-flop with data (D), clock enable (CE), asynchronous preset (PRE), and asynchronous clear (CLR) inputs and data output (Q). The asynchronous PRE, when High, sets the Q output High; CLR, when High, resets the output Low. Data on the D input is loaded into the flip-flop when PRE and CLR are Low and CE is High on the Low-to-High and High-to-Low clock (C) transitions. When CE is Low, the clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Port Descriptions**

Inputs		Outputs			
CLR	PRE	CE	D	С	Q
1	Х	Х	X	X	0
0	1	X	Х	X	1
0	0	0	X	X	No Change
0	0	1	0	-	0
0	0	1	1	-	1
0	0	1	0	-	0
0	0	1	1	-	1

### **Design Entry Method**

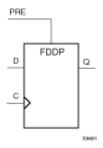
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDP**

#### Macro: Dual Edge Triggered D Flip-Flop with Asynchronous Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a single dual edge triggered D-type flip-flop with data (D) and asynchronous preset (PRE) inputs and data output (Q). The asynchronous PRE, when High, overrides all other inputs and presets the Q output High. The data on the D input is loaded into the flip-flop when PRE is Low on the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Port Descriptions**

Inputs			Outputs
PRE	С	D	Q
1	X	X	1
0	-	1	1
0	-	0	0
0	-	1	1
0	-	0	0

# **Design Entry Method**

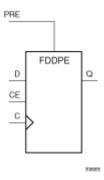
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDPE**

Primitive: Dual Edge Triggered D Flip-Flop with Clock Enable and Asynchronous Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a single dual edge triggered D-type flip-flop with data (D), clock enable (CE), and asynchronous preset (PRE) inputs and data output (Q). The asynchronous PRE, when High, overrides all other inputs and sets the Q output High. Data on the D input is loaded into the flip-flop when PRE is Low and CE is High on the Low-to-High and High-to-Low clock (C) transitions. When CE is Low, the clock transitions are ignored.

Logic connected to the clock enable (CE) input may be implemented using the clock enable product term (p-term) in the macrocell, provided the logic can be completely implemented using the single p-term available for clock enable without requiring feedback from another macrocell. Only FDDCE and FDDPE flip-flops primitives may take advantage of the clock-enable p-term.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs				Outputs
PRE	CE	D	С	Q
1	X	X	Χ	1
0	0	X	X	No Change
0	1	0	-	0
0	1	1	-	1
0	1	0	-	0
0	1	1	-	1

### **Design Entry Method**

This design element is only for use in schematics.

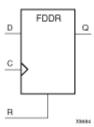


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDR**

#### Macro: Dual Edge Triggered D Flip-Flop with Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a single dual edge triggered D-type flip-flop with data (D) and synchronous reset (R) inputs and data output (Q). The synchronous reset (R) input, when High, overrides all other inputs and resets the Q output Low on the Low-to-High and High-to-Low clock (C) transitions. The data on the D input is loaded into the flip-flop when R is Low during the Low-to-High or High-to-Low clock transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs			Outputs
R	D	С	Q
1	X	-	0
1	X	-	0
0	1	-	1
0	0	-	0
0	1	-	1
0	0	-	0

### **Design Entry Method**

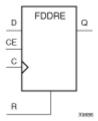
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDRE**

Macro: Dual Edge Triggered D Flip-Flop with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDRE is a single dual edge triggered D-type flip-flop with data (D), clock enable (CE), and synchronous reset (R) inputs and data output (Q). The synchronous reset (R) input, when High, overrides all other inputs and resets the Q output Low on the Low-to-High or High-to-Low clock (C) transition. The data on the D input is loaded into the flip-flop when R is Low and CE is High during the Low-to-High and High-to-Low clock transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs	Outputs			
R	CE	D	С	Q
1	X	Х	-	0
1	X	Х	-	0
0	0	Х	Х	No Change
0	1	1	-	1
0	1	0	-	0
0	1	1	-	1
0	1	0	-	0

### **Design Entry Method**

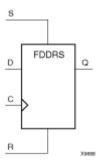
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDRS**

### Macro: Dual Edge Triggered D Flip-Flop with Synchronous Reset and Set



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDRS is a single dual edge triggered D-type flip-flop with data (D), synchronous set (S), and synchronous reset (R) inputs and data output (Q). The synchronous reset (R) input, when High, overrides all other inputs and resets the Q output Low during the Low-to-High or High-to-Low clock (C) transitions. (Reset has precedence over Set.) When S is High and R is Low, the flip-flop is set, output High, during the Low-to-High or High-to-Low clock transition. When R and S are Low, data on the (D) input is loaded into the flip-flop during the Low-to-High and High-to-Low clock transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Port Descriptions**

Inputs	Outputs			
R	S	D	С	Q
1	X	X	-	0
1	X	X	-	0
0	1	X	-	1
0	1	X	-	1
0	0	1	-	1
0	0	1	-	1
0	0	0	-	0
0	0	0	-	0

### **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

• See the appropriate CPLD User Guide.

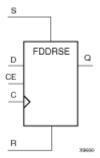


• See the appropriate CPLD Data Sheets.



# **FDDRSE**

Macro: Dual Edge Triggered D Flip-Flop with Synchronous Reset and Set and Clock Enable



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDRSE is a single dual edge triggered D-type flip-flop with synchronous reset (R), synchronous set (S), and clock enable (CE) inputs and data output (Q). The reset (R) input, when High, overrides all other inputs and resets the Q output Low during the Low-to-High or High-to-Low clock transitions. (Reset has precedence over Set.) When the set (S) input is High and R is Low, the flip-flop is set, output High, during the Low-to-High or High-to-Low clock (C) transition. Data on the D input is loaded into the flip-flop when R and S are Low and CE is High during the Low-to-High and High-to-Low clock transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### Logic Table

Inputs				Outputs	
R	s	CE	D	С	Q
1	X	Х	Х	-	0
1	Х	Х	Х	-	0
0	1	Х	Х	-	1
0	1	X	Х	-	1
0	0	0	Х	X	No Change
0	0	1	1	-	1
0	0	1	0	-	0
0	0	1	1	-	1
0	0	1	0	-	0

# **Design Entry Method**

This design element is only for use in schematics.

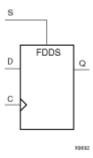


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDS**

#### Macro: Dual Edge Triggered D Flip-Flop with Synchronous Set



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDS is a single dual edge triggered D-type flip-flop with data (D) and synchronous set (S) inputs and data output (Q). The synchronous set input, when High, sets the Q output High on the Low-to-High or High-to-Low clock (C) transition. The data on the D input is loaded into the flip-flop when S is Low during the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs			Outputs
S	D	С	Q
1	X	-	1
1	X	-	1
0	1	-	1
0	0	-	0
0	1	-	1
0	0	-	0

# **Design Entry Method**

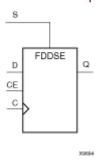
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDSE**

#### Macro: D Flip-Flop with Clock Enable and Synchronous Set



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDSE is a single dual edge triggered D-type flip-flop with data (D), clock enable (CE), and synchronous set (S) inputs and data output (Q). The synchronous set (S) input, when High, overrides the clock enable (CE) input and sets the Q output High during the Low-to-High or High-to-Low clock (C) transition. The data on the D input is loaded into the flip-flop when S is Low and CE is High during the Low-to-High and High-to-Low clock (C) transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### Logic Table

Inputs	Outputs			
S	CE	D	С	Q
1	Х	X	-	1
1	Х	Х	-	1
0	0	Χ	Χ	No Change
0	1	1	-	1
0	1	0	-	0
0	1	1	-	1
0	1	0	-	0

# **Design Entry Method**

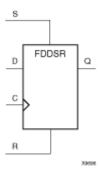
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDSR**

Macro: Dual Edge Triggered D Flip-Flop with Synchronous Set and Reset



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDSR is a single dual edge triggered D-type flip-flop with data (D), synchronous reset (R) and synchronous set (S) inputs and data output (Q). When the set (S) input is High, it overrides all other inputs and sets the Q output High during the Low-to-High or High-to-Low clock transition. (Set has precedence over Reset.) When reset (R) is High and S is Low, the flip-flop is reset, output Low, on the Low-to-High or High-to-Low clock transition. Data on the D input is loaded into the flip-flop when S and R are Low on the Low-to-High and High-to-Low clock transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs				Outputs
S	R	D	С	Q
1	X	X	-	1
1	Х	Х	-	1
0	1	Х	-	0
0	1	Х	-	0
0	0	1	-	1
0	0	0	-	0
0	0	1	-	1
0	0	0	-	0

# **Design Entry Method**

This design element is only for use in schematics.

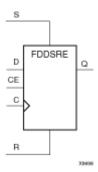


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDDSRE**

Macro: Dual Edge Triggered D Flip-Flop with Synchronous Set and Reset and Clock Enable



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

FDDSRE is a single dual edge triggered D-type flip-flop with synchronous set (S), synchronous reset (R), and clock enable (CE) inputs and data output (Q). When synchronous set (S) is High, it overrides all other inputs and sets the Q output High during the Low-to-High or High-to-Low clock transition. (Set has precedence over Reset.) When synchronous reset (R) is High and S is Low, output Q is reset Low during the Low-to-High or High-to-Low clock transition. Data is loaded into the flip-flop when S and R are Low and CE is High during the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs	Inputs				
S	R	CE	D	С	Q
1	X	X	X	-	1
1	X	X	Х	-	1
0	1	Х	Х	-	0
0	1	Х	Х	-	0
0	0	0	X	Х	No Change
0	0	1	1	-	1
0	0	1	0	-	0
0	0	1	1	-	1
0	0	1	0	-	0

# **Design Entry Method**

This design element is only for use in schematics.

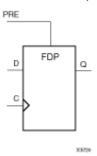


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDP**

### Macro: D Flip-Flop with Asynchronous Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with data (D) and asynchronous preset (PRE) inputs and data output (Q). The asynchronous PRE, when High, overrides all other inputs and presets the (Q) output High. The data on the (D) input is loaded into the flip-flop when PRE is Low on the Low-to-High clock (C) transition.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs	Outputs		
PRE	С	D	Q
1	X	X	1
0	-	D	D
0	-	0	0

# **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration

### For More Information

• See the appropriate CPLD User Guide.

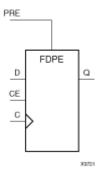


• See the appropriate CPLD Data Sheets.



# **FDPE**

### Primitive: D Flip-Flop with Clock Enable and Asynchronous Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a single D-type flip-flop with data (D), clock enable (CE), and asynchronous preset (PRE) inputs and data output (Q). The asynchronous PRE, when High, overrides all other inputs and sets the (Q) output High. Data on the (D) input is loaded into the flip-flop when PRE is Low and CE is High on the Low-to-High clock (C) transition. When CE is Low, the clock transitions are ignored.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs			
PRE	CE	D	С	Q
1	Χ	X	X	1
0	0	X	X	No Change
0	1	D	-	D

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration

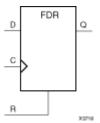


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDR**

#### Macro: D Flip-Flop with Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with data (D) and synchronous reset (R) inputs and data output (Q). The synchronous reset (R) input, when High, overrides all other inputs and resets the (Q) output Low on the Low-to-High clock (C) transition. The data on the (D) input is loaded into the flip-flop when R is Low during the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs			Outputs
R	D	С	Q
1	X	-	0
0	D	-	D

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

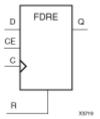
Attribute	Туре	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDRE**

#### Macro: D Flip-Flop with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single D-type flip-flop with data (D), clock enable (CE), and synchronous reset (R) inputs and data output (Q). The synchronous reset (R) input, when High, overrides all other inputs and resets the (Q) output Low on the Low-to-High clock (C) transition. The data on the (D) input is loaded into the flip-flop when R is Low and CE is High during the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### Logic Table

Inputs	Outputs			
R	CE	D	С	Q
1	X	X	-	0
0	0	X	X	No Change
0	1	D	-	D

### **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

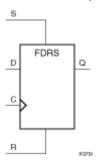
Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDRS**

### Macro: D Flip-Flop with Synchronous Reset and Set



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

FDRS is a single D-type flip-flop with data (D), synchronous set (S), and synchronous reset (R) inputs and data output (Q). The synchronous reset (R) input, when High, overrides all other inputs and resets the (Q) output Low during the Low-to-High clock (C) transition. (Reset has precedence over Set.) When S is High and R is Low, the flip-flop is set, output High, during the Low-to-High clock transition. When R and S are Low, data on the (D) input is loaded into the flip-flop during the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs	Outputs			
R	S	D	С	Q
1	X	Χ	-	0
0	1	X	-	1
0	0	D	-	D

## **Design Entry Method**

This design element is only for use in schematics.

### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration.

### For More Information

• See the appropriate CPLD User Guide.

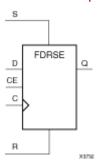


• See the appropriate CPLD Data Sheets.



# **FDRSE**

### Macro: D Flip-Flop with Synchronous Reset and Set and Clock Enable



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

FDRSE is a single D-type flip-flop with synchronous reset (R), synchronous set (S), clock enable (CE) inputs. The reset (R) input, when High, overrides all other inputs and resets the Q output Low during the Low-to-High clock transition. (Reset has precedence over Set.) When the set (S) input is High and R is Low, the flip-flop is set, output High, during the Low-to-High clock (C) transition. Data on the D input is loaded into the flip-flop when R and S are Low and CE is High during the Low-to-High clock transition.

Upon power-up, the initial value of this component is specified by the INIT attribute. If a subsequent GSR (Global Set/Reset) is asserted, the flop is asynchronously set to the INIT value.

# **Logic Table**

Inputs	Inputs				
R	S	CE	D	С	Q
1	X	X	X	-	0
0	1	X	X	-	1
0	0	0	X	X	No Change
0	0	1	1	-	1
0	0	1	0	-	0

# **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.



#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration and on GSR.

# **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- FDRSE: Single Data Rate D Flip-Flop with Synchronous Clear, Set and
          Clock Enable (posedge clk). All families.
-- Xilinx HDL Libraries Guide, version 10.1.1
FDRSE_inst : FDRSE
generic map (
INIT => '0') -- Initial value of register ('0' or '1')
port map (
Q => Q,
            -- Data output
C => C,
            -- Clock input
            -- Clock enable input
CE => CE,
D => D,
            -- Data input
R => R,
            -- Synchronous reset input
            -- Synchronous set input
S => S
);
-- End of FDRSE_inst instantiation
```

# **Verilog Instantiation Template**

```
// FDRSE: Single Data Rate D Flip-Flop with Synchronous Clear, Set and
         Clock Enable (posedge clk).
         All families.
// Xilinx HDL Libraries Guide, version 10.1.1
FDRSE #(
.INIT(1'b0) // Initial value of register (1'b0 or 1'b1)
) FDRSE_inst (
           // Data output
.Q(Q),
.C(C),
           // Clock input
         // Clock enable input
.CE(CE),
.D(D),
           // Data input
.R(R),
           // Synchronous reset input
.S(S)
           // Synchronous set input
);
// End of FDRSE inst instantiation
```

#### **For More Information**

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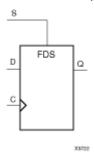
- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.

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# **FDS**

#### Macro: D Flip-Flop with Synchronous Set



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

FDS is a single D-type flip-flop with data (D) and synchronous set (S) inputs and data output (Q). The synchronous set input, when High, sets the Q output High on the Low-to-High clock (C) transition. The data on the D input is loaded into the flip-flop when S is Low during the Low-to-High clock (C) transition.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs	Outputs		
S	D	С	Q
1	X	-	1
0	D	-	D

# **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

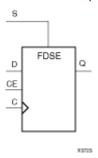
Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDSE**

#### Macro: D Flip-Flop with Clock Enable and Synchronous Set



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

FDSE is a single D-type flip-flop with data (D), clock enable (CE), and synchronous set (S) inputs and data output (Q). The synchronous set (S) input, when High, overrides the clock enable (CE) input and sets the Q output High during the Low-to-High clock (C) transition. The data on the D input is loaded into the flip-flop when S is Low and CE is High during the Low-to-High clock (C) transition.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
S	CE	D	С	Q
1	X	X	-	1
0	0	Χ	X	No Change
0	1	D	-	D

# **Design Entry Method**

This design element is only for use in schematics.

### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	1-Bit Binary	0 or 1	0	Sets the initial value of Q output after configuration.

### For More Information

• See the appropriate CPLD User Guide.



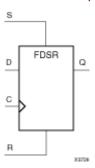
365

• See the appropriate CPLD Data Sheets.



# **FDSR**

#### Macro: D Flip-Flop with Synchronous Set and Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

FDSR is a single D-type flip-flop with data (D), synchronous reset (R) and synchronous set (S) inputs and data output (Q). When the set (S) input is High, it overrides all other inputs and sets the Q output High during the Low-to-High clock transition. (Set has precedence over Reset.) When reset (R) is High and S is Low, the flip-flop is reset, output Low, on the Low-to-High clock transition. Data on the D input is loaded into the flip-flop when S and R are Low on the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Port Descriptions**

Inputs				Outputs
S	R	D	С	Q
1	X	X	-	1
0	1	X	-	0
0	0	1	-	1
0	0	0	-	0

# **Design Entry Method**

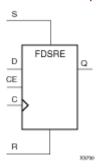
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FDSRE**

#### Macro: D Flip-Flop with Synchronous Set and Reset and Clock Enable



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

FDSRE is a single D-type flip-flop with synchronous set (S), synchronous reset (R), and clock enable (CE) inputs and data output (Q). When synchronous set (S) is High, it overrides all other inputs and sets the Q output High during the Low-to-High clock transition. (Set has precedence over Reset.) When synchronous reset (R) is High and S is Low, output Q is reset Low during the Low-to-High clock transition. Data is loaded into the flip-flop when S and R are Low and CE is High during the Low-to-high clock transition. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Port Descriptions**

Inputs	Outputs				
S	R	CE	D	С	Q
1	X	X	X	-	1
0	1	X	Х	-	0
0	0	0	Х	X	No Change
0	0	1	1	-	1
0	0	1	0	-	0

## **Design Entry Method**

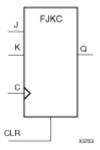
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKC**

#### Macro: J-K Flip-Flop with Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single J-K-type flip-flop with J, K, and asynchronous clear (CLR) inputs and data output (Q). The asynchronous clear (CLR) input, when High, overrides all other inputs and resets the Q output Low. When CLR is Low, the output responds to the state of the J and K inputs, as shown in the following logic table, during the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
CLR	J	K	С	Q
1	Χ	Χ	Χ	0
0	0	0	-	No Change
0	0	1	-	0
0	1	0	-	1
0	1	1	-	Toggle

### **Design Entry Method**

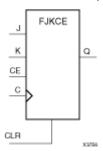
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKCE**

#### Macro: J-K Flip-Flop with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single J-K-type flip-flop with J, K, clock enable (CE), and asynchronous clear (CLR) inputs and data output (Q). The asynchronous clear (CLR), when High, overrides all other inputs and resets the Q output Low. When CLR is Low and CE is High, Q responds to the state of the J and K inputs, as shown in the following logic table, during the Low-to-High clock transition. When CE is Low, the clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Outputs				
CLR	CE	J	K	С	Q
1	X	X	X	Χ	0
0	0	X	X	X	No Change
0	1	0	0	X	No Change
0	1	0	1	-	0
0	1	1	0	-	1
0	1	1	1	-	Toggle

# **Design Entry Method**

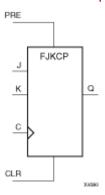
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKCP**

#### Macro: J-K Flip-Flop with Asynchronous Clear and Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a single J-K-type flip-flop with J, K, asynchronous clear (CLR), and asynchronous preset (PRE) inputs and data output (Q). When the asynchronous clear (CLR) is High, all other inputs are ignored and Q is reset 0. The asynchronous preset (PRE), when High, and CLR set to Low overrides all other inputs and sets the Q output High. When CLR and PRE are Low, Q responds to the state of the J and K inputs during the Low-to-High clock transition, as shown in the following logic table.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
CLR	PRE	J	K	С	Q
1	X	Χ	X	X	0
0	1	X	X	X	1
0	0	0	0	X	No Change
0	0	0	1	1	0
0	0	1	0	1	1
0	0	1	1		Toggle

### **Design Entry Method**

This design element is only for use in schematics.

### For More Information

• See the appropriate CPLD User Guide.

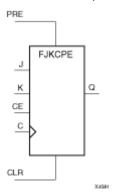


• See the appropriate CPLD Data Sheets.



# **FJKCPE**

Macro: J-K Flip-Flop with Asynchronous Clear and Preset and Clock Enable



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single J-K-type flip-flop with J, K, asynchronous clear (CLR), asynchronous preset (PRE), and clock enable (CE) inputs and data output (Q). When the asynchronous clear (CLR) is High, all other inputs are ignored and Q is reset 0. The asynchronous preset (PRE), when High, and CLR set to Low overrides all other inputs and sets the Q output High. When CLR and PRE are Low and CE is High, Q responds to the state of the J and K inputs, as shown in the following logic table, during the Low-to-High clock transition. Clock transitions are ignored when CE is Low.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

#### **Logic Table**

Inputs	Inputs						
CLR	PRE	CE	J	K	С	Q	
1	X	X	X	Х	X	0	
0	1	X	X	X	X	1	
0	0	0	0	Х	X	No Change	
0	0	1	0	0	X	No Change	
0	0	1	0	1	1	0	
0	0	1	1	0	1	1	
0	0	1	1	1	1	Toggle	

## **Design Entry Method**

This design element is only for use in schematics.

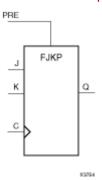


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKP**

#### Macro: J-K Flip-Flop with Asynchronous Preset



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single J-K-type flip-flop with J, K, and asynchronous preset (PRE) inputs and data output (Q). The asynchronous preset (PRE) input, when High, overrides all other inputs and sets the (Q) output High. When (PRE) is Low, the (Q) output responds to the state of the J and K inputs, as shown in the following logic table, during the Low-to-High clock transition.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
PRE	J	K	С	Q
1	Χ	X	X	1
0	0	0	X	No Change
0	0	1	-	0
0	1	0	-	1
0	1	1	-	Toggle

# **Design Entry Method**

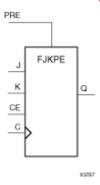
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKPE**

#### Macro: J-K Flip-Flop with Clock Enable and Asynchronous Preset



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single J-K-type flip-flop with J, K, clock enable (CE), and asynchronous preset (PRE) inputs and data output (Q). The asynchronous preset (PRE), when High, overrides all other inputs and sets the (Q) output High. When (PRE) is Low and (CE) is High, the (Q) output responds to the state of the J and K inputs, as shown in the logic table, during the Low-to-High clock (C) transition. When (CE) is Low, clock transitions are ignored.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs					
PRE	CE	J	K	С	Q	
1	X	X	X	X	1	
0	0	X	X	X	No Change	
0	1	0	0	X	No Change	
0	1	0	1	-	0	
0	1	1	0	-	1	
0	1	1	1	-	Toggle	

### **Design Entry Method**

This design element is only for use in schematics.

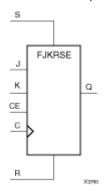


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKRSE**

Macro: J-K Flip-Flop with Clock Enable and Synchronous Reset and Set



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a single J-K-type flip-flop with J, K, synchronous reset (R), synchronous set (S), and clock enable (CE) inputs and data output (Q). When synchronous reset (R) is High during the Low-to-High clock (C) transition, all other inputs are ignored and output (Q) is reset Low. When synchronous set (S) is High and (R) is Low, output (Q) is set High. When (R) and (S) are Low and (CE) is High, output (Q) responds to the state of the J and K inputs, according to the following logic table, during the Low-to-High clock (C) transition. When (CE) is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs					
R	S	CE	J	K	С	Q
1	X	X	X	X	-	0
0	1	Χ	Χ	X	-	1
0	0	0	X	X	X	No Change
0	0	1	0	0	X	No Change
0	0	1	0	1	-	0
0	0	1	1	1	-	Toggle
0	0	1	1	0	-	1

# **Design Entry Method**

This design element is only for use in schematics.

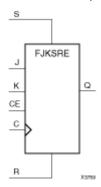


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FJKSRE**

#### Macro: J-K Flip-Flop with Clock Enable and Synchronous Set and Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a single J-K-type flip-flop with J, K, synchronous set (S), synchronous reset (R), and clock enable (CE) inputs and data output (Q). When synchronous set (S) is High during the Low-to-High clock (C) transition, all other inputs are ignored and output (Q) is set High. When synchronous reset (R) is High and (S) is Low, output (Q) is reset Low. When (S) and (R) are Low and (CE) is High, output (Q) responds to the state of the J and K inputs, as shown in the following logic table, during the Low-to-High clock (C) transition. When (CE) is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Inputs					
S	R	CE	J	K	С	Q
1	X	X	X	Х	-	1
0	1	X	Х	Х	-	0
0	0	0	Х	Х	Х	No Change
0	0	1	0	0	X	No Change
0	0	1	0	1	-	0
0	0	1	1	0	-	1
0	0	1	1	1	-	Toggle

# **Design Entry Method**

This design element is only for use in schematics.

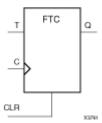


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTC**

#### Primitive: Toggle Flip-Flop with Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a synchronous, resettable toggle flip-flop. The asynchronous clear (CLR) input, when High, overrides all other inputs and resets the data output (Q) Low. The (Q) output toggles, or changes state, when the toggle enable (T) input is High and (CLR) is Low during the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Outputs		
CLR	T	С	Q
1	X	X	0
0	0	X	No Change
0	1	-	Toggle

# **Design Entry Method**

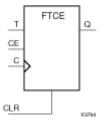
You can instantiate this element when targeting a CPLD, but not when you are targeting an FPGA.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTCE**

Macro: Toggle Flip-Flop with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle flip-flop with toggle and clock enable and asynchronous clear. When the asynchronous clear (CLR) input is High, all other inputs are ignored and the data output (Q) is reset Low. When CLR is Low and toggle enable (T) and clock enable (CE) are High, Q output toggles, or changes state, during the Low-to-High clock (C) transition. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
CLR	СЕ	T	С	Q
1	Х	Χ	X	0
0	0	X	X	No Change
0	1	0	X	No Change
0	1	1	-	Toggle

# **Design Entry Method**

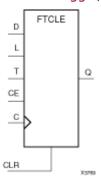
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FTCLE**

Macro: Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle/loadable flip-flop with toggle and clock enable and asynchronous clear. When the asynchronous clear input (CLR) is High, all other inputs are ignored and output Q is reset Low. When load enable input (L) is High and CLR is Low, clock enable (CE) is overridden and the data on data input (D) is loaded into the flip-flop during the Low-to-High clock (C) transition. When toggle enable (T) and CE are High and L and CLR are Low, output Q toggles, or changes state, during the Low- to-High clock transition. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs						Outputs
CLR	L	CE	Т	D	С	Q
1	X	X	X	X	X	0
0	1	X	X	D	-	D
0	0	0	X	X	X	No Change
0	0	1	0	X	X	No Change
0	0	1	1	Х	-	Toggle

### **Design Entry Method**

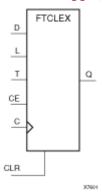
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTCLEX**

Macro: Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle/loadable flip-flop with toggle and clock enable and asynchronous clear. When the asynchronous clear input (CLR) is High, all other inputs are ignored and output Q is reset Low. When load enable input (L) is High, CLR is Low, and CE is High, the data on data input (D) is loaded into the flip-flop during the Low-to-High clock (C) transition. When toggle enable (T) and CE are High and L and CLR are Low, output Q toggles, or changes state, during the Low- to-High clock transition. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs						Outputs
CLR	L	CE	Т	D	С	Q
1	X	X	X	Х	X	0
0	1	Х	X	D	-	D
0	0	0	X	X	Х	No Change
0	0	1	0	X	X	No Change
0	0	1	1	X	-	Toggle

## **Design Entry Method**

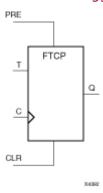
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FTCP**

### Primitive: Toggle Flip-Flop with Asynchronous Clear and Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle flip-flop with toggle enable and asynchronous clear and preset. When the asynchronous clear (CLR) input is High, all other inputs are ignored and the output (Q) is reset Low. When the asynchronous preset (PRE) is High and CLR is Low, all other inputs are ignored and Q is set High. When the toggle enable input (T) is High and CLR and PRE are Low, output Q toggles, or changes state, during the Low-to-High clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
CLR	PRE	T	С	Q
1	Χ	X	Χ	0
0	1	Χ	X	1
0	0	0	Χ	No Change
0	0	1	-	Toggle

### **Design Entry Method**

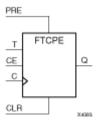
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTCPE**

Macro: Toggle Flip-Flop with Clock Enable and Asynchronous Clear and Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a toggle flip-flop with toggle and clock enable and asynchronous clear and preset. When the asynchronous clear (CLR) input is High, all other inputs are ignored and the output (Q) is reset Low. When the asynchronous preset (PRE) is High and CLR is Low, all other inputs are ignored and Q is set High. When the toggle enable input (T) and the clock enable input (CE) are High and CLR and PRE are Low, output Q toggles, or changes state, during the Low-to-High clock (C) transition. Clock transitions are ignored when CE is Low.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs				
CLR	PRE	CE	T	С	Q
1	X	Χ	Χ	Χ	0
0	1	Χ	Χ	Χ	1
0	0	0	Χ	Χ	No Change
0	0	1	0	Χ	No Change
0	0	1	1	-	Toggle

### **Design Entry Method**

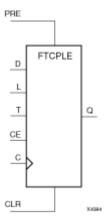
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FTCPLE**

Macro: Loadable Toggle Flip-Flop with Clock Enable and Asynchronous Clear and Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a loadable toggle flip-flop with toggle and clock enable and asynchronous clear and preset. When the asynchronous clear (CLR) input is High, all other inputs are ignored and the output (Q) is reset Low. When the asynchronous preset (PRE) is High and CLR is Low, all other inputs are ignored and Q is set High. When the load input (L) is High, the clock enable input (CE) is overridden and data on data input (D) is loaded into the flip-flop during the Low-to-High clock transition. When the toggle enable input (T) and the clock enable input (CE) are High and CLR, PRE, and L are Low, output Q toggles, or changes state, during the Low-to-High clock (C) transition. Clock transitions are ignored when CE is Low.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Inputs						
CLR	PRE	L	CE	T	С	D	Q
1	X	Χ	Χ	Χ	Χ	X	0
0	1	Χ	Χ	Χ	Χ	Χ	1
0	0	1	X	X	-	0	0
0	0	1	Х	X	-	1	1
0	0	0	0	X	X	X	No Change
0	0	0	1	0	X	X	No Change
0	0	0	1	1	-	X	Toggle



# **Design Entry Method**

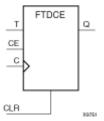
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTDCE**

Macro: Dual-Edge Triggered Toggle Flip-Flop with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered toggle flip-flop with toggle and clock enable and asynchronous clear. When the asynchronous clear (CLR) input is High, all other inputs are ignored and the data output (Q) is reset Low. When CLR is Low and toggle enable (T) and clock enable (CE) are High, Q output toggles, or changes state, during the Low-to-High and High-to-Low clock (C) transitions. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs			
CLR	CE	Т	С	Q
1	X	Χ	Х	0
0	0	X	Х	No Change
0	1	0	Х	No Change
0	1	1	-	Toggle
0	1	1	-	Toggle

## **Design Entry Method**

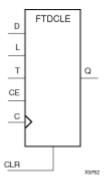
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FTDCLE**

Macro: Dual-Edge Triggered Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered toggle/loadable flip-flop with toggle and clock enable and asynchronous clear. When the asynchronous clear input (CLR) is High, all other inputs are ignored and output Q is reset Low. When load enable input (L) is High and CLR is Low, clock enable (CE) is overridden and the data on data input (D) is loaded into the flip-flop during the Low-to-High and High-to-Low clock (C) transitions. When toggle enable (T) and CE are High and L and CLR are Low, output Q toggles, or changes state, during the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### Logic Table

Inputs						Outputs
CLR	L	CE	Т	D	С	Q
1	Χ	X	X	X	X	0
0	1	X	X	1	-	1
0	1	X	Х	1	-	1
0	1	X	X	0	-	0
0	1	X	X	0	-	0
0	0	0	X	X	Х	No Change
0	0	1	0	X	Х	No Change
0	0	1	1	X	-	Toggle
0	0	1	1	Х	-	Toggle

# **Design Entry Method**

This design element is only for use in schematics.

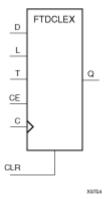


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **FTDCLEX**

Macro: Dual-Edge Triggered Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element is a dual edge triggered toggle/loadable flip-flop with toggle and clock enable and asynchronous clear. When the asynchronous clear input (CLR) is High, all other inputs are ignored and output Q is reset Low. When load enable input (L) is High, CLR is Low, and CE is High, the data on data input (D) is loaded into the flip-flop during the Low-to-High and High-to-Low clock (C) transitions. When toggle enable (T) and CE are High and L and CLR are Low, output Q toggles, or changes state, during the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs						
CLR	L	CE	T	D	С	Q	
1	X	Х	X	X	X	0	
0	1	1	Х	1	-	1	
0	1	1	X	1	-	1	
0	1	1	X	0	-	0	
0	1	1	X	0	-	0	
0	0	0	Х	Х	Х	No Change	
0	0	1	0	X	X	No Change	
0	0	1	1	X	-	Toggle	
0	0	1	1	X	-	Toggle	

# **Design Entry Method**

This design element is only for use in schematics.

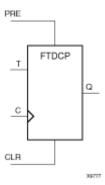


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTDCP**

Primitive: Dual-Edge Triggered Toggle Flip-Flop with Asynchronous Clear and Preset



### **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a toggle flip-flop with toggle enable and asynchronous clear and preset. When the asynchronous clear (CLR) input is High, all other inputs are ignored and the output (Q) is reset Low. When the asynchronous preset (PRE) is High and CLR is Low, all other inputs are ignored and Q is set High. When the toggle enable input (T) is High and CLR and PRE are Low, output Q toggles, or changes state, during the Low-to-High and High-to-Low clock (C) transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs			
CLR	PRE	Т	С	Q
1	X	Χ	Χ	0
0	1	Χ	Χ	1
0	0	0	Χ	No Change
0	0	1	-	Toggle
0	0	1	-	Toggle

### **Design Entry Method**

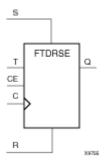
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTDRSE**

Macro: Dual-Edge Triggered Toggle Flip-Flop with Synchronous Reset, Set, and Clock Enable



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered toggle flip-flop with toggle and clock enable and synchronous reset and set. When the synchronous reset input (R) is High, it overrides all other inputs and the data output (Q) is reset Low. When the synchronous set input (S) is High and R is Low, clock enable input (CE) is overridden and output Q is set High. (Reset has precedence over Set.) When toggle enable input (T) and CE are High and R and S are Low, output Q toggles, or changes state, during the Low-to-High and High-to-Low clock transitions.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs						
R	S	CE	Т	С	Q		
1	X	X	X	-	0		
1	Х	Х	Х	-	0		
0	1	Х	Х	-	1		
0	1	Х	Х	-	1		
0	0	0	X	X	No Change		
0	0	1	0	X	No Change		
0	0	1	1	-	Toggle		
0	0	1	1	-	Toggle		

### **Design Entry Method**

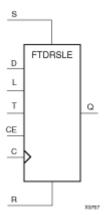
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTDRSLE**

Macro: Dual-Edge Triggered Toggle Flip-Flop with Clock Enable and Synchronous Reset and Set



## **Supported Architectures**

This design element is supported in the following architectures only: CoolRunner-II

#### Introduction

This design element is a dual edge triggered toggle/loadable flip-flop with toggle and clock enable and synchronous reset and set. The synchronous reset input (R), when High, overrides all other inputs and resets the data output (Q) Low. (Reset has precedence over Set.) When R is Low and synchronous set input (S) is High, the clock enable input (CE) is overridden and output Q is set High. When R and S are Low and load enable input (L) is High, CE is overridden and data on data input (D) is loaded into the flip-flop during the Low-to-High and High-to-Low clock transitions. When R, S, and L are Low and CE is High, output Q toggles, or changes state, during the Low-to-High and High-to-Low clock transitions. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs	Inputs						
R	S	L	CE	T	D	С	Q
1	0	X	X	X	X	-	0
1	0	X	X	X	X	-	0
0	1	X	X	X	X	-	1
0	1	X	X	X	X	-	1
0	0	1	X	X	1	-	1
0	0	1	X	X	1	-	1
0	0	1	X	X	0	-	0
0	0	1	X	X	0	-	0
0	0	0	0	X	Χ	Χ	No Change
0	0	0	1	0	X	X	No Change



Inputs							Outputs
R	S	L	CE	T	D	С	Q
0	0	0	1	1	X	-	Toggle
0	0	0	1	1	X	-	Toggle

# **Design Entry Method**

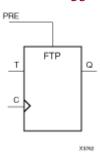
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTP**

#### Macro: Toggle Flip-Flop with Asynchronous Preset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a toggle flip-flop with toggle enable and asynchronous preset. When the asynchronous preset (PRE) input is High, all other inputs are ignored and output (Q) is set High. When toggle-enable input (T) is High and (PRE) is Low, output (Q) toggles, or changes state, during the Low-to-High clock (C) transition.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs	Outputs		
PRE	Т	С	Q
1	X	X	1
0	0	X	No Change
0	1	-	Toggle

# **Design Entry Method**

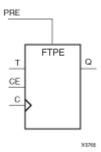
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTPE**

#### Macro: Toggle Flip-Flop with Clock Enable and Asynchronous Preset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle flip-flop with toggle and clock enable and asynchronous preset. When the asynchronous preset (PRE) input is High, all other inputs are ignored and output (Q) is set High. When the toggle enable input (T) is High, clock enable (CE) is High, and (PRE) is Low, output (Q) toggles, or changes state, during the Low-to-High clock transition. When (CE) is Low, clock transitions are ignored.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs	Outputs			
PRE	CE	T	С	Q
1	Χ	Χ	Χ	1
0	0	Χ	Χ	No Change
0	1	0	X	No Change
0	1	1	-	Toggle

### **Design Entry Method**

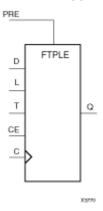
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTPLE**

Macro: Toggle/Loadable Flip-Flop with Clock Enable and Asynchronous Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle/loadable flip-flop with toggle and clock enable and asynchronous preset. When the asynchronous preset input (PRE) is High, all other inputs are ignored and output (Q) is set High. When the load enable input (L) is High and (PRE) is Low, the clock enable (CE) is overridden and the data (D) is loaded into the flip-flop during the Low-to-High clock transition. When L and PRE are Low and toggle-enable input (T) and (CE) are High, output (Q) toggles, or changes state, during the Low-to-High clock transition. When (CE) is Low, clock transitions are ignored.

For CPLD devices, this flip-flop is asynchronously cleared, output Low, when power is applied. You can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs						
PRE	L	CE	Т	D	С	Q	
1	X	X	X	Х	X	1	
0	1	X	X	D	?	D	
0	0	0	Х	Х	X	No Change	
0	0	1	0	Х	X	No Change	
0	0	1	1	Х	?	Toggle	

### **Design Entry Method**

This design element is only for use in schematics.

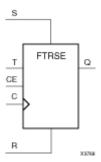


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTRSE**

Macro: Toggle Flip-Flop with Clock Enable and Synchronous Reset and Set



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle flip-flop with toggle and clock enable and synchronous reset and set. When the synchronous reset input (R) is High, it overrides all other inputs and the data output (Q) is reset Low. When the synchronous set input (S) is High and (R) is Low, clock enable input (CE) is overridden and output (Q) is set High. (Reset has precedence over Set.) When toggle enable input (T) and (CE) are High and (R) and (S) are Low, output (Q) toggles, or changes state, during the Low-to-High clock transition.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Outputs				
R	S	CE	T	С	Q
1	Х	X	Х	-	0
0	1	X	Х	-	1
0	0	0	Χ	Χ	No Change
0	0	1	0	X	No Change
0	0	1	1	-	Toggle

# **Design Entry Method**

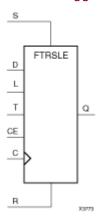
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTRSLE**

Macro: Toggle/Loadable Flip-Flop with Clock Enable and Synchronous Reset and Set



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle/loadable flip-flop with toggle and clock enable and synchronous reset and set. The synchronous reset input (R), when High, overrides all other inputs and resets the data output (Q) Low. (Reset has precedence over Set.) When R is Low and synchronous set input (S) is High, the clock enable input (CE) is overridden and output Q is set High. When R and S are Low and load enable input (L) is High, CE is overridden and data on data input (D) is loaded into the flip-flop during the Low-to-High clock transition. When R, S, and L are Low, CE is High and T is High, output Q toggles, or changes state, during the Low-to-High clock transition. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs						Outputs	
R	S	L	CE	T	D	С	Q
1	0	Х	Х	Х	Х	-	0
0	1	Х	Х	Х	Х	-	1
0	0	1	Х	Х	1	-	1
0	0	1	Х	Х	0	-	0
0	0	0	0	Х	Х	Х	No Change
0	0	0	1	0	Х	Х	No Change
0	0	0	1	1	X	-	Toggle



# **Design Entry Method**

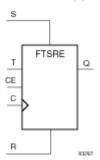
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **FTSRE**

Macro: Toggle Flip-Flop with Clock Enable and Synchronous Set and Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle flip-flop with toggle and clock enable and synchronous set and reset. The synchronous set input, when High, overrides all other inputs and sets data output (Q) High. (Set has precedence over Reset.) When synchronous reset input (R) is High and S is Low, clock enable input (CE) is overridden and output Q is reset Low. When toggle enable input (T) and CE are High and S and R are Low, output Q toggles, or changes state, during the Low-to-High clock transition. When CE is Low, clock transitions are ignored.

This flip-flop is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs					Outputs
S	R	CE	T	С	Q
1	Х	X	Х	-	1
0	1	X	Х	-	0
0	0	0	Х	X	No Change
0	0	1	0	Χ	No Change
0	0	1	1	-	Toggle

### **Design Entry Method**

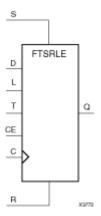
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **FTSRLE**

Macro: Toggle/Loadable Flip-Flop with Clock Enable and Synchronous Set and Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a toggle/loadable flip-flop with toggle and clock enable and synchronous set and reset. The synchronous set input (S), when High, overrides all other inputs and sets data output (Q) High. (Set has precedence over Reset.) When synchronous reset (R) is High and (S) is Low, clock enable input (CE) is overridden and output (Q) is reset Low. When load enable input (L) is High and S and R are Low, CE is overridden and data on data input (D) is loaded into the flip-flop during the Low-to-High clock transition. When the toggle enable input (T) and (CE) are High and (S), (R), and (L) are Low, output (Q) toggles, or changes state, during the Low-to-High clock transition. When (CE) is Low, clock transitions are ignored.

For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs						Outputs	
S	R	L	CE	Т	D	С	Q
1	Х	Х	Х	Х	Х	-	1
0	1	Х	Х	Х	Х	-	0
0	0	1	Х	Х	1	-	1
0	0	1	Х	Х	0	-	0
0	0	0	0	Х	Х	Х	No Change
0	0	0	1	0	Х	X	No Change
0	0	0	1	1	Х	-	Toggle



# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **GND**

Primitive: Ground-Connection Signal Tag



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

The GND signal tag, or parameter, forces a net or input function to a Low logic level. A net tied to GND cannot have any other source.

When the logic-trimming software or fitter encounters a net or input function tied to GND, it removes any logic that is disabled by the GND signal. The GND signal is only implemented when the disabled logic cannot be removed.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **IBUF**

#### Primitive: Input Buffer

IBUF O

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is automatically inserted (inferred) by the synthesis tool to any signal directly connected to a top-level input or in-out port of the design. You should generally let the synthesis tool infer this buffer. However, it can be instantiated into the design if required. In order to do so, connect the input port (I) directly to the associated top-level input or in-out port, and connect the output port (O) to the logic sourced by that port. Modify any necessary generic maps (VHDL) or named parameter value assignment (Verilog) in order to change the default behavior of the component.

# **Port Descriptions**

Name	Direction	Width	Function
О	Output	1-Bit	Buffer input
I	Input	1-Bit	Buffer output

# **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

In general, this element is inferred by the synthesis tool for any specified top-level input port to the design. It is generally not necessary to specify them in the source code however if desired, they be manually instantiated by either copying the instantiation code from the ISE Libaries Guide HDL Template and paste it into the top-level entity/module of your code. It is recommended to always put all I/O components on the top-level of the design to help facilitate hierarchical design methods. Connect the I port directly to the top-level input port of the design and the O port to the logic in which this input is to source. Specify the desired generic/defparam values in order to configure the proper behavior of the buffer.



#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
IOSTANDARD	String	See Note Below	DEFAULT	Sets the programmable I/O standard for the input.
IBUF_DELAY _VALUE	Binary	0 thru 12	0	Specifies the amount of additional delay to add to the non-registered path out of the IOB
IFD_DELAY _VALUE	Binary	AUTO, 0 thru 6	AUTO	Specifies the amount of additional delay to add to the registered path within the IOB

Note Consult the device user guide or databook for the allowed values and the default value.

### **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- IBUF: Single-ended Input Buffer
        All devices
-- Xilinx HDL Libraries Guide, version 10.1.1
IBUF_inst : IBUF
generic map (
IBUF_DELAY_VALUE => "0", -- Specify the amount of added input delay for buffer, "0"-"16" (Spartan-3E/3A only)
IFD_DELAY_VALUE => "AUTO", -- Specify the amount of added delay for input register, "AUTO", "0"-"8" (Spartan-3E/3A c
IOSTANDARD => "DEFAULT")
port map (
0 => 0,
            -- Buffer output
I => I
           -- Buffer input (connect directly to top-level port)
);
-- End of IBUF_inst instantiation
```

# **Verilog Instantiation Template**

```
// IBUF: Single-ended Input Buffer
// All devices
// Xilinx HDL Libraries Guide, version 10.1.1

IBUF #(
.IBUF_DELAY_VALUE("0"), // Specify the amount of added input delay for
// the buffer, "0"-"16" (Spartan-3E/3A only)
.IFD_DELAY_VALUE("AUTO"), // Specify the amount of added delay for input
// register, "AUTO", "0"-"8" (Spartan-3E/3A only)
.IOSTANDARD("DEFAULT") // Specify the input I/O standard
)IBUF_inst (
.O(0), // Buffer output
.I(I) // Buffer input (connect directly to top-level port)
);
// End of IBUF_inst instantiation
```

### For More Information

• See the appropriate CPLD User Guide.



• See the appropriate CPLD Data Sheets.



## **IBUF16**

Macro: 16-Bit Input Buffer

IBUF16



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

Input Buffers isolate the internal circuit from the signals coming into the chip. This design element is contained in input/output blocks (IOBs) and allows the specification of the particular I/O Standard to configure the I/O. In general, an this element should be used for all single-ended data input or bidirectional pins.

## **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

In general, this element is inferred by the synthesis tool for any specified top-level input port to the design. It is generally not necessary to specify them in the source code however if desired, they be manually instantiated by either copying the instantiation code from the ISE Libaries Guide HDL Template and paste it into the top-level entity/module of your code. It is recommended to always put all I/O components on the top-level of the design to help facilitate hierarchical design methods. Connect the I port directly to the top-level input port of the design and the O port to the logic in which this input is to source. Specify the desired generic/defparam values in order to configure the proper behavior of the buffer.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
IOSTANDARD	String	See Note Below	DEFAULT	Sets the programmable I/O standard for the input.
IBUF_DELAY _VALUE	Binary	0 thru 12	0	Specifies the amount of additional delay to add to the non-registered path out of the IOB
IFD_DELAY _VALUE	Binary	AUTO, 0 thru 6	AUTO	Specifies the amount of additional delay to add to the registered path within the IOB



Note Consult the device user guide or databook for the allowed values and the default value.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **IBUF4**

#### Macro: 4-Bit Input Buffer

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

Input Buffers isolate the internal circuit from the signals coming into the chip. This design element is contained in input/output blocks (IOBs) and allows the specification of the particular I/O Standard to configure the I/O. In general, an this element should be used for all single-ended data input or bidirectional pins.

# **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

In general, this element is inferred by the synthesis tool for any specified top-level input port to the design. It is generally not necessary to specify them in the source code however if desired, they be manually instantiated by either copying the instantiation code from the ISE Libaries Guide HDL Template and paste it into the top-level entity/module of your code. It is recommended to always put all I/O components on the top-level of the design to help facilitate hierarchical design methods. Connect the I port directly to the top-level input port of the design and the O port to the logic in which this input is to source. Specify the desired generic/defparam values in order to configure the proper behavior of the buffer.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
IOSTANDARD	String	See Note Below	DEFAULT	Sets the programmable I/O standard for the input.



Attribute	Type	Allowed Values	Default	Description
IBUF_DELAY _VALUE	Binary	0 thru 12	0	Specifies the amount of additional delay to add to the non-registered path out of the IOB
IFD_DELAY _VALUE	Binary	AUTO, 0 thru 6	AUTO	Specifies the amount of additional delay to add to the registered path within the IOB

Note Consult the device user guide or databook for the allowed values and the default value.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **IBUF8**

Macro: 8-Bit Input Buffer

IBUF8



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

Input Buffers isolate the internal circuit from the signals coming into the chip. This design element is contained in input/output blocks (IOBs) and allows the specification of the particular I/O Standard to configure the I/O. In general, an this element should be used for all single-ended data input or bidirectional pins.

## **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

In general, this element is inferred by the synthesis tool for any specified top-level input port to the design. It is generally not necessary to specify them in the source code however if desired, they be manually instantiated by either copying the instantiation code from the ISE Libaries Guide HDL Template and paste it into the top-level entity/module of your code. It is recommended to always put all I/O components on the top-level of the design to help facilitate hierarchical design methods. Connect the I port directly to the top-level input port of the design and the O port to the logic in which this input is to source. Specify the desired generic/defparam values in order to configure the proper behavior of the buffer.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
IOSTANDARD	String	See Note Below	DEFAULT	Sets the programmable I/O standard for the input.
IBUF_DELAY _VALUE	Binary	0 thru 12	0	Specifies the amount of additional delay to add to the non-registered path out of the IOB
IFD_DELAY _VALUE	Binary	AUTO, 0 thru 6	AUTO	Specifies the amount of additional delay to add to the registered path within the IOB



Note Consult the device user guide or databook for the allowed values and the default value.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



Primitive: Inverter

X10955

INV

### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a single inverter that identifies signal inversions in a schematic.

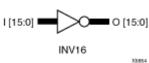
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 16 Inverters



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a multiple inverter that identifies signal inversions in a schematic.

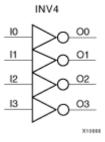
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: Four Inverters



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple inverter that identifies signal inversions in a schematic.

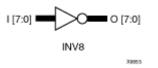
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: Eight Inverters



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple inverter that identifies signal inversions in a schematic.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **IOBUFE**

#### Primitive: Bi-Directional Buffer

ю

IOBUFE

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a bi-directional buffer that is a composite of the IBUF and OBUFE elements. The O output is X (unknown) when IO (input/output) is Z. You can also implement IOBUFEs as interconnections of their component elements.

### **Logic Table**

Inputs		Bidirectional	Outputs
E	I	IO	0
0	0	Z	X
0	1	Z	X
1	0	0	0
1	1	1	1

### **Design Entry Method**

This design element is only for use in schematics.

### **VHDL Instantiation Template**

-- End of IOBUFE\_inst instantiation

Unless they already exist, copy the following two statements and paste them before the entity declaration.

```
Library UNISIM;
use UNISIM.vcomponents.all;
-- IOBUFE: Bi-Directional Buffer
-- XC9500XL/CoolRunner-II/XPLA-3
-- Xilinx HDL Language Template, version 10.1

IOBUFE_inst : IOBUFE
port map (0 => user_O,
IO => user_IO,
I => user_I,
E => user_E);
```



# **Verilog Instantiation Template**

```
// IOBUFE: Bi-Directional Buffer
// XC9500XL/CoolRunner-II/XPLA-3
// Xilinx HDL Language Template, version 10.1

IOBUFE IOBUFE_inst (.O (user_O),
.IO (user_IO),
.I (user_I),
.E (user_E));

// End of IOBUFE_inst instantiation
```

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **KEEPER**

## Primitive: KEEPER Symbol



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner-II

#### Introduction

The design element is a weak keeper element that retains the value of the net connected to its bidirectional O pin. For example, if a logic 1 is being driven onto the net, KEEPER drives a weak/resistive 1 onto the net. If the net driver is then 3-stated, KEEPER continues to drive a weak/resistive 1 onto the net.

# **Port Descriptions**

Name	Direction	Width	Function
О	Output	1-Bit	Keeper output

# **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

This element can be connected to a net in the following locations on a top-level schematic file:

- A net connected to an input IO Marker
- A net connected to both an output IO Marker and 3-statable IO element, such as an OBUFT.

# **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- KEEPER: I/O Buffer Weak Keeper
-- All FPGA, CoolRunner-II
-- Xilinx HDL Libraries Guide, version 10.1.1

KEEPER_inst : KEEPER
port map (
O => O -- Keeper output (connect directly to top-level port)
);
```



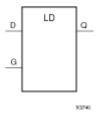
```
-- End of KEEPER_inst instantiation
```

# **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: Transparent Data Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

LD is a transparent data latch. The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is High. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains Low.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs		Outputs
G	D	Q
1	D	D
0	Х	No Change
-	D	D

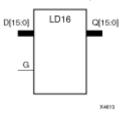
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: Multiple Transparent Data Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element has 16 transparent data latches with a common gate enable (G). The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is High. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains Low.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs		Outputs
G	D	Q
1	D	D
0	Х	No Change
-	Dn	Dn

# **Design Entry Method**

This design element is only for use in schematics.

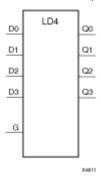
### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	Binary	Any 16-Bit Value	All zeros	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: Multiple Transparent Data Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

This design element has four transparent data latches with a common gate enable (G). The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is High. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains Low.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs		Outputs
G	D	Q
1	D	D
0	Х	No Change
-	Dn	Dn

## **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

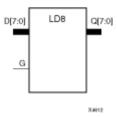
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	Any 4-Bit Value	All zeros	Sets the initial value of Q output after configuration



- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: Multiple Transparent Data Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

This design element has 8 transparent data latches with a common gate enable (G). The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is High. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains Low.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs		Outputs
G	D	Q
1	D	D
0	Х	No Change
-	Dn	Dn

## **Design Entry Method**

This design element is only for use in schematics.

### **Available Attributes**

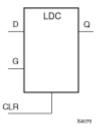
Attribute	Type	Allowed Values	Default	Description
INIT	Binary	Any 8-Bit Value	All zeros	Sets the initial value of Q output after configuration

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **LDC**

Macro: Transparent Data Latch with Asynchronous Clear



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a transparent data latch with asynchronous clear. When the asynchronous clear input (CLR) is High, it overrides the other inputs and resets the data (Q) output Low. (Q) reflects the data (D) input while the gate enable (G) input is High and (CLR) is Low. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains low.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs			Outputs
CLR	G	D	Q
1	X	X	0
0	1	D	D
0	0	X	No Change
0	?	D	D

## **Design Entry Method**

This design element is only for use in schematics.

### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	INTEGER	0 or 1	0	Sets the initial value of Q output after configuration

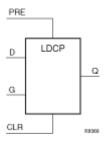


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **LDCP**

Primitive: Transparent Data Latch with Asynchronous Clear and Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

The design element is a transparent data latch with data (D), asynchronous clear (CLR) and preset (PRE) inputs. When (CLR) is High, it overrides the other inputs and resets the data (Q) output Low. When PRE is High and (CLR) is low, it presets the data (Q) output High. (Q) reflects the data (D) input while the gate (G) input is High and (CLR) and PRE are Low. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains Low.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs	Outputs			
CLR	PRE	G	D	Q
1	X	X	Χ	0
0	1	X	Χ	1
0	0	1	D	D
0	0	0	X	No Change
0	0	-	D	D

# **Design Entry Method**

This design element is only for use in schematics.

### **Available Attributes**

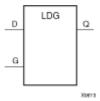


Attribute	Туре	Allowed Values	Default	Description
INIT	INTEGER	0 or 1	0	Specifies the initial value upon power-up or the assertion of GSR for the (Q) port.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: Transparent Datagate Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a transparent DataGate latch used for gating input signals to decrease power dissipation. The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is Low. The data on the D input during the Low-to-High gate transition is stored in the latch. The data on the Q output remains unchanged as long as G remains High.

The D input(s) of the LDG must be connected to a device input pad(s) and must have no other fan-outs (must not branch). The CPLD fitter maps the G input to the device's DataGate Enable control pin (DGE). There must be no more than one DataGate Enable signal in the design. The DataGate Enable signal may be driven either by a device input pin or any on-chip logic source. The DataGate Enable signal may be reused by other ordinary logic in the design.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs		Outputs
G	D	Q
0	0	0
0	1	1
1	Х	No Change
-	D	D

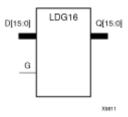
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 16-bit Transparent Datagate Latch



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element has 16 transparent DataGate latches with a common gate enable (G). These latches are used to gate input signals in order to decrease power dissipation during periods when activity on the input pins is not of interest to the CPLD. The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is Low. The data on the D input during the Low-to-High gate transition is stored in the latch. The data on the Q output remains unchanged as long as G remains High.

The D input(s) of the LDG must be connected to a device input pad(s) and must have no other fan-outs (must not branch). The CPLD fitter maps the G input to the device's DataGate Enable control pin (DGE). There must be no more than one DataGate Enable signal in the design. The DataGate Enable signal may be driven either by a device input pin or any on-chip logic source. The DataGate Enable signal may be reused by other ordinary logic in the design.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

### Logic Table

Inputs		Outputs
G	D	Q
0	0	0
0	1	1
1	Х	No Change
-	D	D

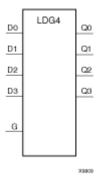
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 4-Bit Transparent Datagate Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element has 4 transparent DataGate latches with a common gate enable (G). These latches are used to gate input signals in order to decrease power dissipation during periods when activity on the input pins is not of interest to the CPLD. The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is Low. The data on the D input during the Low-to-High gate transition is stored in the latch. The data on the Q output remains unchanged as long as G remains High.

The D input(s) of the LDG must be connected to a device input pad(s) and must have no other fan-outs (must not branch). The CPLD fitter maps the G input to the device's DataGate Enable control pin (DGE). There must be no more than one DataGate Enable signal in the design. The DataGate Enable signal may be driven either by a device input pin or any on-chip logic source. The DataGate Enable signal may be reused by other ordinary logic in the design.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

# **Logic Table**

Inputs		Outputs
G	D	Q
0	0	0
0	1	1
1	X	No Change
-	D	D

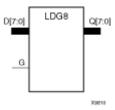
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 8-Bit Transparent Datagate Latch



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element has 8 transparent DataGate latches with a common gate enable (G). These latches are used to gate input signals in order to decrease power dissipation during periods when activity on the input pins is not of interest to the CPLD. The data output (Q) of the latch reflects the data (D) input while the gate enable (G) input is Low. The data on the D input during the Low-to-High gate transition is stored in the latch. The data on the Q output remains unchanged as long as G remains High.

The D input(s) of the LDG must be connected to a device input pad(s) and must have no other fan-outs (must not branch). The CPLD fitter maps the G input to the device's DataGate Enable control pin (DGE). There must be no more than one DataGate Enable signal in the design. The DataGate Enable signal may be driven either by a device input pin or any on-chip logic source. The DataGate Enable signal may be reused by other ordinary logic in the design.

This latch is asynchronously cleared, outputs Low, when power is applied. For CPLD devices, you can simulate power-on by applying a High-level pulse on the PRLD global net.

## **Logic Table**

Inputs		Outputs
G	D	Q
0	0	0
0	1	1
1	X	No Change
-	D	D

# **Design Entry Method**

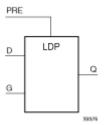
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **LDP**

### Macro: Transparent Data Latch with Asynchronous Preset



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a transparent data latch with asynchronous preset (PRE). When the (PRE) input is High, it overrides the other inputs and presets the data (Q) output High. (Q) reflects the data (D) input while gate (G) input is High and (PRE) is Low. The data on the (D) input during the High-to-Low gate transition is stored in the latch. The data on the (Q) output remains unchanged as long as (G) remains Low.

The latch is asynchronously preset, output High, when power is applied.

## **Logic Table**

Inputs			Outputs
PRE	G	D	Q
1	X	X	1
0	1	0	0
0	1	1	1
0	0	X	No Change
0	-	D	D

# **Design Entry Method**

This design element is only for use in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
INIT	INTEGER	0 or 1	0	Specifies the initial value upon power-up or the assertion of GSR for the (Q) port.

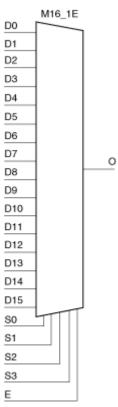


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M16\_1E

## Macro: 16-to-1 Multiplexer with Enable



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

This design element is a 16-to-1 multiplexer with enable. When the enable input (E) is High, the M16\_1E multiplexer chooses one data bit from 16 sources (D15 – D0) under the control of the select inputs (S3 – S0). The output (O) reflects the state of the selected input as shown in the logic table. When (E) is Low, the output is Low.

## **Logic Table**

Inputs					Outputs	
E	S3	S2	S1	S0	D15-D0	О
0	Х	X	X	Х	X	0
1	0	0	0	0	D0	D0
1	0	0	0	1	D1	D1
1	0	0	1	0	D2	D2



Inputs					Outputs	
E	S3	S2	S1	S0	D15-D0	0
1	0	0	1	1	D3	D3
1	1	1	0	0	D12	D12
1	1	1	0	1	D13	D13
1	1	1	1	0	D14	D14
1	1	1	1	1	D15	D15

# **Design Entry Method**

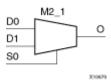
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M<sub>2</sub> 1

### Macro: 2-to-1 Multiplexer



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element chooses one data bit from two sources (D1 or D0) under the control of the select input (S0). The output (O) reflects the state of the selected data input. When Low, S0 selects D0 and when High, S0 selects D1.

## **Logic Table**

Inputs			Outputs
S0	D1	D0	0
1	D1	X	D1
0	Χ	D0	D0

## **Design Entry Method**

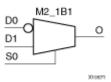
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M<sub>2</sub> 1B<sub>1</sub>

## Macro: 2-to-1 Multiplexer with D0 Inverted



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element chooses one data bit from two sources (D1 or D0) under the control of select input (S0). When S0 is Low, the output (O) reflects the inverted value of (D0). When S0 is High, (O) reflects the state of D1.

## **Logic Table**

Inputs			Outputs
S0	D1	D0	0
1	1	X	1
1	0	X	0
0	X	1	0
0	X	0	1

## **Design Entry Method**

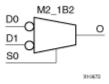
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M2 1B2

## Macro: 2-to-1 Multiplexer with D0 and D1 Inverted



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

This design element chooses one data bit from two sources (D1 or D0) under the control of select input (S0). When S0 is Low, the output (O) reflects the inverted value of D0. When S0 is High, O reflects the inverted value of D1.

# **Logic Table**

Inputs			Outputs
S0	D1	D0	0
1	1	X	0
1	0	X	1
0	X	1	0
0	Х	0	1

# **Design Entry Method**

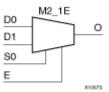
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M2 1E

### Macro: 2-to-1 Multiplexer with Enable



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

This design element is a 2-to-1 multiplexer with enable. When the enable input (E) is High, the M2\_1E chooses one data bit from two sources (D1 or D0) under the control of select input (S0). When Low, S0 selects D0 and when High, S0 selects D1. When (E) is Low, the output is Low.

# **Logic Table**

Inputs	Outputs			
E	S0	D1	D0	0
0	Χ	Χ	Χ	0
1	0	Χ	1	1
1	0	X	0	0
1	1	1	Χ	1
1	1	0	Х	0

# **Design Entry Method**

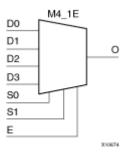
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M4 1E

### Macro: 4-to-1 Multiplexer with Enable



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a 4-to-1 multiplexer with enable. When the enable input (E) is High, the M4\_1E multiplexer chooses one data bit from four sources (D3, D2, D1, or D0) under the control of the select inputs (S1 – S0). The output (O) reflects the state of the selected input as shown in the logic table. When (E) is Low, the output is Low.

## **Logic Table**

Inputs							Outputs
Е	S1	S0	D0	D1	D2	D3	0
0	Х	Х	Х	Х	Х	Х	0
1	0	0	D0	Х	Х	Х	D0
1	0	1	Х	D1	Х	Х	D1
1	1	0	Х	Х	D2	Х	D2
1	1	1	X	Х	X	D3	D3

# **Design Entry Method**

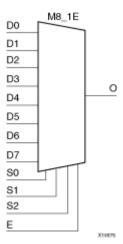
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# M8 1E

#### Macro: 8-to-1 Multiplexer with Enable



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

This design element is an 8-to-1 multiplexer with enable. When the enable input (E) is High, the  $M8_1E$  multiplexer chooses one data bit from eight sources (D7 – D0) under the control of the select inputs (S2 – S0). The output (O) reflects the state of the selected input as shown in the logic table. When (E) is Low, the output is Low.

# **Logic Table**

Inputs	Outputs				
E	S2	S1	S0	D7-D0	О
0	X	X	X	X	0
1	0	0	0	D0	D0
1	0	0	1	D1	D1
1	0	1	0	D2	D2
1	0	1	1	D3	D3
1	1	0	0	D4	D4
1	1	0	1	D5	D5
1	1	1	0	D6	D6
1	1	1	1	D7	D7

# **Design Entry Method**

This design element is only for use in schematics.



- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NAND2

## Primitive: 2-Input NAND Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND2B1

### Primitive: 2-Input NAND Gate with 1 Inverted and 1 Non-Inverted Inputs

NAND2B1



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND2B2

## Primitive: 2-Input NAND Gate with Inverted Inputs

NAND2B2



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND3

## Primitive: 3-Input NAND Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND3B1

### Primitive: 3-Input NAND Gate with 1 Inverted and 2 Non-Inverted Inputs

NAND3B1 12 11 10 X10748

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND3B2

## Primitive: 3-Input NAND Gate with 2 Inverted and 1 Non-Inverted Inputs

NAND3B2



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND3B3

### Primitive: 3-Input NAND Gate with Inverted Inputs

NAND3B3



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND4

## Primitive: 4-Input NAND Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND4B1

### Primitive: 4-Input NAND Gate with 1 Inverted and 3 Non-Inverted Inputs

NAND4B1



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NAND4B2

### Primitive: 4-Input NAND Gate with 2 Inverted and 2 Non-Inverted Inputs

NAND4B2



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NAND4B3

#### Primitive: 4-Input NAND Gate with 3 Inverted and 1 Non-Inverted Inputs

NAND4B3



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NAND4B4

#### Primitive: 4-Input NAND Gate with Inverted Inputs

NAND4B4



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

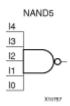
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NAND Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

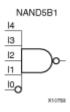
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NAND Gate with 1 Inverted and 4 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NAND Gate with 2 Inverted and 3 Non-Inverted Inputs

NAND5B2



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NAND Gate with 3 Inverted and 2 Non-Inverted Inputs

NAND5B3



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

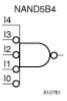
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NAND Gate with 4 Inverted and 1 Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

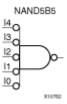
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NAND Gate with Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

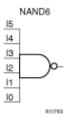
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 6-Input NAND Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

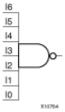
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 7-Input NAND Gate with Non-Inverted Inputs

NAND7



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

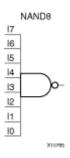
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 8-Input NAND Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

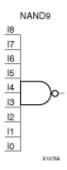
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 9-Input NAND Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NAND gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NAND gates of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NOR<sub>2</sub>

#### Primitive: 2-Input NOR Gate with Non-Inverted Inputs

NOR2

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NOR<sub>2</sub>B<sub>1</sub>

### Primitive: 2-Input NOR Gate with 1 Inverted and 1 Non-Inverted Inputs

NOR2B1 10 NOR2B1 NOR2B1

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### NOR2B2

#### Primitive: 2-Input NOR Gate with Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NOR<sub>3</sub>

### Primitive: 3-Input NOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## NOR3B1

### Primitive: 3-Input NOR Gate with 1 Inverted and 2 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NOR3B2

### Primitive: 3-Input NOR Gate with 2 Inverted and 1 Non-Inverted Inputs

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# NOR3B3

### Primitive: 3-Input NOR Gate with Inverted Inputs

NOR3B3



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

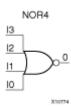
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input NOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input NOR Gate with 1 Inverted and 3 Non-Inverted Inputs

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input NOR Gate with 2 Inverted and 2 Non-Inverted Inputs

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 4-Input NOR Gate with 3 Inverted and 1 Non-Inverted Inputs

NOR4B3



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input NOR Gate with Inverted Inputs

NOR4B4



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

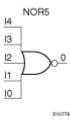
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input NOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

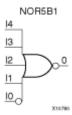
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input NOR Gate with 1 Inverted and 4 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 5-Input NOR Gate with 2 Inverted and 3 Non-Inverted Inputs

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

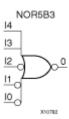
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input NOR Gate with 3 Inverted and 2 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

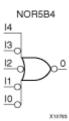
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input NOR Gate with 4 Inverted and 1 Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

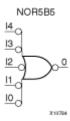
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input NOR Gate with Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

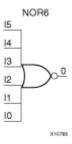
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 6-Input NOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

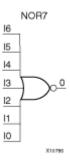
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 7-Input NOR Gate with Non-Inverted Inputs



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

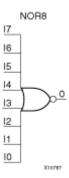
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 8-Input NOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

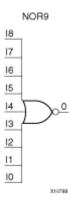
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 9-Input NOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

NOR gates of up to five inputs are available in any combination of inverting and non-inverting inputs. NOR gates of six to nine inputs, 12 inputs, and 16 inputs are available only with non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace gates with unused inputs with gates having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OBUF**

Primitive: Output Buffer

OBUF

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a simple output buffer used to drive output signals to the FPGA device pins that do not need to be 3-stated (constantly driven). Either an OBUF, OBUFT, OBUFDS, or OBUFTDS must be connected to every output port in the design.

This element isolates the internal circuit and provides drive current for signals leaving a chip. It exists in input/output blocks (IOB). Its output (O) is connected to an OPAD or an IOPAD. The interface standard used by this element is LVTTL. Also, this element has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

# **Port Descriptions**

Name	Direction	Width	Function
О	Output	1-bit	Output of OBUF to be connected directly to top-level output port.
I	Input	1-bit	Input of OBUF. Connect to the logic driving the output port.

Instantiation	Yes	
Inference	Recommended	
Coregen and wizards	No	
Macro support	No	

This design element can be used in schematics.

#### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.



Attribute	Type	Allowed Values	Default	Description
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration.

```
Library UNISIM;
use UNISIM.vcomponents.all;

-- OBUF: Single-ended Output Buffer
-- All devices
-- Xilinx HDL Libraries Guide, version 10.1.1

OBUF_inst : OBUF
generic map (
DRIVE => 12,
IOSTANDARD => "DEFAULT",
SLEW => "SLOW")
port map (
O => O, -- Buffer output (connect directly to top-level port)
I => I -- Buffer input
);

-- End of OBUF_inst instantiation
```

## **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OBUF16**

Macro: 16-Bit Output Buffer

OBUF16



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple output buffer.

This element isolates the internal circuit and provides drive current for signals leaving a chip. It exists in input/output blocks (IOB). Its output (O) is connected to an OPAD or an IOPAD. The interface standard used by this element is LVTTL. Also, this element has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.

#### For More Information

• See the appropriate CPLD User Guide.

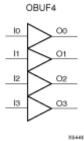


• See the appropriate CPLD Data Sheets.



# **OBUF4**

Macro: 4-Bit Output Buffer



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple output buffer.

This element isolates the internal circuit and provides drive current for signals leaving a chip. It exists in input/output blocks (IOB). Its output (O) is connected to an OPAD or an IOPAD. The interface standard used by this element is LVTTL. Also, this element has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

### **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.



- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **OBUF8**

Macro: 8-Bit Output Buffer

OBUF8



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple output buffer.

This element isolates the internal circuit and provides drive current for signals leaving a chip. It exists in input/output blocks (IOB). Its output (O) is connected to an OPAD or an IOPAD. The interface standard used by this element is LVTTL. Also, this element has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

## **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.

#### For More Information

• See the appropriate CPLD User Guide.



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• See the appropriate CPLD Data Sheets.



### Macro: 3-State Output Buffer with Active-High Output Enable

OBUFE



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 3-state buffer with input I, output O, and active-High output enable (E).

When E is High, data on the inputs of the buffers is transferred to the corresponding outputs. When E is Low, the output is High impedance (off or Z state). An *OBUFE* isolates the internal circuit and provides drive current for signals leaving a chip. An *OBUFE* output is connected to an OPAD or an IOPAD. An *OBUFE* input is connected to the internal circuit.

# **Logic Table**

Inputs		Outputs
Е	I	0
0	X	Z
1	1	1
1	0	0

# **Design Entry Method**

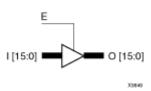
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Macro: 16-Bit 3-State Output Buffer with Active-High Output Enable

OBUFE16



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a 3-state buffer with input I15-I0, output O15-O0, and active-High output enable (E).

When E is High, data on the inputs of the buffers is transferred to the corresponding outputs. When E is Low, the output is High impedance (off or Z state). An *OBUFE* isolates the internal circuit and provides drive current for signals leaving a chip. An *OBUFE* output is connected to an OPAD or an IOPAD. An *OBUFE* input is connected to the internal circuit.

## **Logic Table**

Inputs		Outputs
E	I	0
0	X	Z
1	1	1
1	0	0

## **Design Entry Method**

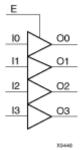
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Macro: 4-Bit 3-State Output Buffer with Active-High Output Enable

OBUFE4



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a 3-state buffer with input I3-I0, output O3-O0, and active-High output enable (E).

When E is High, data on the inputs of the buffers is transferred to the corresponding outputs. When E is Low, the output is High impedance (off or Z state). An *OBUFE* isolates the internal circuit and provides drive current for signals leaving a chip. An *OBUFE* output is connected to an OPAD or an IOPAD. An *OBUFE* input is connected to the internal circuit.

# **Logic Table**

Inputs		Outputs
Е	I	0
0	X	Z
1	1	1
1	0	0

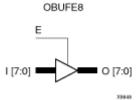
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Macro: 8-Bit 3-State Output Buffer with Active-High Output Enable



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a 3-state buffer with input I7-I0, output O7-O0, and active-High output enable (E).

When E is High, data on the inputs of the buffers is transferred to the corresponding outputs. When E is Low, the output is High impedance (off or Z state). An *OBUFE* isolates the internal circuit and provides drive current for signals leaving a chip. An *OBUFE* output is connected to an OPAD or an IOPAD. An *OBUFE* input is connected to the internal circuit.

# **Logic Table**

Inputs		Outputs
E	I	0
0	X	Z
1	1	1
1	0	0

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OBUFT**

### Primitive: 3-State Output Buffer with Active Low Output Enable

OBUFT



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a single, 3-state output buffer with input I, output O, and active-Low output enables (T). This element uses the LVTTL standard and has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (off or Z state). OBUFTs are generally used when a single-ended output is needed with a 3-state capability, such as the case when building bidirectional I/O.

## **Logic Table**

Inputs		Outputs
Т	I	0
1	Х	Z
0	I	F

# **Port Descriptions**

Name	Direction	Width	Function
0	Output	1-Bit	Buffer output (connect directly to top-level port)
I	Input	1-Bit	Buffer input
T	Input	1-Bit	3-state enable input

# **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.



### **Available Attributes**

Attribute	Туре	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- OBUFT: Single-ended 3-state Output Buffer
         All devices
-- Xilinx HDL Libraries Guide, version 10.1.1
OBUFT_inst : OBUFT
generic map (
DRIVE \Rightarrow 12.
IOSTANDARD => "DEFAULT",
SLEW => "SLOW")
port map (
0 => 0,
            -- Buffer output (connect directly to top-level port)
           -- Buffer input
I => I,
           -- 3-state enable input
T => T
);
-- End of OBUFT_inst instantiation
```

# **Verilog Instantiation Template**

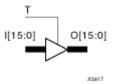
- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OBUFT16**

## Macro: 16-Bit 3-State Output Buffer with Active Low Output Enable

OBUFT16



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple, 3-state output buffer with input I, output O, and active-Low output enables (T). This element uses the LVTTL standard and has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (off or Z state). OBUFTs are generally used when a single-ended output is needed with a 3-state capability, such as the case when building bidirectional I/O.

## **Logic Table**

Inputs		Outputs
T	I	0
1	Х	Z
0	I	F

## **Design Entry Method**

This design element is only for use in schematics.

## **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.



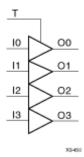
- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OBUFT4**

## Macro: 4-Bit 3-State Output Buffers with Active-Low Output Enable

OBUFT4



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element is a multiple, 3-state output buffer with input I, output O, and active-Low output enables (T). This element uses the LVTTL standard and has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (off or Z state). OBUFTs are generally used when a single-ended output is needed with a 3-state capability, such as the case when building bidirectional I/O.

## **Logic Table**

Inputs		Outputs
T	I	0
1	Х	Z
0	I	F

### **Design Entry Method**

This design element is only for use in schematics.

## **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.



Attribute	Type	Allowed Values	Default	Description
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OBUFT8**

### Macro: 8-Bit 3-State Output Buffers with Active-Low Output Enable

OBUFT8



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a multiple, 3-state output buffer with input I, output O, and active-Low output enables (T). This element uses the LVTTL standard and has selectable drive and slew rates using the DRIVE and SLOW or FAST constraints. The defaults are DRIVE=12 mA and SLOW slew.

When T is Low, data on the inputs of the buffers is transferred to the corresponding outputs. When T is High, the output is high impedance (off or Z state). OBUFTs are generally used when a single-ended output is needed with a 3-state capability, such as the case when building bidirectional I/O.

# Logic Table

Inputs		Outputs
Т	I	0
1	Х	Z
0	I	F

# **Design Entry Method**

This design element is only for use in schematics.

## **Available Attributes**

Attribute	Type	Allowed Values	Default	Description
DRIVE	Integer	2, 4, 6, 8, 12, 16, 24	12	Specifies the output current drive strength of the I/O. It is suggested that you set this to the lowest setting tolerable for the design drive and timing requirements.
IOSTANDARD	String	Consult the product Data Sheet.	"DEFAULT"	Specifies the I/O standard to be used for this output.
SLEW	String	"SLOW" or "FAST"	"SLOW"	Specifies the slew rate of the output driver. Consult the product Data Sheet for recommendations of the best setting for this attribute.



- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## OR<sub>2</sub>

### Primitive: 2-Input OR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **OR2B1**

## Primitive: 2-Input OR Gate with 1 Inverted and 1 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# OR2B2

### Primitive: 2-Input OR Gate with Inverted Inputs

OR2B2

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# OR3

## Primitive: 3-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OR3B1**

## Primitive: 3-Input OR Gate with 1 Inverted and 2 Non-Inverted Inputs

OR3B1



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OR3B2**

## Primitive: 3-Input OR Gate with 2 Inverted and 1 Non-Inverted Inputs

OR3B2

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# **OR3B3**

### Primitive: 3-Input OR Gate with Inverted Inputs

OR3B3

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

## **Design Entry Method**

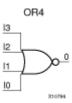
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## OR4

## Primitive: 4-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: 4-Input OR Gate with 1 Inverted and 3 Non-Inverted Inputs

OR4B1



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: 4-Input OR Gate with 2 Inverted and 2 Non-Inverted Inputs

OR4B2

## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

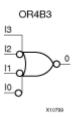
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: 4-Input OR Gate with 3 Inverted and 1 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

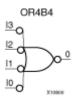
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input OR Gate with Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

## Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

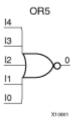
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## OR5

## Primitive: 5-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

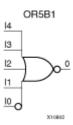
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: 5-Input OR Gate with 1 Inverted and 4 Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

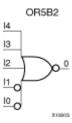
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input OR Gate with 2 Inverted and 3 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

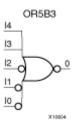
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: 5-Input OR Gate with 3 Inverted and 2 Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

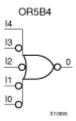
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## Primitive: 5-Input OR Gate with 4 Inverted and 1 Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

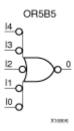
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **OR5B5**

### Primitive: 5-Input OR Gate with Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

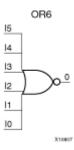
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 6-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

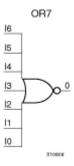
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 7-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

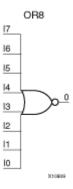
### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 8-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

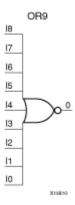
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Macro: 9-Input OR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

OR functions of up to five inputs are available in any combination of inverting and non-inverting inputs. OR functions of six to nine inputs, 12 inputs, and 16 inputs are available with only non-inverting inputs. To invert some or all inputs, use external inverters. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **PULLDOWN**

Primitive: Resistor to GND for Input Pads, Open-Drain, and 3-State Outputs

#### PULLDOWN



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This resistor element is connected to input, output, or bidirectional pads to guarantee a logic Low level for nodes that might float.

### **Port Descriptions**

Name	Name Direction		Function			
0	Output	1-Bit	Pulldown output (connect directly to top level port)			

### **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

This element can be connected to a net in the following locations on a top-level schematic file:

- A net connected to an input IO Marker
- A net connected to both an output IO Marker and 3-statable IO element, such as an OBUFT.

## **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- PULLDOWN: I/O Buffer Weak Pull-down
-- All FPGA
-- Xilinx HDL Libraries Guide, version 10.1.1

PULLDOWN_inst: PULLDOWN
port map (
0 => 0 -- Pulldown output (connect directly to top-level port)
```



```
);
-- End of PULLDOWN_inst instantiation
```

## **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### **PULLUP**

Primitive: Resistor to VCC for Input PADs, Open-Drain, and 3-State Outputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element allows for an input, 3-state output or bi-directional port to be driven to a weak high value when not being driven by an internal or external source. This element establishes a High logic level for open-drain elements and macros when all the drivers are off.

## **Port Descriptions**

Name	Name Direction		Function			
О	Output	1-Bit	Pullup output (connect directly to top level port)			

## **Design Entry Method**

Instantiation	Yes
Inference	Recommended
Coregen and wizards	No
Macro support	No

This design element can be used in schematics.

This element can be connected to a net in the following locations on a top-level schematic file:

- A net connected to an input IO Marker
- A net connected to both an output IO Marker and 3-statable IO element, such as an OBUFT.

### **VHDL Instantiation Template**

Unless they already exist, copy the following two statements and paste them before the entity declaration. Library UNISIM;

```
use UNISIM.vcomponents.all;
-- PULLUP: I/O Buffer Weak Pull-up
-- All FPGA, CoolRunner-II
-- Xilinx HDL Libraries Guide, version 10.1.1

PULLUP_inst: PULLUP
port map (
0 => 0 -- Pullup output (connect directly to top-level port)
);
```



```
-- End of PULLUP_inst instantiation
```

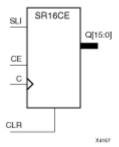
## **Verilog Instantiation Template**

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SR16CE

Macro: 16-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with a shift-left serial input (SLI), parallel outputs (Q), and clock enable (CE) and asynchronous clear (CLR) inputs. The (CLR) input, when High, overrides all other inputs and resets the data outputs (Q) Low. When (CE) is High and (CLR) is Low, the data on the SLI input is loaded into the first bit of the shift register during the Low-to- High clock (C) transition and appears on the (Q0) output. During subsequent Low-to- High clock transitions, when (CE) is High and (CLR) is Low, data shifts to the next highest bit position as new data is loaded into (Q0) (SLI® Q0, Q0® Q1, Q1® Q2, and so forth). The register ignores clock transitions when (CE) is Low.

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), and (CLR) in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs		Outputs			
CLR	CE	SLI	С	Q0	Qz - Q1
1	Χ	Χ	Χ	0	0
0	0	X	X	No Change	No Change
0	1	SLI	-	SLI	qn-1

z = bit width - 1

qn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

See the appropriate CPLD User Guide.

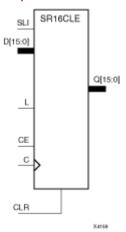


• See the appropriate CPLD Data Sheets.



## SR16CLE

Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with a shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and asynchronous clear (CLR). The register ignores clock transitions when (L) and (CE) are Low. The asynchronous (CLR), when High, overrides all other inputs and resets the data outputs (Q) Low. When (L) is High and (CLR) is Low, data on the Dn-D0 inputs is loaded into the corresponding Qn-Q0 bits of the register.

When (CE) is High and (L) and (CLR) are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the (Q0) output. During subsequent clock transitions, when (CE) is High and (L) and (CLR) are Low, the data shifts to the next highest bit position as new data is loaded into (Q)0 (for example, SLI® Q0, Q0® Q1, and Q1® Q2).

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), (L), and (CLR) inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

#### Logic Table

Inputs						Outputs			
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1		
1	Χ	X	X	Χ	Х	0	0		
0	1	X	X	Dn – D0	-	D0	Dn		
0	0	1	SLI	Х	-	SLI	qn-1		
0	0	0	Χ	Х	Х	No Change	No Change		
z = bitwidth									

qn-1 = state of referenced output one setup time prior to active clock transition



# **Design Entry Method**

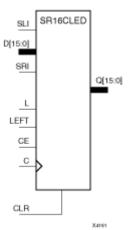
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SR16CLED

Macro: 16-Bit Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs: clock enable (CE), load enable (L), shift left/right (LEFT), and asynchronous clear (CLR). The register ignores clock transitions when (CE) and (L) are Low. The asynchronous clear, when High, overrides all other inputs and resets the data outputs (Qn) Low.

When (L) is High and (CLR) is Low, the data on the (D) inputs is loaded into the corresponding (Q) bits of the register. When (CE) is High and (L) and (CLR) are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on the SLI is loaded into (Q0) during the Low-to-High clock transition and shifted left (for example, to Q1 or Q2) during subsequent clock transitions. If LEFT is Low, data on the SRI is loaded into the last (Q) output during the Low-to-High clock transition and shifted right during subsequent clock transitions. The logic tables indicate the state of the (Q) outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs	Inputs								Outputs		
CLR	L	CE	LEFT	SLI	SRI	D15 - D0	С	Q0	Q15	Q14 - Q1	
1	Х	X	Х	Х	Χ	Х	Х	0	0	0	
0	1	Χ	Χ	X	X	D15-D0	-	D0	D15	Dn	
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change	
0	0	1	1	SLI	X	X	-	SLI	q14	qn-1	



Inputs	Inputs							Outputs		
CLR	L	CE	LEFT	SLI	SRI	D15 - D0	С	Q0	Q15	Q14 - Q1
0	0	1	0	Х	SRI	X	-	q1	SRI	qn+1

qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition.

# **Design Entry Method**

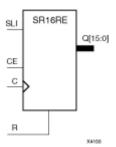
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### SR16RE

Macro: 16-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left serial input (SLI), parallel outputs (Qn), clock enable (CE), and synchronous reset (R) inputs. The R input, when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low.

When (CE) is High and (R) is Low, the data on the (SLI) is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the (Q0) output. During subsequent Low-to-High clock transitions, when (CE) is High and R is Low, data shifts to the next highest bit position as new data is loaded into (Q0) (for example, SLI® Q0, Q0® Q1, and Q1® Q2). The register ignores clock transitions when (CE) is Low.

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), and (R) in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs		Outputs			
R	CE	SLI	С	Q0	Qz - Q1
1	X	X	-	0	0
0	0	Χ	Χ	No Change	No Change
0	1	SLI	-	SLI	qn-1

z = bitwidth -1

### **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

See the appropriate CPLD User Guide.

qn-1 = state of referenced output one setup time prior to active clock transition

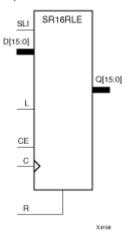


• See the appropriate CPLD Data Sheets.



## SR16RLE

Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and synchronous reset (R). The register ignores clock transitions when (L) and (CE) are Low. The synchronous (R), when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low. When (L) is High and (R) is Low during the Low-to-High clock transition, data on the (D) inputs is loaded into the corresponding Q bits of the register.

When (CE) is High and (L) and (R) are Low, data on the (SLI) input is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when (CE) is High and (L) and (R) are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, (CE), (L), and (R) inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

## **Logic Table**

Inputs						Outputs	
R	L	CE	SLI	Dz - D0	С	Q0	Qz - Q1
1	X	X	X	X	-	0	0
0	1	X	X	Dz – D0	-	D0	Dn
0	0	1	SLI	Х	-	SLI	qn-1



Inputs							Outputs	
R	L CE SLI Dz - D0 C						Qz - Q1	
0	0	0	X	Χ	X	No Change	No Change	
z = bitwidth -1								
qn-1 = state of	qn-1 = state of referenced output one setup time prior to active clock transition							

# **Design Entry Method**

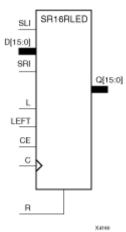
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SR16RLED

Macro: 16-Bit Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q) and four control inputs — clock enable (CE), load enable (L), shift left/right (LEFT), and synchronous reset (R). The register ignores clock transitions when (CE) and (L) are Low. The synchronous (R), when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low. When (L) is High and (R) is Low during the Low-to-High clock transition, the data on the (D) inputs is loaded into the corresponding (Q) bits of the register.

When (CE) is High and (L) and (R) are Low, data shifts right or left, depending on the state of the LEFT input. If LEFT is High, data on (SLI) is loaded into (Q0) during the Low-to-High clock transition and shifted left (for example, to Q1 and Q2) during subsequent clock transitions. If LEFT is Low, data on the (SRI) is loaded into the last (Q) output during the Low-to-High clock transition and shifted right) during subsequent clock transitions. The logic tables below indicates the state of the (Q) outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied.

### Logic Table

Inputs	Inputs								Outputs		
R	L	CE	LEFT	SLI	SRI	D15 - D0	С	Q0	Q15	Q14 - Q1	
1	X	Χ	X	X	Χ	Χ	-	0	0	0	
0	1	Х	X	X	Х	D15-D0	-	D0	D15	Dn	
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change	
0	0	1	1	SLI	X	Χ	-	SLI	q14	qn-1	



Inputs	Inputs									Outputs			
R	L	CE	LEFT	SLI	SRI	D15 - D0	С	Q0	Q15	Q14 - Q1			
0	0	1	0	Х	SRI	Х	-	q1	SRI	qn+1			

qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition

# **Design Entry Method**

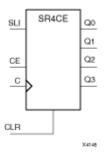
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SR4CE

Macro: 4-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear



### **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with a shift-left serial input (SLI), parallel outputs (Q), and clock enable (CE) and asynchronous clear (CLR) inputs. The (CLR) input, when High, overrides all other inputs and resets the data outputs (Q) Low. When (CE) is High and (CLR) is Low, the data on the SLI input is loaded into the first bit of the shift register during the Low-to- High clock (C) transition and appears on the (Q0) output. During subsequent Low-to- High clock transitions, when (CE) is High and (CLR) is Low, data shifts to the next highest bit position as new data is loaded into (Q0) (SLI® Q0, Q0® Q1, Q1® Q2, and so forth). The register ignores clock transitions when (CE) is Low.

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), and (CLR) in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs			Outputs			
CLR	CE	SLI	С	Q0	Qz – Q1	
1	X	Χ	Χ	0	0	
0	0	X	X	No Change	No Change	
0	1	SLI	-	SLI	qn-1	

z = bit width - 1

qn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

• See the appropriate CPLD User Guide.

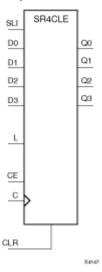


• See the appropriate CPLD Data Sheets.



## SR4CLE

Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with a shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and asynchronous clear (CLR). The register ignores clock transitions when (L) and (CE) are Low. The asynchronous (CLR), when High, overrides all other inputs and resets the data outputs (Q) Low. When (L) is High and (CLR) is Low, data on the Dn-D0 inputs is loaded into the corresponding Qn-Q0 bits of the register.

When (CE) is High and (L) and (CLR) are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the (Q0) output. During subsequent clock transitions, when (CE) is High and (L) and (CLR) are Low, the data shifts to the next highest bit position as new data is loaded into (Q)0 (for example, SLI® Q0, Q0® Q1, and Q1® Q2).

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), (L), and (CLR) inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs			Outputs				
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1
1	Χ	Χ	Χ	Χ	Χ	0	0
0	1	X	X	Dn – D0	-	D0	Dn
0	0	1	SLI	X	-	SLI	qn-1



Inputs			Outputs							
CLR	CLR L CE SLI Dn – D0 C					Q0 Qz - Q1				
0	0	0	Χ	X	X	No Change	No Change			
z = bitwidth	z = bitwidth -1									

qn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

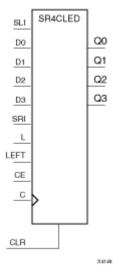
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **SR4CLED**

Macro: 4-Bit Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs: clock enable (CE), load enable (L), shift left/right (LEFT), and asynchronous clear (CLR). The register ignores clock transitions when (CE) and (L) are Low. The asynchronous clear, when High, overrides all other inputs and resets the data outputs (Qn) Low.

When (L) is High and (CLR) is Low, the data on the (D) inputs is loaded into the corresponding (Q) bits of the register. When (CE) is High and (L) and (CLR) are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on the SLI is loaded into (Q0) during the Low-to-High clock transition and shifted left (for example, to Q1 or Q2) during subsequent clock transitions. If LEFT is Low, data on the SRI is loaded into the last (Q) output during the Low-to-High clock transition and shifted right during subsequent clock transitions. The logic tables indicate the state of the (Q) outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied.

## Logic Table

Inputs			Outputs							
CLR	L	CE	LEFT	SLI	SRI	D3 – D0	С	Q0	Q3	Q2 – Q1
1	X	Х	Χ	Χ	Χ	X	Χ	0	0	0
0	1	X	Χ	X	Χ	D3- D0	-	D0	D3	Dn
0	0	0	Х	X	Х	Х	X	No Change	No Change	No Change
0	0	1	1	SLI	X	X	-	SLI	q2	qn-1



Inputs	Inputs									
CLR	L	CE	LEFT	SLI	SRI	D3 – D0	С	Q0	Q3	Q2 – Q1
0	0 0 1 0 X SRI X - q1 SRI qn+1									
qn-1 and qn+1 = state of referenced output one setup time prior to active clock transition.										

# **Design Entry Method**

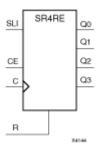
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SR4RE

Macro: 4-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left serial input (SLI), parallel outputs (Qn), clock enable (CE), and synchronous reset (R) inputs. The R input, when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low.

When (CE) is High and (R) is Low, the data on the (SLI) is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the (Q0) output. During subsequent Low-to-High clock transitions, when (CE) is High and R is Low, data shifts to the next highest bit position as new data is loaded into (Q0) (for example, SLI® Q0, Q0® Q1, and Q1® Q2). The register ignores clock transitions when (CE) is Low.

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), and (R) in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs		Outputs				
R	CE	SLI	С	Q0	Qz – Q1	
1	X	X	-	0	0	
0	0	X	X	No Change	No Change	
0	1	SLI	-	SLI	qn-1	

z = bitwidth -1

## **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

See the appropriate CPLD User Guide.

qn-1 = state of referenced output one setup time prior to active clock transition

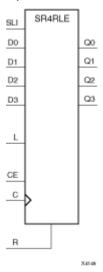


• See the appropriate CPLD Data Sheets.



## SR4RLE

Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and synchronous reset (R). The register ignores clock transitions when (L) and (CE) are Low. The synchronous (R), when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low. When (L) is High and (R) is Low during the Low-to-High clock transition, data on the (D) inputs is loaded into the corresponding Q bits of the register.

When (CE) is High and (L) and (R) are Low, data on the (SLI) input is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when (CE) is High and (L) and (R) are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, (CE), (L), and (R) inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

## Logic Table

Inputs		Outputs					
R	L	CE	SLI	Dz - D0	С	Q0	Qz - Q1
1	X	X	X	X	-	0	0
0	1	X	X	Dz – D0	-	D0	Dn
0	0	1	SLI	Х	-	SLI	qn-1



Inputs		Outputs						
R	L CE SLI Dz – D0 C					Q0	Qz - Q1	
0	0 0 X X X No Change							
z = bitwidth -1								
qn-1 = state of referenced output one setup time prior to active clock transition								

# **Design Entry Method**

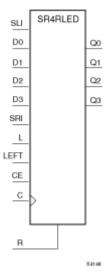
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **SR4RLED**

Macro: 4-Bit Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q) and four control inputs — clock enable (CE), load enable (L), shift left/right (LEFT), and synchronous reset (R). The register ignores clock transitions when (CE) and (L) are Low. The synchronous (R), when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low. When (L) is High and (R) is Low during the Low-to-High clock transition, the data on the (D) inputs is loaded into the corresponding (Q) bits of the register.

When (CE) is High and (L) and (R) are Low, data shifts right or left, depending on the state of the LEFT input. If LEFT is High, data on (SLI) is loaded into (Q0) during the Low-to-High clock transition and shifted left (for example, to Q1 and Q2) during subsequent clock transitions. If LEFT is Low, data on the (SRI) is loaded into the last (Q) output during the Low-to-High clock transition and shifted right) during subsequent clock transitions. The logic tables below indicates the state of the (Q) outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied.

### Logic Table

Inputs	Inputs									Outputs			
R	L	CE	LEFT	SLI	SRI	D3 – D0	С	Q0	Q3	Q2 – Q1			
1	Х	Х	Х	Х	Х	Х	-	0	0	0			
0	1	Χ	Χ	X	Χ	D3 – D0	-	D0	D3	Dn			
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change			



Inputs	Inputs									Outputs			
R	L	CE	LEFT	SLI	SRI	D3 – D0	С	Q0	Q3	Q2 – Q1			
0	0	1	1	SLI	Х	Х	-	SLI	q2	qn-1			
0	0	1	0	Х	SRI	Х	-	q1	SRI	qn+1			

qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition

# **Design Entry Method**

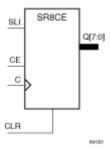
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SR8CE

Macro: 8-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with a shift-left serial input (SLI), parallel outputs (Q), and clock enable (CE) and asynchronous clear (CLR) inputs. The (CLR) input, when High, overrides all other inputs and resets the data outputs (Q) Low. When (CE) is High and (CLR) is Low, the data on the SLI input is loaded into the first bit of the shift register during the Low-to- High clock (C) transition and appears on the (Q0) output. During subsequent Low-to- High clock transitions, when (CE) is High and (CLR) is Low, data shifts to the next highest bit position as new data is loaded into (Q0) (SLI® Q0, Q0® Q1, Q1® Q2, and so forth). The register ignores clock transitions when (CE) is Low.

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), and (CLR) in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

### **Logic Table**

Inputs		Outputs			
CLR	CE	SLI	С	Q0	Qz - Q1
1	X	Χ	X	0	0
0	0	X	Χ	No Change	No Change
0	1	SLI	-	SLI	qn-1

z = bit width - 1

gn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

See the appropriate CPLD User Guide.

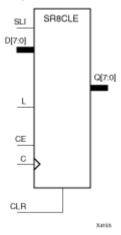


• See the appropriate CPLD Data Sheets.



## SR8CLE

Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with a shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and asynchronous clear (CLR). The register ignores clock transitions when (L) and (CE) are Low. The asynchronous (CLR), when High, overrides all other inputs and resets the data outputs (Q) Low. When (L) is High and (CLR) is Low, data on the Dn-D0 inputs is loaded into the corresponding Qn-Q0 bits of the register.

When (CE) is High and (L) and (CLR) are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the (Q0) output. During subsequent clock transitions, when (CE) is High and (L) and (CLR) are Low, the data shifts to the next highest bit position as new data is loaded into (Q)0 (for example, SLI® Q0, Q0® Q1, and Q1® Q2).

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), (L), and (CLR) inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

#### Logic Table

Inputs						Outputs	Outputs	
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1	
1	Х	Χ	Х	X	X	0	0	
0	1	X	X	Dn – D0	-	D0	Dn	
0	0	1	SLI	Х	-	SLI	qn-1	
0	0	0	Х	Х	Х	No Change	No Change	
z = bitwidt	h -1	-	-	•		•	•	

qn-1 = state of referenced output one setup time prior to active clock transition



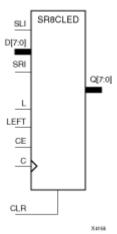
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **SR8CLED**

Macro: 8-Bit Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs: clock enable (CE), load enable (L), shift left/right (LEFT), and asynchronous clear (CLR). The register ignores clock transitions when (CE) and (L) are Low. The asynchronous clear, when High, overrides all other inputs and resets the data outputs (Qn) Low.

When (L) is High and (CLR) is Low, the data on the (D) inputs is loaded into the corresponding (Q) bits of the register. When (CE) is High and (L) and (CLR) are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on the SLI is loaded into (Q0) during the Low-to-High clock transition and shifted left (for example, to Q1 or Q2) during subsequent clock transitions. If LEFT is Low, data on the SRI is loaded into the last (Q) output during the Low-to-High clock transition and shifted right during subsequent clock transitions. The logic tables indicate the state of the (Q) outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied.

Inputs								Outputs		
CLR	L	CE	LEFT	SLI	SRI	D7 – D0	С	Q0	Q7	Q6 – Q1
1	Х	Х	Х	Х	Х	Х	Х	0	0	0
0	1	Х	Х	X	X	D7 – D0	-	D0	D7	Dn
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change
0	0	1	1	SLI	X	X	-	SLI	q6	qn-1
0	0	1	0	Х	SRI	Х	-	q1	SRI	qn+1
qn-1 or	qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition.									



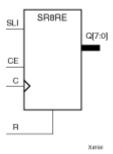
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **SR8RE**

Macro: 8-Bit Serial-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left serial input (SLI), parallel outputs (Qn), clock enable (CE), and synchronous reset (R) inputs. The R input, when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low.

When (CE) is High and (R) is Low, the data on the (SLI) is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the (Q0) output. During subsequent Low-to-High clock transitions, when (CE) is High and R is Low, data shifts to the next highest bit position as new data is loaded into (Q0) (for example, SLI® Q0, Q0® Q1, and Q1® Q2). The register ignores clock transitions when (CE) is Low.

Registers can be cascaded by connecting the last (Q) output of one stage to the SLI input of the next stage and connecting clock, (CE), and (R) in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

## **Logic Table**

Inputs		Outputs			
R	CE	SLI	С	Q0	Qz - Q1
1	Х	X	-	0	0
0	0	Χ	Χ	No Change	No Change
0	1	SLI	-	SLI	qn-1

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition

### **Design Entry Method**

This design element is only for use in schematics.

#### For More Information

• See the appropriate CPLD User Guide.

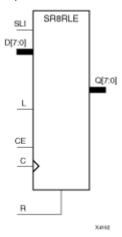


• See the appropriate CPLD Data Sheets.



## **SR8RLE**

Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and synchronous reset (R). The register ignores clock transitions when (L) and (CE) are Low. The synchronous (R), when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low. When (L) is High and (R) is Low during the Low-to-High clock transition, data on the (D) inputs is loaded into the corresponding Q bits of the register.

When (CE) is High and (L) and (R) are Low, data on the (SLI) input is loaded into the first bit of the shift register during the Low-to-High clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when (CE) is High and (L) and (R) are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, (CE), (L), and (R) inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied.

Inputs			Outputs				
R	L	CE	SLI	Dz – D0	С	Q0	Qz - Q1
1	X	X	X	X	-	0	0
0	1	X	X	Dz – D0	-	D0	Dn
0	0	1	SLI	Х	-	SLI	qn-1



Inputs		Outputs							
R	L CE SLI Dz – D0 C					Q0	Qz - Q1		
0	0 0 X X X No Change								
z = bitwidth -1									
qn-1 = state of referenced output one setup time prior to active clock transition									

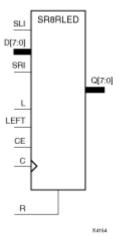
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **SR8RLED**

Macro: 8-Bit Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

This design element is a shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q) and four control inputs — clock enable (CE), load enable (L), shift left/right (LEFT), and synchronous reset (R). The register ignores clock transitions when (CE) and (L) are Low. The synchronous (R), when High, overrides all other inputs during the Low-to-High clock (C) transition and resets the data outputs (Q) Low. When (L) is High and (R) is Low during the Low-to-High clock transition, the data on the (D) inputs is loaded into the corresponding (Q) bits of the register.

When (CE) is High and (L) and (R) are Low, data shifts right or left, depending on the state of the LEFT input. If LEFT is High, data on (SLI) is loaded into (Q0) during the Low-to-High clock transition and shifted left (for example, to Q1 and Q2) during subsequent clock transitions. If LEFT is Low, data on the (SRI) is loaded into the last (Q) output during the Low-to-High clock transition and shifted right) during subsequent clock transitions. The logic tables below indicates the state of the (Q) outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied.

Inputs		Outputs								
R	L	CE	LEFT	SLI	SRI	D7- D0	С	Q0	Q7	Q6 – Q1
1	Χ	Χ	X	X	X	Χ	-	0	0	0
0	1	Х	X	X	X	D7 – D0	-	D0	D7	Dn
0	0	0	X	X	Х	Х	Х	No Change	No Change	No Change
0	0	1	1	SLI	Х	Х	-	SLI	q6	qn-1



Inputs	Inputs									
R	L	CE	LEFT	SLI	SRI	D7- D0	С	Q0	Q7	Q6 - Q1
0	0	1	0	X	SRI	X	-	q1	SRI	qn+1
qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition										

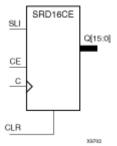
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD16CE

Macro: 16-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with a shift-left serial input (SLI), parallel outputs (Q), clock enable (CE) and asynchronous clear (CLR) inputs. The CLR input, when High, overrides all other inputs and resets the data outputs (Q) Low. When CE is High and CLR is Low, the data on the SLI input is loaded into the first bit of the shift register during the Low-to-High (or High-to-Low) clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and CLR is Low, data shifts to the next highest bit position as new data is loaded into Q0. The register ignores clock transitions when CE is Low.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, and CLR in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. .

The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs				Outputs	
CLR	CE	SLI	SLI C		Qz - Q1
1	X	X	X	0	0
0	0	X	X	No Change	No Change
0	1	1	-	1	qn-1
0	1	1	-	1	qn-1
0	1	0	-	0	qn-1
0	1	0	-	0	qn-1

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

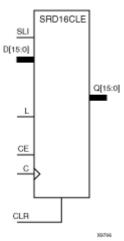


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD16CLE

Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with a shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and asynchronous clear (CLR). The register ignores clock transitions when L and CE are Low. The asynchronous CLR, when High, overrides all other inputs and resets the data outputs (Q) Low. When L is High and CLR is Low, data on the Dn – D0 inputs is loaded into the corresponding Qn – Q0 bits of the register. When CE is High and L and CLR are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High (or High-to-Low) clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and L and CLR are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output) of one stage to the SLI input of the next stage and connecting clock, CE, L, and CLR inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs				Outputs			
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1
1	X	Χ	Х	X	X	0	0
0	1	Х	X	Dn – D0	-	D0	Dn
0	1	Х	Х	Dn – D0	-	D0	Dn
0	0	1	SLI	Х	-	SLI	qn-1
0	0	1	SLI	Х	-	SLI	qn-1



Inputs		Outputs								
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1			
0	0 0 X X X No Change No Change									
z = bitwidth	z = bitwidth -1									
qn-1 = state o	qn-1 = state of referenced output one setup time prior to active clock transition									

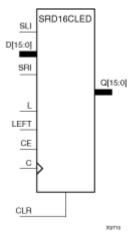
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD16CLED

Macro: 16-Bit Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

### Introduction

This design element is a dual edge triggered shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs: clock enable (CE), load enable (L), shift left/right (LEFT), and asynchronous clear (CLR). The register ignores clock transitions when CE and L are Low. The asynchronous clear, when High, overrides all other inputs and resets the data outputs (Qn) Low. When L is High and CLR is Low, the data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and CLR are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on the SLI is loaded into Q0 during the Low-to-High or High-to-Low clock transition and shifted left during subsequent clock transitions. If LEFT is Low, data on the SRI is loaded into the last Q output during the Low-to-High or High-to-Low clock transition and shifted right during subsequent clock transitions. The logic table indicates the state of the Q outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs								Outputs		
CLR	L	CE	LEFT	SLI	SRI	D15 - D0	С	Q0	Q15	Q14 - Q1
1	Х	Х	Х	Х	Х	Х	Х	0	0	0
0	1	Х	Х	Х	Х	D15-D0	-	D0	D15	Dn
0	1	Х	Х	Х	Х	D15-D0	-	D0	D15	Dn
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change
0	0	1	1	SLI	Х	Х	-	SLI	q14	qn-1
0	0	1	1	SLI	Х	Х	-	SLI	q14	qn-1



Inputs	nputs									Outputs		
CLR	L	CE	LEFT	SLI	SRI	D15 - D0	С	Q0	Q15	Q14 - Q1		
0	0	1	0	Х	SRI	Х	-	q1	SRI	qn+1		
0	0	1	0	Х	SRI	Х	-	q1	SRI	qn+1		
qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition.												

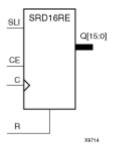
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### SRD16RE

Macro: 16-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left serial input (SLI), parallel outputs (Qn), clock enable (CE), and synchronous reset (R) inputs. The R input, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When CE is High and R is Low, the data on the SLI is loaded into the first bit of the shift register during the Low-to-High clock or High-to-Low (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and R is Low, data shifts to the next highest bit position as new data is loaded into Q0. The register ignores clock transitions when CE is Low.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, and R in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs				Outputs	
R	CE	SLI	С	Q0	Qz – Q1
1	X	X	-	0	0
1	X	Х	-	0	0
0	0	Х	X	No Change	No Change
0	1	1	-	1	qn-1
0	1	1	-	1	qn-1
0	1	0	-	0	qn-1
0	1	0	-	0	qn-1

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition



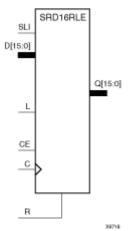
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD16RLE

Macro: 16-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and synchronous reset (R). The register ignores clock transitions when L and CE are Low. The synchronous R, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When L is High and R is Low, data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and R are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High or High-to-Low clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and L and R are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, L, and R inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs						Outputs		
R	L	CE	SLI	Dz – D0	С	Q0	Qz - Q1	
1	X	Χ	Х	X	-	0	0	
1	X	Χ	X	X	-	0	0	
0	1	Χ	X	Dz – D0	-	D0	Dn	
0	1	Χ	X	Dz – D0	-	D0	Dn	
0	0	1	SLI	X	-	SLI	qn-1	
0	0	1	SLI	X	-	SLI	qn-1	



Inputs						Outputs			
R	R L CE SLI Dz – D0 C					Q0 Qz - Q1			
0	0	0	Χ	No Change	No Change				
z = bitw	z = bitwidth -1								

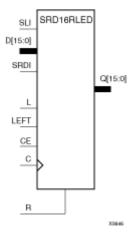
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD16RLED

Macro: 16-Bit Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only: CoolRunner-II

### Introduction

This design element is a dual edge triggered shift register with shift-left (SLI) and shift-right (SRDI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs — clock enable (CE), load enable (L), shift left/right (LEFT), and synchronous reset (R). The register ignores clock transitions when CE and L are Low. The synchronous R, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When L is High and R is Low, the data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and R are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on SLI is loaded into Q0 during the Low-to-High or High-to-Low clock transition and shifted left (to Q1, Q2, etc.) during subsequent clock transitions. If LEFT is Low, data on the SRDI is loaded into the last Q output during the Low-to-High or High-to-Low clock transition and shifted right during subsequent clock transitions. The logic table indicates the state of the Q outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs	S							Outputs		
R	L	CE	LEFT	SLI	SRDI	D15 - D0	С	Q0	Q15	Q14 - Q1
1	X	X	Х	X	X	X	-	0	0	0
1	Х	Х	Х	Х	Х	Х		0	0	0
0	1	Х	Х	Х	Х	D15-D0	-	D0	D15	Dn
0	1	Х	Х	Х	Х	D15-D0	-	D0	D15	Dn
0	0	0	X	Х	Х	X	Х	No Change	No Change	No Change
0	0	1	1	SLI	Х	Х	-	SLI	q14	qn-1



Inputs								Outputs	Outputs		
R	L	CE	LEFT	SLI	SRDI	D15 - D0	С	Q0	Q15	Q14 - Q1	
0	0	1	1	SLI	Х	Х	-	SLI	q14	qn-1	
0	0	1	0	Х	SRDI	Х	-	q1	SRDI	qn+1	
0	0	1	0	Х	SRDI	Х	-	q1	SRDI	qn+1	
qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition											

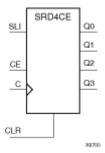
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD4CE

Macro: 4-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with a shift-left serial input (SLI), parallel outputs (Q), clock enable (CE) and asynchronous clear (CLR) inputs. The CLR input, when High, overrides all other inputs and resets the data outputs (Q) Low. When CE is High and CLR is Low, the data on the SLI input is loaded into the first bit of the shift register during the Low-to-High (or High-to-Low) clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and CLR is Low, data shifts to the next highest bit position as new data is loaded into Q0. The register ignores clock transitions when CE is Low.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, and CLR in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. .

The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs				Outputs	
CLR	CE	SLI	С	Q0	Qz - Q1
1	X	X	X	0	0
0	0	X	X	No Change	No Change
0	1	1	-	1	qn-1
0	1	1	-	1	qn-1
0	1	0	-	0	qn-1
0	1	0	-	0	qn-1

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

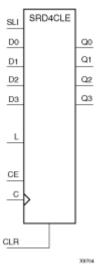


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD4CLE

Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with a shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and asynchronous clear (CLR). The register ignores clock transitions when L and CE are Low. The asynchronous CLR, when High, overrides all other inputs and resets the data outputs (Q) Low. When L is High and CLR is Low, data on the Dn – D0 inputs is loaded into the corresponding Qn – Q0 bits of the register. When CE is High and L and CLR are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High (or High-to-Low) clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and L and CLR are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output) of one stage to the SLI input of the next stage and connecting clock, CE, L, and CLR inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs						Outputs		
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz – Q1	
1	X	Х	X	X	X	0	0	
0	1	Χ	X	Dn – D0	-	D0	Dn	
0	1	Х	X	Dn – D0	-	D0	Dn	
0	0	1	SLI	Х	-	SLI	qn-1	
0	0	1	SLI	Х	-	SLI	qn-1	



Inputs						Outputs			
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1		
0 0 X X X No Change No Change							No Change		
z = bitwidth -1									
an-1 = state o	an-1 = state of referenced output one setup time prior to active clock transition								

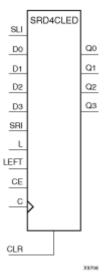
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD4CLED

Macro: 4-Bit Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs: clock enable (CE), load enable (L), shift left/right (LEFT), and asynchronous clear (CLR). The register ignores clock transitions when CE and L are Low. The asynchronous clear, when High, overrides all other inputs and resets the data outputs (Qn) Low. When L is High and CLR is Low, the data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and CLR are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on the SLI is loaded into Q0 during the Low-to-High or High-to-Low clock transition and shifted left during subsequent clock transitions. If LEFT is Low, data on the SRI is loaded into the last Q output during the Low-to-High or High-to-Low clock transition and shifted right during subsequent clock transitions. The logic table indicates the state of the Q outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs								Outputs		
CLR	L	CE	LEFT	SLI	SRI	D3 – D0	С	Q0	Q3	Q2 – Q1
1	Х	Χ	Χ	Χ	Χ	Χ	Χ	0	0	0
0	1	Χ	X	Χ	X	D3- D0	-	D0	D3	Dn
0	1	Х	Х	Х	X	D3- D0	-	D0	D3	Dn
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change
0	0	1	1	SLI	X	X	-	SLI	q2	qn-1



Inputs								Outputs	Outputs		
CLR	L	CE	LEFT	SLI	SRI	D3 – D0	С	Q0	Q3	Q2 – Q1	
0	0	1	1	SLI	Х	Х	-	SLI	q2	qn-1	
0	0	1	0	Χ	SRI	X	-	q1	SRI	qn+1	
0	0	1	0	Х	SRI	Х	-	q1	SRI	qn+1	
qn-1 and	qn-1 and qn+1 = state of referenced output one setup time prior to active clock transition										

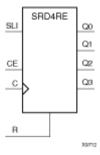
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### SRD4RE

Macro: 4-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left serial input (SLI), parallel outputs (Qn), clock enable (CE), and synchronous reset (R) inputs. The R input, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When CE is High and R is Low, the data on the SLI is loaded into the first bit of the shift register during the Low-to-High clock or High-to-Low (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and R is Low, data shifts to the next highest bit position as new data is loaded into Q0. The register ignores clock transitions when CE is Low.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, and R in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs				Outputs	
R	CE	SLI	С	Q0	Qz - Q1
1	X	X	-	0	0
1	X	X	-	0	0
0	0	Х	X	No Change	No Change
0	1	1	-	1	qn-1
0	1	1	-	1	qn-1
0	1	0	-	0	qn-1
0	1	0	-	0	qn-1

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition



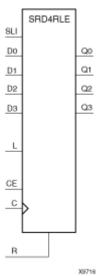
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD4RLE

Macro: 4-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and synchronous reset (R). The register ignores clock transitions when L and CE are Low. The synchronous R, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When L is High and R is Low, data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and R are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High or High-to-Low clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and L and R are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, L, and R inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs				Outputs			
R	L	CE	SLI	Dz – D0	С	Q0	Qz - Q1
1	X	Χ	Х	X	-	0	0
1	X	Χ	X	X	-	0	0
0	1	Χ	X	Dz – D0	-	D0	Dn
0	1	Χ	X	Dz – D0	-	D0	Dn



Inputs						Outputs		
R	L	CE	SLI	Dz – D0	С	Q0	Qz – Q1	
0	0	1	SLI	X	-	SLI	qn-1	
0	0	1	SLI	X	-	SLI	qn-1	
0	0	0	Х	Х	Х	No Change	No Change	
z = bitwidth -1								

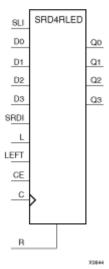
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### SRD4RLED

Macro: 4-Bit Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left (SLI) and shift-right (SRDI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs — clock enable (CE), load enable (L), shift left/right (LEFT), and synchronous reset (R). The register ignores clock transitions when CE and L are Low. The synchronous R, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When L is High and R is Low, the data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and R are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on SLI is loaded into Q0 during the Low-to-High or High-to-Low clock transition and shifted left (to Q1, Q2, etc.) during subsequent clock transitions. If LEFT is Low, data on the SRDI is loaded into the last Q output during the Low-to-High or High-to-Low clock transition and shifted right during subsequent clock transitions. The logic table indicates the state of the Q outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs								Outputs		
R	L	CE	LEFT	SLI	SRDI	D3 – D0	С	Q0	Q3	Q2 – Q1
1	X	X	X	X	X	X-	-	0	0	0
1	X	X	X	X	X	Х	-	0	0	0
0	1	X	X	X	X	D3 – D0	-	D0	D3	Dn
0	1	X	X	X	X	D3 – D0	-	D0	D3	Dn
0	0	0	X	X	X	Х	Х	No Change	No Change	No Change



Inputs								Outputs		
R	L	CE	LEFT	SLI	SRDI	D3 – D0	С	Q0	Q3	Q2 – Q1
0	0	1	1	SLI	X	X	-	SLI	q2	qn-1
0	0	1	1	SLI	X	X	-	SLI	q2	qn-1
0	0	1	0	X	SRDI	X	-	q1	SRDI	qn+1
0	0	1	0	X	SRDI	X	-	q1	SRDI	qn+1

qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

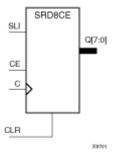
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD8CE

Macro: 8-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with a shift-left serial input (SLI), parallel outputs (Q), clock enable (CE) and asynchronous clear (CLR) inputs. The CLR input, when High, overrides all other inputs and resets the data outputs (Q) Low. When CE is High and CLR is Low, the data on the SLI input is loaded into the first bit of the shift register during the Low-to-High (or High-to-Low) clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and CLR is Low, data shifts to the next highest bit position as new data is loaded into Q0. The register ignores clock transitions when CE is Low.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, and CLR in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. .

The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs			Outputs	Outputs		
CLR	CE	SLI	С	Q0	Qz - Q1	
1	X	X	X	0	0	
0	0	X	X	No Change	No Change	
0	1	1	-	1	qn-1	
0	1	1	-	1	qn-1	
0	1	0	-	0	qn-1	
0	1	0	-	0	qn-1	

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition

## **Design Entry Method**

This design element is only for use in schematics.

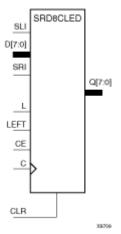


- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD8CLE

Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with a shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and asynchronous clear (CLR). The register ignores clock transitions when L and CE are Low. The asynchronous CLR, when High, overrides all other inputs and resets the data outputs (Q) Low. When L is High and CLR is Low, data on the Dn – D0 inputs is loaded into the corresponding Qn – Q0 bits of the register. When CE is High and L and CLR are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High (or High-to-Low) clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and L and CLR are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output) of one stage to the SLI input of the next stage and connecting clock, CE, L, and CLR inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

Inputs		Outputs					
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1
1	Χ	Χ	Χ	X	X	0	0
0	1	Χ	X	Dn – D0	-	D0	Dn
0	1	X	X	Dn – D0	-	D0	Dn
0	0	1	SLI	Х	-	SLI	qn-1
0	0	1	SLI	Х	=	SLI	qn-1



Inputs		Outputs						
CLR	L	CE	SLI	Dn – D0	С	Q0	Qz - Q1	
0 0 0 X X X No Change No Change						No Change		
z = bitwidth -1								
qn-1 = state o	qn-1 = state of referenced output one setup time prior to active clock transition							

# **Design Entry Method**

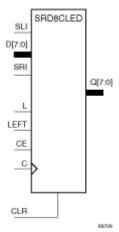
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# SRD8CLED

Macro: 8-Bit Dual Edge Triggered Shift Register with Clock Enable and Asynchronous Clear



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left (SLI) and shift-right (SRI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs: clock enable (CE), load enable (L), shift left/right (LEFT), and asynchronous clear (CLR). The register ignores clock transitions when CE and L are Low. The asynchronous clear, when High, overrides all other inputs and resets the data outputs (Qn) Low. When L is High and CLR is Low, the data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and CLR are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on the SLI is loaded into Q0 during the Low-to-High or High-to-Low clock transition and shifted left during subsequent clock transitions. If LEFT is Low, data on the SRI is loaded into the last Q output during the Low-to-High or High-to-Low clock transition and shifted right during subsequent clock transitions. The logic table indicates the state of the Q outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

# Logic Table

Inputs	Inputs									Outputs		
CLR	L	CE	LEFT	SLI	SRI	D7 – D0	CLK	Q0	Q7	Q6 - Q1		
1	Х	Х	X	X	Х	X	X	0	0	0		
0	1	X	X	X	X	D7 – D0	-	D0	D7	Dn		
0	1	Х	Х	Х	Х	D7 – D0	-	D0	D7	Dn		
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change		
0	0	1	1	SLI	X	Χ	-	SLI	q6	qn-1		
0	0	1	1	SLI	Х	Х	Ø	SLI	q6	qn-1		
0	0	1	0	X	SRI	Χ	-	q1	SRI	qn+1		



Inputs	Inputs							Outputs			
CLR	L	CE	LEFT	SLI	SRI	D7 – D0	CLK	Q0	Q7	Q6 - Q1	
0 0 1 0 X SRI X - q1 SRI qn+1											
qn-1 or qr	qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition										

# **Design Entry Method**

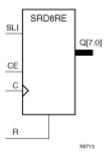
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## SRD8RE

Macro: 8-Bit Serial-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



# **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left serial input (SLI), parallel outputs (Qn), clock enable (CE), and synchronous reset (R) inputs. The R input, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When CE is High and R is Low, the data on the SLI is loaded into the first bit of the shift register during the Low-to-High clock or High-to-Low (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and R is Low, data shifts to the next highest bit position as new data is loaded into Q0. The register ignores clock transitions when CE is Low.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, and R in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs		Outputs			
R	CE	SLI	С	Q0	Qz - Q1
1	X	X	-	0	0
1	X	X	-	0	0
0	0	X	X	No Change	No Change
0	1	1	-	1	qn-1
0	1	1	-	1	qn-1
0	1	0	-	0	qn-1
0	1	0	-	0	qn-1

z = bitwidth -1

qn-1 = state of referenced output one setup time prior to active clock transition



# **Design Entry Method**

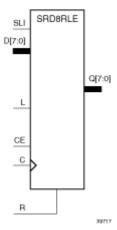
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# SRD8RLE

Macro: 8-Bit Loadable Serial/Parallel-In Parallel-Out Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left serial input (SLI), parallel inputs (D), parallel outputs (Q), and three control inputs: clock enable (CE), load enable (L), and synchronous reset (R). The register ignores clock transitions when L and CE are Low. The synchronous R, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When L is High and R is Low, data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and R are Low, data on the SLI input is loaded into the first bit of the shift register during the Low-to-High or High-to-Low clock (C) transition and appears on the Q0 output. During subsequent clock transitions, when CE is High and L and R are Low, the data shifts to the next highest bit position as new data is loaded into Q0.

Registers can be cascaded by connecting the last Q output of one stage to the SLI input of the next stage and connecting clock, CE, L, and R inputs in parallel.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

## Logic Table

Inputs	Inputs						
R	L	CE	SLI	Dz – D0	С	Q0	Qz - Q1
1	X	Χ	Х	X	-	0	0
1	X	Χ	Χ	X	-	0	0
0	1	Χ	X	Dz – D0	-	D0	Dn
0	1	Χ	X	Dz – D0	-	D0	Dn
0	0	1	SLI	X	-	SLI	qn-1
0	0	1	SLI	X	-	SLI	qn-1



Inputs	Inputs					Outputs		
R	L	CE	SLI	Dz - D0	С	Q0	Qz - Q1	
0	0	0	Χ	X	X	No Change	No Change	
z = bitw	z = bitwidth -1							

# **Design Entry Method**

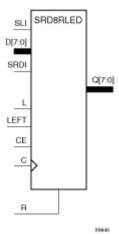
This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# SRD8RLED

Macro: 8-Bit Dual Edge Triggered Shift Register with Clock Enable and Synchronous Reset



## **Supported Architectures**

This design element is supported in the following architectures only:

CoolRunner-II

#### Introduction

This design element is a dual edge triggered shift register with shift-left (SLI) and shift-right (SRDI) serial inputs, parallel inputs (D), parallel outputs (Q), and four control inputs — clock enable (CE), load enable (L), shift left/right (LEFT), and synchronous reset (R). The register ignores clock transitions when CE and L are Low. The synchronous R, when High, overrides all other inputs during the Low-to-High or High-to-Low clock (C) transition and resets the data outputs (Q) Low. When L is High and R is Low, the data on the D inputs is loaded into the corresponding Q bits of the register. When CE is High and L and R are Low, data is shifted right or left, depending on the state of the LEFT input. If LEFT is High, data on SLI is loaded into Q0 during the Low-to-High or High-to-Low clock transition and shifted left (to Q1, Q2, etc.) during subsequent clock transitions. If LEFT is Low, data on the SRDI is loaded into the last Q output during the Low-to-High or High-to-Low clock transition and shifted right during subsequent clock transitions. The logic table indicates the state of the Q outputs under all input conditions.

This register is asynchronously cleared, outputs Low, when power is applied. The power-on condition can be simulated by applying a High-level pulse on the PRLD global net.

### **Logic Table**

Inputs	Inputs								utputs		
R	L	CE	LEFT	SLI	SRDI	D7- D0	С	Q0	Q7	Q6 - Q1	
1	X	X	X	X	X	X	-	0	0	0	
1	X	Χ	Χ	X	X	X	-	0	0	0	
0	1	Х	Х	X	Х	D7 – D0	-	D0	D7	Dn	
0	1	Χ	Χ	X	X	D7 – D0	-	D0	D7	Dn	
0	0	0	Х	Х	Х	Х	Х	No Change	No Change	No Change	
0	0	1	1	SLI	X	Χ	-	SLI	q6	qn-1	



Inputs	Inputs									
R	L	CE	LEFT	SLI	SRDI	D7- D0	С	Q0	Q7	Q6 - Q1
0	0	1	1	SLI	X	X	-	SLI	q6	qn-1
0	0	1	0	X	SRDI	X	-	q1	SRDI	qn+1
0	0	1	0	Х	SRDI	X	-	q1	SRDI	qn+1
qn-1 or q	qn-1 or qn+1 = state of referenced output one setup time prior to active clock transition									

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



## **VCC**

### Primitive: VCC-Connection Signal Tag



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

This design element serves as a signal tag, or parameter, that forces a net or input function to a logic High level. A net tied to this element cannot have any other source.

When the placement and routing software encounters a net or input function tied to this element, it removes any logic that is disabled by the Vcc signal, which is only implemented when the disabled logic cannot be removed.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 2-Input XNOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



#### Primitive: 3-Input XNOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

### **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

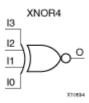
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input XNOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

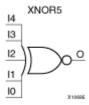
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input XNOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

## **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

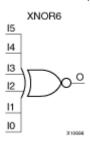
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 6-Input XNOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

## **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

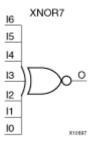
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 7-Input XNOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

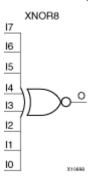
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 8-Input XNOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

### Logic Table

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

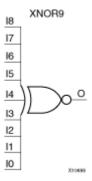
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 9-Input XNOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XNOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Logic Table**

Input	Output
I0 Iz	0
Odd number of 1	0
Even number of 1	1

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



# XOR<sub>2</sub>

### Primitive: 2-Input XOR Gate with Non-Inverted Inputs

XOR2 10 XOR2

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 3-Input XOR Gate with Non-Inverted Inputs

XOR3

# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

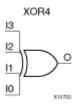
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 4-Input XOR Gate with Non-Inverted Inputs



# **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

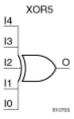
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Primitive: 5-Input XOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

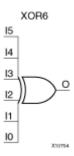
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 6-Input XOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

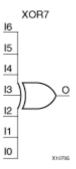
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 7-Input XOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

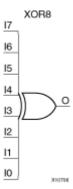
# **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 8-Input XOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

#### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

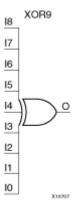
## **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.



### Macro: 9-Input XOR Gate with Non-Inverted Inputs



## **Supported Architectures**

This design element is supported in the following architectures only:

- XC9500XL
- CoolRunner XPLA3
- CoolRunner-II

### Introduction

XOR functions of up to nine inputs are available. All inputs are non-inverting. Because each input uses a CLB resource, replace functions with unused inputs with functions having the necessary number of inputs.

### **Design Entry Method**

This design element is only for use in schematics.

- See the appropriate CPLD User Guide.
- See the appropriate CPLD Data Sheets.